Data Sheet

ADAU1761

SigmaDSP Stereo, Low Power, 96 kHz, 24-Bit Audio Codec with Integrated PLL

FEATURES

- SigmaDSP 28-/56-bit, 50 MIPS digital audio processor
- Fully programmable with SigmaStudio graphical tool
- 24-bit stereo audio ADC and DAC: >98 dB SNR
- Sampling rates from 8 kHz to 96 kHz
- Low power: 7 mW record, 7 mW playback, 48 kHz at 1.8 V
- 6 analog input pins, configurable for single-ended or differential inputs
- Flexible analog input/output mixers
- Stereo digital microphone input
- Analog outputs: 2 differential stereo, 2 single-ended stereo, 1 mono headphone output driver
- PLL supporting input clocks from 8 MHz to 27 MHz
- Analog automatic level control (ALC)
- Microphone bias reference voltage
- Analog and digital I/O: 1.8 V to 3.65 V
- I^2C and SPI control interfaces
- Digital audio serial data I/O: stereo and time-division multiplexing (TDM) modes
- Software-controllable clickless mute
- Software power-down
- GPIO pins for digital controls and outputs
- 32-lead, 5 mm × 5 mm LFCSP
- −40°C to +85°C operating temperature range

APPLICATIONS

- Smartphones/multimedia phones
- Digital still cameras/digital video cameras
- Portable media players/portable audio players
- Phone accessories products

GENERAL DESCRIPTION

The ADAU1761 is a low power, stereo audio codec with integrated digital audio processing that supports stereo 48 kHz record and playback at 14 mW from a 1.8 V analog supply. The stereo audio ADCs and DACs support sample rates from 8 kHz to 96 kHz as well as a digital volume control.

The SigmaDSP® core features 28-bit processing (56-bit double precision). The processor allows system designers to compensate for the real-world limitations of microphones, speakers, amplifiers, and listening environments, resulting in a dramatic improvement in the perceived audio quality through equalization, multiband compression, limiting, and third-party branded algorithms.

The SigmaStudio™ graphical development tool is used to program the ADAU1761. This software includes audio processing blocks such as filters, dynamics processors, mixers, and low level DSP functions for fast development of custom signal flows.

The record path includes an integrated microphone bias circuit and six inputs. The inputs can be mixed and muxed before the ADC, or they can be configured to bypass the ADC. The ADAU1761 includes a stereo digital microphone input.

The ADAU1761 includes five high power output drivers (two differential and three single-ended), supporting stereo headphones, an earpiece, or other output transducer. AC-coupled or capless configurations are supported. Individual fine level controls are supported on all analog outputs. The output mixer stage allows for flexible routing of audio.

FUNCTIONAL BLOCK DIAGRAM

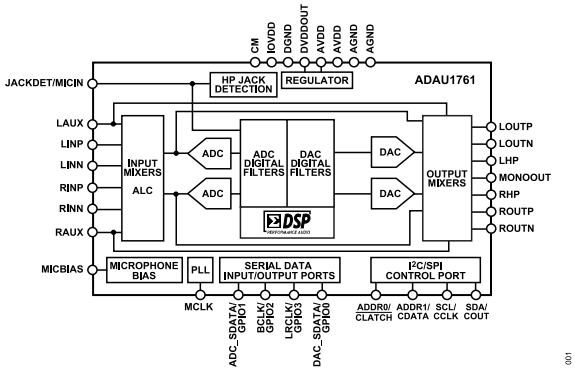


Figure 1.

Rev. F

DOCUMENT FEEDBACK

TECHNICAL SUPPORT

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REVISION HISTORY

8/2024—Rev. E to Rev. F

Changed Master to Main and Slave to Subordinate (Throughout) ... 1
Updated Outline Dimensions ... 92
Changes to Ordering Guide ... 92

SPECIFICATIONS

Supply voltage (AVDD) = 3.3 V, T_A = 25°C, main clock = 12.288 MHz (48 kHz f_S, 256 × f_S mode), input sample rate = 48 kHz, measurement bandwidth = 20 Hz to 20 kHz, word width = 24 bits, C_{LOAD} (digital output) = 20 pF, I_{LOAD} (digital output) = 2 mA, V_{IH} = 2 V, V_{IL} = 0.8 V, unless otherwise noted. Performance of all channels is identical, exclusive of the interchannel gain mismatch and interchannel phase deviation specifications.

ANALOG PERFORMANCE SPECIFICATIONS

Specifications guaranteed at 25°C (ambient).

Table 1.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
ANALOG-TO-DIGITAL CONVERTERS	ADC performance excludes mixers and PGA				
ADC Resolution	All ADCs		24		Bits
Digital Attenuation Step			0.375		dB
Digital Attenuation Range			95		dB
INPUT RESISTANCE					
Single-Ended Line Input	−12 dB gain		83		kΩ
	0 dB gain		21		kΩ
	6 dB gain		10.5		kΩ
PGA Inverting Inputs	−12 dB gain		84.5		kΩ
	0 dB gain		53		kΩ
	35.25 dB gain		2		kΩ
PGA Noninverting Inputs	All gains		105		kΩ
SINGLE-ENDED LINE INPUT					
Full-Scale Input Voltage (0 dB)	Scales linearly with AVDD		AVDD/3.3		V rms
	AVDD = 1.8 V		0.55 (1.56)		V rms (V p-p)
	AVDD = 3.3 V		1.0 (2.83)		V rms (V p-p)
Dynamic Range	20 Hz to 20 kHz, −60 dB input				
With A-Weighted Filter (RMS)	AVDD = 1.8 V		94		dB
	AVDD = 3.3 V		99		dB
No Filter (RMS)	AVDD = 1.8 V		91		dB
	AVDD = 3.3 V		96		dB
Total Harmonic Distortion + Noise	−1 dBFS				
	AVDD = 1.8 V		−88		dB
	AVDD = 3.3 V		−90		dB
Signal-to-Noise Ratio					
With A-Weighted Filter (RMS)	AVDD = 1.8 V		94		dB
	AVDD = 3.3 V		99		dB
No Filter (RMS)	AVDD = 1.8 V		91		dB
	AVDD = 3.3 V		96		dB
Gain per Step			3		dB
Total Gain Range		−12		+6	dB
Mute Attenuation			−87		dB
Interchannel Gain Mismatch			0.005		dB
Offset Error			0		mV
Gain Error			−12		%
Interchannel Isolation			68		dB
Power Supply Rejection Ratio	CM capacitor = 20 µF				
	100 mV p-p @ 217 Hz		65		dB
	100 mV p-p @ 1 kHz		67		dB

SPECIFICATIONS

Table 1. (Continued)

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
PSEUDO-DIFFERENTIAL PGA INPUT					
Full-Scale Input Voltage (0 dB)	Scales linearly with AVDD		AVDD/3.3		V rms
	AVDD = 1.8 V		0.55 (1.56)		V rms (V p-p)
	AVDD = 3.3 V		1.0 (2.83)		V rms (V p-p)
Dynamic Range	20 Hz to 20 kHz, −60 dB input				
With A-Weighted Filter (RMS)	AVDD = 1.8 V		92		dB
	AVDD = 3.3 V		98		dB
No Filter (RMS)	AVDD = 1.8 V		90		dB
	AVDD = 3.3 V		95		dB
Total Harmonic Distortion + Noise	−1 dBFS				
	AVDD = 1.8 V		−88		dB
	AVDD = 3.3 V		−89		dB
Signal-to-Noise Ratio					
With A-Weighted Filter (RMS)	AVDD = 1.8 V		92		dB
	AVDD = 3.3 V		98		dB
No Filter (RMS)	AVDD = 1.8 V		90		dB
	AVDD = 3.3 V		95		dB
Volume Control Step	PGA gain		0.75		dB
Volume Control Range	PGA gain	−12		+35.25	dB
PGA Boost			20		dB
Mute Attenuation			−87		dB
Interchannel Gain Mismatch			0.005		dB
Offset Error			0		mV
Gain Error			−14		%
Interchannel Isolation			83		dB
Common-Mode Rejection Ratio	100 mV rms, 1 kHz		65		dB
	100 mV rms, 20 kHz		65		dB
FULL DIFFERENTIAL PGA INPUT	Differential PGA inputs				
Full-Scale Input Voltage (0 dB)	Scales linearly with AVDD		AVDD/3.3		V rms
	AVDD = 1.8 V		0.55 (1.56)		V rms (V p-p)
	AVDD = 3.3 V		1.0 (2.83)		V rms (V p-p)
Dynamic Range	20 Hz to 20 kHz, −60 dB input				
With A-Weighted Filter (RMS)	AVDD = 1.8 V		92		dB
	AVDD = 3.3 V		98		dB
No Filter (RMS)	AVDD = 1.8 V		90		dB
	AVDD = 3.3 V		95		dB
Total Harmonic Distortion + Noise	−1 dBFS				
	AVDD = 1.8 V		−70		dB
	AVDD = 3.3 V		−78		dB
Signal-to-Noise Ratio					
With A-Weighted Filter (RMS)	AVDD = 1.8 V		92		dB
	AVDD = 3.3 V		98		dB
No Filter (RMS)	AVDD = 1.8 V		90		dB
	AVDD = 3.3 V		95		dB
Volume Control Step	PGA gain		0.75		dB
Volume Control Range	PGA gain	−12		+35.25	dB
PGA Boost			20		dB
Mute Attenuation			−87		dB
Interchannel Gain Mismatch			0.005		dB

SPECIFICATIONS

Table 1. (Continued)

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
Offset Error			0		mV
Gain Error			−14		%
Interchannel Isolation			83		dB
Common-Mode Rejection Ratio	100 mV rms, 1 kHz		65		dB
	100 mV rms, 20 kHz		65		dB
MICROPHONE BIAS	MBIEN = 1				
Bias Voltage					
0.65 × AVDD	AVDD = 1.8 V, MBI = 1		1.17		V
	AVDD = 3.3 V, MBI = 1		2.145		V
0.90 × AVDD	AVDD = 1.8 V, MBI = 0		1.62		V
	AVDD = 3.3 V, MBI = 0		2.97		V
Bias Current Source	AVDD = 3.3 V, MBI = 0, MPERF = 1			3	mA
Noise in the Signal Bandwidth	AVDD = 3.3 V, 1 kHz to 20 kHz				
	MBI = 0, MPERF = 0		42		nV/√Hz
	MBI = 0, MPERF = 1		85		nV/√Hz
	MBI = 1, MPERF = 0		25		nV/√Hz
	MBI = 1, MPERF = 1		37		nV/√Hz
DIGITAL-TO-ANALOG CONVERTERS	DAC performance excludes mixers and headphone amplifier				
DAC Resolution	All DACs		24		Bits
Digital Attenuation Step			0.375		dB
Digital Attenuation Range			95		dB
DAC TO LINE OUTPUT					
Full-Scale Output Voltage (0 dB)	Scales linearly with AVDD		AVDD/3.3		V rms
	AVDD = 1.8 V		0.50 (1.41)		V rms (V p-p)
	AVDD = 3.3 V		0.92 (2.60)		V rms (V p-p)
Analog Volume Control Step	Line output volume control		0.75		dB
Analog Volume Control Range	Line output volume control	−57	1	+6	dB
Mute Attenuation			−87		dB
Dynamic Range	20 Hz to 20 kHz, −60 dB input, line output mode				
With A-Weighted Filter (RMS)	AVDD = 1.8 V		96		dB
	AVDD = 3.3 V		101		dB
No Filter (RMS)	AVDD = 1.8 V		93.5		dB
	AVDD = 3.3 V		98		dB
Total Harmonic Distortion + Noise	−1 dBFS, line output mode				dB
	AVDD = 1.8 V		−90		dB
	AVDD = 3.3 V		−92		dB
Signal-to-Noise Ratio	Line output mode				
With A-Weighted Filter (RMS)	AVDD = 1.8 V		96		dB
	AVDD = 3.3 V		101		dB
No Filter (RMS)	AVDD = 1.8 V		93.5		dB
	AVDD = 3.3 V		98		dB
Power Supply Rejection Ratio	CM capacitor = 20 µF				
	100 mV p-p @ 217 Hz		56		dB
	100 mV p-p @ 1 kHz		70		dB
Gain Error			3		%
Interchannel Gain Mismatch			0.005		dB
Offset Error			0		mV
Interchannel Isolation	1 kHz, 0 dBFS input signal		100		dB

SPECIFICATIONS

Table 1. (Continued)

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
DAC TO HEADPHONE/EARPIECE OUTPUT	P_O = output power per channel				
Full-Scale Output Voltage (0 dB)	Scales linearly with AVDD		AVDD/3.3		V rms
	AVDD = 1.8 V		0.50 (1.41)		V rms (V p-p)
	AVDD = 3.3 V		0.92 (2.60)		V rms (V p-p)
Total Harmonic Distortion + Noise	−4 dBFS				
16 Ω load	AVDD = 1.8 V, P_O = 6.4 mW		−76		dB
	AVDD = 3.3 V, P_O = 21.1 mW		−82		dB
32 Ω load	AVDD = 1.8 V, P_O = 3.8 mW		−82		dB
	AVDD = 3.3 V, P_O = 10.6 mW		−82		dB
Power Supply Rejection Ratio	CM capacitor = 20 µF				
	100 mV p-p @ 217 Hz		56		dB
	100 mV p-p @ 1 kHz		67		dB
Interchannel Isolation	1 kHz, 0 dBFS input signal, 32 Ω load, AVDD = 3.3 V				
	Referred to GND		73		dB
	Referred to CM (capless headphone mode)		50		dB
REFERENCE					
Common-Mode Reference Output	CM pin		AVDD/2		V

POWER SUPPLY SPECIFICATIONS

Table 2.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
SUPPLIES					
Voltage	DVDDOUT		1.56		V
	AVDD	1.8	3.3	3.65	V
	IOVDD	1.63	3.3	3.65	V
Digital I/O Current (IOVDD = 1.8 V)	20 pF capacitive load on all digital pins				
Subordinate Mode	f_S = 48 kHz		0.25		mA
	f_S = 96 kHz		0.48		mA
	f_S = 8 kHz		0.07		mA
Main Mode	f_S = 48 kHz		0.62		mA
	f_S = 96 kHz		1.23		mA
	f_S = 8 kHz		0.11		mA
Digital I/O Current (IOVDD = 3.3 V)	20 pF capacitive load on all digital pins				
Subordinate Mode	f_S = 48 kHz		0.48		mA
	f_S = 96 kHz		0.9		mA
	f_S = 8 kHz		0.13		mA
Main Mode	f_S = 48 kHz		1.51		mA
	f_S = 96 kHz		3		mA
	f_S = 8 kHz		0.27		mA
Analog Current (AVDD)	See Table 3				

SPECIFICATIONS

TYPICAL CURRENT CONSUMPTION

Main clock = 12.288 MHz, input sample rate = 48 kHz, input tone = 1 kHz, normal power management settings, ADC input @ −1 dBFS, DAC input @ 0 dBFS. For total power consumption, add the IOVDD current listed in Table 2.

Table 3.

Operating Voltage	Audio Path	Clock Generation	Typical AVDD Current Consumption (mA)
AVDD = IOVDD = 3.3 V	Record stereo differential to ADC	Direct MCLK	5.24
		Integer PLL	6.57
	DAC stereo playback to line output (10 kΩ)	Direct MCLK	5.55
		Integer PLL	6.90
	DAC stereo playback to headphone (16 Ω)	Direct MCLK	55.5
		Integer PLL	56.8
	DAC stereo playback to headphone (32 Ω)	Direct MCLK	30.9
		Integer PLL	32.25
	DAC stereo playback to capless headphone (32 Ω)	Direct MCLK	56.75
		Integer PLL	58
	Record aux stereo bypass to line output (10 kΩ)	Direct MCLK	1.9
		Integer PLL	3.3
AVDD = IOVDD = 1.8 V	Record stereo differential to ADC	Direct MCLK	4.25
		Integer PLL	5.55
	DAC stereo playback to line output (10 kΩ)	Direct MCLK	4.7
		Integer PLL	5.7
	DAC stereo playback to headphone (16 Ω)	Direct MCLK	30.81
		Integer PLL	32
	DAC stereo playback to headphone (32 Ω)	Direct MCLK	18.3
		Integer PLL	19.5
	DAC stereo playback to capless headphone (32 Ω)	Direct MCLK	32.6
		Integer PLL	33.7
	Record aux stereo bypass to line output (10 kΩ)	Direct MCLK	1.9
		Integer PLL	3.07

TYPICAL POWER MANAGEMENT MEASUREMENTS

Main clock = 12.288 MHz, integer PLL, input sample rate = 48 kHz, input tone = 1 kHz. Pseudo-differential input to ADCs, DACs to line output with 10 kΩ load. ADC input @ −1 dBFS, DAC input @ 0 dBFS. In Table 4, the mixer boost and power management conditions are set for MXBIAS[1:0], ADCBIAS[1:0], HPBIAS[1:0], and DACBIAS[1:0]. RBIAS[1:0] and PBIAS[1:0] do not have an extreme power saving mode and are therefore set for power saving mode in the extreme power saving rows in Table 4.

Table 4.

Operating Voltage	Power Management Setting	Mixer Boost Setting	Typical AVDD Current Consumption (mA)	Typical ADC THD + N (dB)	Typical Line Output THD + N (dB)
AVDD = IOVDD = 3.3 V	Normal (default)	Normal operation	9.6	−91	−92.5
		Boost Level 1	9.75	−91.5	−92.5
		Boost Level 2	9.92	−91.5	−92.5
		Boost Level 3	10.25	−91.5	−92.5
	Extreme power saving	Normal operation	7.09	−84.5	−87
		Boost Level 1	7.19	−84.8	−87.1
		Boost Level 2	7.29	−84.8	−87.1
		Boost Level 3	7.49	−85	−87.1

SPECIFICATIONS

Table 4. (Continued)

Operating Voltage	Power Management Setting	Mixer Boost Setting	Typical AVDD Current Consumption (mA)	Typical ADC THD + N (dB)	Typical Line Output THD + N (dB)
	Power saving	Normal operation	7.67	−89.5	−90
		Boost Level 1	7.77	−89.5	−90
		Boost Level 2	7.86	−89.8	−90
		Boost Level 3	8.07	−89.8	−90
	Enhanced performance	Normal operation	10.55	−91	−93.5
		Boost Level 1	10.74	−91	−93.5
		Boost Level 2	10.93	−91	−93.5
		Boost Level 3	11.33	−91	−93.5
AVDD = IOVDD = 1.8 V	Normal (default)	Normal operation	8.1	−88	−91.2
		Boost Level 1	8.26	−88	−91.2
		Boost Level 2	8.41	−88	−91.2
		Boost Level 3	8.73	−88	−91.2
	Extreme power saving	Normal operation	5.73	−85	−86
		Boost Level 1	5.82	−85.4	−86
		Boost Level 2	5.91	−85.5	−86
		Boost Level 3	6.1	−85.5	−86
	Power saving	Normal operation	6.27	−86	−89.4
		Boost Level 1	6.36	−86.1	−89.5
		Boost Level 2	6.46	−86.3	−89.5
		Boost Level 3	6.65	−86.3	−89.5
	Enhanced performance	Normal operation	9.01	−88	−91.5
		Boost Level 1	9.2	−88	−91.5
		Boost Level 2	9.38	−88	−91.5
		Boost Level 3	9.76	−88	−91.5

DIGITAL FILTERS

Table 5.

Parameter	Mode	Factor	Min	Typ	Max	Unit
ADC DECIMATION FILTER	All modes, typ @ 48 kHz					
Pass Band		0.4375 f_S		21		kHz
Pass-Band Ripple				±0.015		dB
Transition Band		0.5 f_S		24		kHz
Stop Band		0.5625 f_S		27		kHz
Stop-Band Attenuation				67		dB
Group Delay		22.9844/f_S		479		µs
DAC INTERPOLATION FILTER						
Pass Band	48 kHz mode, typ @ 48 kHz	0.4535 f_S		22		kHz
	96 kHz mode, typ @ 96 kHz	0.3646 f_S		35		kHz
Pass-Band Ripple	48 kHz mode, typ @ 48 kHz				±0.01	dB
	96 kHz mode, typ @ 96 kHz				±0.05	dB
Transition Band	48 kHz mode, typ @ 48 kHz	0.5 f_S		24		kHz
	96 kHz mode, typ @ 96 kHz	0.5 f_S		48		kHz
Stop Band	48 kHz mode, typ @ 48 kHz	0.5465 f_S		26		kHz
	96 kHz mode, typ @ 96 kHz	0.6354 f_S		61		kHz
Stop-Band Attenuation	48 kHz mode, typ @ 48 kHz			69		dB
	96 kHz mode, typ @ 96 kHz			68		dB

SPECIFICATIONS

Table 5. (Continued)

Parameter	Mode	Factor	Min	Typ	Max	Unit
Group Delay	48 kHz mode, typ @ 48 kHz	$25/f_S$		521		µs
	96 kHz mode, typ @ 96 kHz	$11/f_S$		115		µs

DIGITAL INPUT/OUTPUT SPECIFICATIONS

−40°C < T_A < +85°C, IOVDD = 3.3 V ± 10%.

Table 6.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
INPUT SPECIFICATIONS					
Input Voltage High (V_{IH})		0.7 × IOVDD			V
Input Voltage Low (V_{IL})				0.3 × IOVDD	V
Input Leakage					
Pull-Ups/Pull-Downs Disabled	I_{IH} @ V_{IH} = 3.3 V	−0.17		+0.17	µA
	I_{IL} @ V_{IL} = 0 V	−0.17		+0.17	µA
	I_{IL} @ V_{IL} = 0 V (MCLK pin)	−13.5		−0.5	µA
Pull-Ups Enabled	I_{IH} @ V_{IH} = 3.3 V	−0.7		+0.7	µA
	I_{IL} @ V_{IL} = 0 V	−13.5		−0.5	µA
Pull-Downs Enabled	I_{IH} @ V_{IH} = 3.3 V	2.7		8.3	µA
	I_{IL} @ V_{IL} = 0 V	−0.18		+0.18	µA
Input Capacitance				5	pF
OUTPUT SPECIFICATIONS					
Output Voltage High (V_{OH})	I_{OH} = 2 mA @ 3.3 V, 0.85 mA @ 1.8 V	0.8 × IOVDD			V
Output Voltage Low (V_{OL})	I_{OL} = 2 mA @ 3.3 V, 0.85 mA @ 1.8 V			0.1 × IOVDD	V

DIGITAL TIMING SPECIFICATIONS

−40°C < T_A < +85°C, IOVDD = 3.3 V ± 10%.

Table 7. Digital Timing

	Limit			
Parameter	t_{MIN}	t_{MAX}	Unit	Description
MAIN CLOCK				
t_{MP}	74	488	ns	MCLK period, 256 × f_S mode.
t_{MP}	37	244	ns	MCLK period, 512 × f_S mode.
t_{MP}	24.7	162.7	ns	MCLK period, 768 × f_S mode.
t_{MP}	18.5	122	ns	MCLK period, 1024 × f_S mode.
SERIAL PORT				
t_{BIL}	5		ns	BCLK pulse width low.
t_{BIH}	5		ns	BCLK pulse width high.
t_{LIS}	5		ns	LRCLK setup. Time to BCLK rising.
t_{LIH}	5		ns	LRCLK hold. Time from BCLK rising.
t_{SIS}	5		ns	DAC_SDATA setup. Time to BCLK rising.
t_{SIH}	5		ns	DAC_SDATA hold. Time from BCLK rising.
t_{SOD}		50	ns	ADC_SDATA delay. Time from BCLK falling in main or subordinate mode. Full range of IOVDD.
		25	ns	ADC_SDATA delay. Time from BCLK falling in main or subordinate mode. IOVDD = 3.3V ± 10%.
SPI PORT				
f_{CCLK}		10	MHz	CCLK frequency.

SPECIFICATIONS

Table 7. Digital Timing (Continued)

Parameter	Limit t_{MIN}	Limit t_{MAX}	Unit	Description
t_{CCPL}	10		ns	CCLK pulse width low.
t_{CCPH}	10		ns	CCLK pulse width high.
t_{CLS}	5		ns	$\overline{CLATCH}$ setup. Time to CCLK rising.
t_{CLH}	10		ns	$\overline{CLATCH}$ hold. Time from CCLK rising.
t_{CLPH}	10		ns	$\overline{CLATCH}$ pulse width high.
t_{CDS}	5		ns	CDATA setup. Time to CCLK rising.
t_{CDH}	5		ns	CDATA hold. Time from CCLK rising.
t_{COD}		50	ns	COUT three-stated. Time from $\overline{CLATCH}$ rising.
I^2C PORT				
f_{SCL}		400	kHz	SCL frequency.
t_{SCLH}	0.6		µs	SCL high.
t_{SCLL}	1.3		µs	SCL low.
t_{SCS}	0.6		µs	Setup time; relevant for repeated start condition.
t_{SCH}	0.6		µs	Hold time. After this period, the first clock is generated.
t_{DS}	100		ns	Data setup time.
t_{SCR}		300	ns	SCL rise time.
t_{SCF}		300	ns	SCL fall time.
t_{SDR}		300	ns	SDA rise time.
t_{SDF}		300	ns	SDA fall time.
t_{BFT}	0.6		µs	Bus-free time. Time between stop and start.
DIGITAL MICROPHONE				R_{LOAD} = 1 MΩ, C_{LOAD} = 14 pF.
t_{DCF}		10	ns	Digital microphone clock fall time.
t_{DCR}		10	ns	Digital microphone clock rise time.
t_{DDV}	22	30	ns	Digital microphone delay time for valid data.
t_{DDH}	0	12	ns	Digital microphone delay time for data three-stated.

DIGITAL TIMING DIAGRAMS

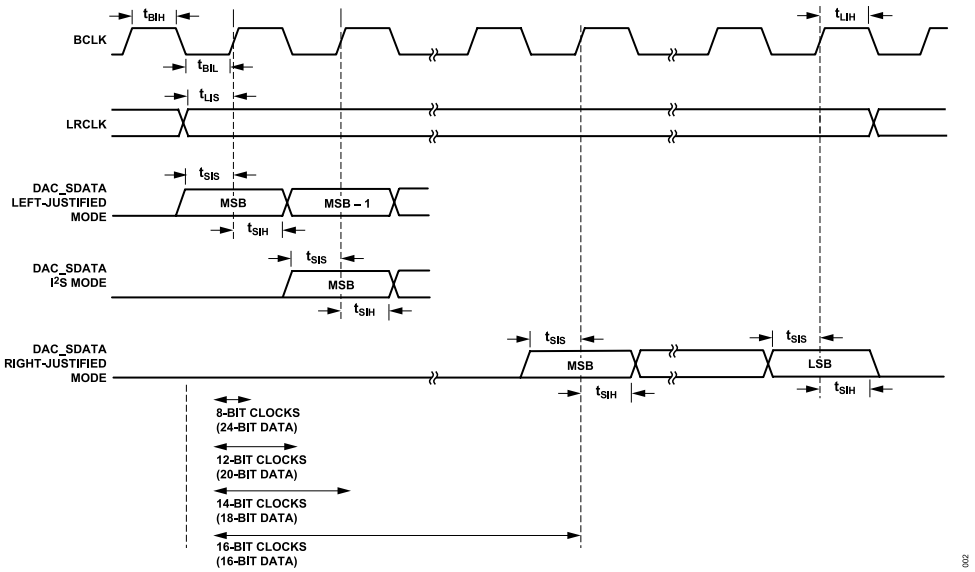


Figure 2. Serial Input Port Timing

SPECIFICATIONS

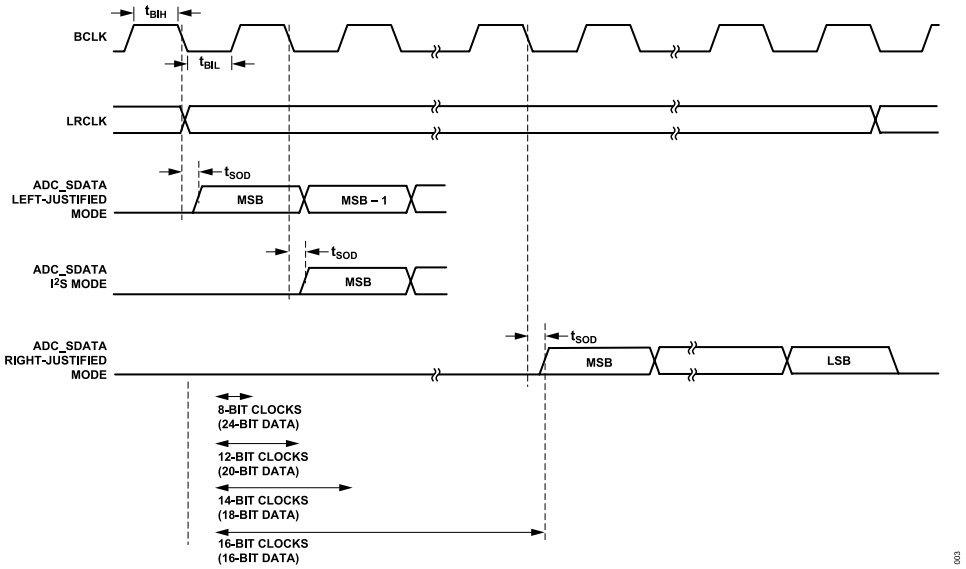


Figure 3. Serial Output Port Timing

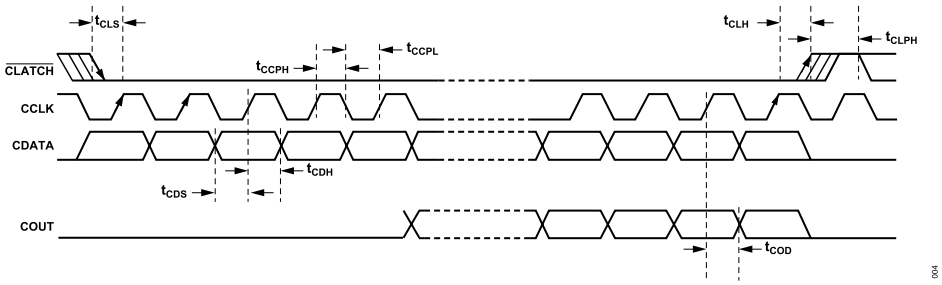


Figure 4. SPI Port Timing

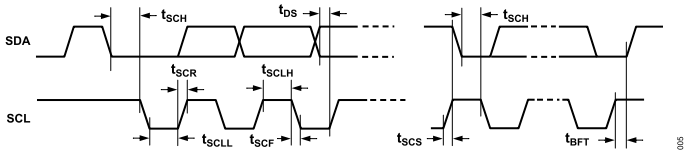


Figure 5. I²C Port Timing

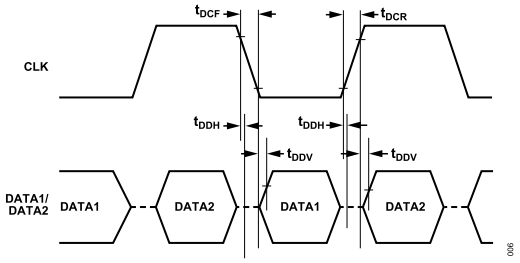


Figure 6. Digital Microphone Timing

ABSOLUTE MAXIMUM RATINGS

Table 8.

Parameter	Rating
Power Supply (AVDD)	−0.3 V to +3.65 V
Input Current (Except Supply Pins)	±20 mA
Analog Input Voltage (Signal Pins)	−0.3 V to AVDD + 0.3 V
Digital Input Voltage (Signal Pins)	−0.3 V to IOVDD + 0.3 V
Operating Temperature Range	−40°C to +85°C
Storage Temperature Range	−65°C to +150°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

θ_{JA} represents thermal resistance, junction-to-ambient; θ_{JC} represents thermal resistance, junction-to-case. All characteristics are for a 4-layer board.

Table 9. Thermal Resistance

Package Type	θ_{JA}	θ_{JC}	Unit
32-Lead LFCSP	50.1	17	°C/W

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

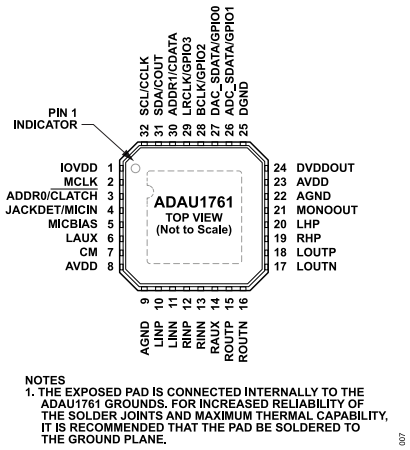


Figure 7. Pin Configuration

Table 10. Pin Function Descriptions

Pin No.	Mnemonic	Type[1]	Description
1	IOVDD	PWR	Supply for Digital Input and Output Pins. The digital output pins are supplied from IOVDD, which also sets the highest input voltage that should be seen on the digital input pins. IOVDD should be set between 1.8 V and 3.3 V. The current draw of this pin is variable because it is dependent on the loads of the digital outputs. IOVDD should be decoupled to DGND with a 100 nF capacitor and a 10 µF capacitor.
2	MCLK	D_IN	External Main Clock Input.
3	ADDR0/$\overline{\text{CLATCH}}$	D_IN	I^2C Address Bit 0 (ADDR0).SPI Latch Signal ($\overline{\text{CLATCH}}$). Must go low at the beginning of an SPI transaction and high at the end of a transaction. Each SPI transaction can take a different number of CCLKs to complete, depending on the address and read/write bit that are sent at the beginning of the SPI transaction.
4	JACKDET/MICIN	D_IN	Detect Insertion/Removal of Headphone Plug (JACKDET).Digital Microphone Stereo Input (MICIN).
5	MICBIAS	A_OUT	Bias Voltage for Electret Microphone.
6	LAUX	A_IN	Left Channel Single-Ended Auxiliary Input. Biased at AVDD/2.
7	CM	A_OUT	AVDD/2 V Common-Mode Reference. A 10 µF to 47 µF standard decoupling capacitor should be connected between this pin and AGND to reduce crosstalk between the ADCs and DACs. This pin can be used to bias external analog circuits, as long as they are not drawing current from CM (for example, the noninverting input of an op amp).
8	AVDD	PWR	1.8 V to 3.65 V Analog Supply for DAC and Microphone Bias. This pin should be decoupled locally to AGND with a 100 nF capacitor.
9	AGND	PWR	Analog Ground. The AGND and DGND pins can be tied together on a common ground plane. AGND should be decoupled locally to AVDD with a 100 nF capacitor.
10	LINP	A_IN	Left Channel Noninverting Input or Single-Ended Input 0. Biased at AVDD/2.
11	LINN	A_IN	Left Channel Inverting Input or Single-Ended Input 1. Biased at AVDD/2.
12	RINP	A_IN	Right Channel Noninverting Input or Single-Ended Input 2. Biased at AVDD/2.
13	RINN	A_IN	Right Channel Inverting Input or Single-Ended Input 3. Biased at AVDD/2.
14	RAUX	A_IN	Right Channel Single-Ended Auxiliary Input. Biased at AVDD/2.
15	ROUTP	A_OUT	Right Line Output, Positive. Biased at AVDD/2.
16	ROUTN	A_OUT	Right Line Output, Negative. Biased at AVDD/2.
17	LOUTN	A_OUT	Left Line Output, Negative. Biased at AVDD/2.
18	LOUTP	A_OUT	Left Line Output, Positive. Biased at AVDD/2.
19	RHP	A_OUT	Right Headphone Output. Biased at AVDD/2.
20	LHP	A_OUT	Left Headphone Output. Biased at AVDD/2.
21	MONOOUT	A_OUT	Mono Output or Virtual Ground for Capless Headphone. Biased at AVDD/2 when set as mono output.
22	AGND	PWR	Analog Ground. The AGND and DGND pins can be tied together on a common ground plane. AGND should be decoupled locally to AVDD with a 100 nF capacitor.
23	AVDD	PWR	1.8 V to 3.3 V Analog Supply for ADC, Output Driver, and Input to Digital Supply Regulator. This pin should be decoupled locally to AGND with a 100 nF capacitor.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

Table 10. Pin Function Descriptions (Continued)

Pin No.	Mnemonic	Type[1]	Description
24	DVDDOUT	PWR	Digital Core Supply Decoupling Point. The digital supply is generated from an on-board regulator and does not require an external supply. DVDDOUT should be decoupled to DGND with a 100 nF capacitor and a 10 µF capacitor.
25	DGND	PWR	Digital Ground. The AGND and DGND pins can be tied together on a common ground plane. DGND should be decoupled to DVDDOUT and to IOVDD with 100 nF capacitors and 10 µF capacitors.
26	ADC_SDATA/GPIO1	D_IO	ADC Serial Output Data (ADC_SDATA). General-Purpose Input/Output 1 (GPIO1).
27	DAC_SDATA/GPIO0	D_IO	DAC Serial Input Data (DAC_SDATA).General-Purpose Input/Output 0 (GPIO0).
28	BCLK/GPIO2	D_IO	Serial Data Port Bit Clock (BCLK).General-Purpose Input/Output 2 (GPIO2).
29	LRCLK/GPIO3	D_IO	Serial Data Port Frame Clock (LRCLK).General-Purpose Input/Output 3 (GPIO3).
30	ADDR1/CDATA	D_IN	I^2C Address Bit 1 (ADDR1). SPI Data Input (CDATA).
31	SDA/COUT	D_IO	I^2C Data (SDA). This pin is a bidirectional open-collector input/output. The line connected to this pin should have a 2 kΩ pull-up resistor.SPI Data Output (COUT). This pin is used for reading back registers and memory locations. It is three-state when an SPI read is not active.
32	SCL/CCLK	D_IN	I^2C Clock (SCL). This pin is always an open-collector input when in I^2C control mode. The line connected to this pin should have a 2 kΩ pull-up resistor.SPI Clock (CCLK). This pin can run continuously or be gated off between SPI transactions.
EP	Exposed Pad		Exposed Pad. The exposed pad is connected internally to the ADAU1761 grounds. For increased reliability of the solder joints and maximum thermal capability, it is recommended that the pad be soldered to the ground plane. See the Exposed Pad PCB Design section for more information.

[1] A_IN = analog input, A_OUT = analog output, D_IN = digital input, D_IO = digital input/output, PWR = power.

TYPICAL PERFORMANCE CHARACTERISTICS

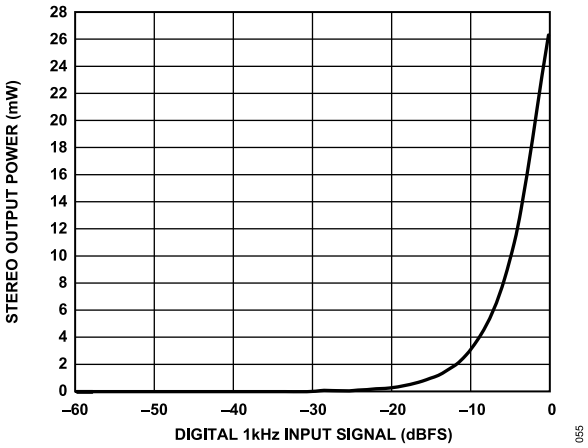


Figure 8. Headphone Amplifier Power vs. Input Level, 16 Ω Load

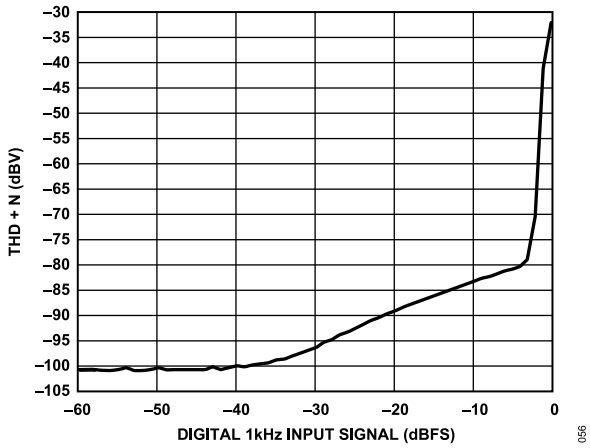


Figure 11. Headphone Amplifier THD + N vs. Input Level, 16 Ω Load

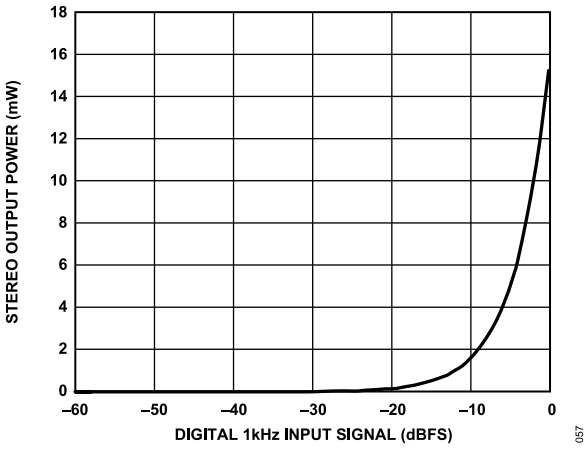


Figure 9. Headphone Amplifier Power vs. Input Level, 32 Ω Load

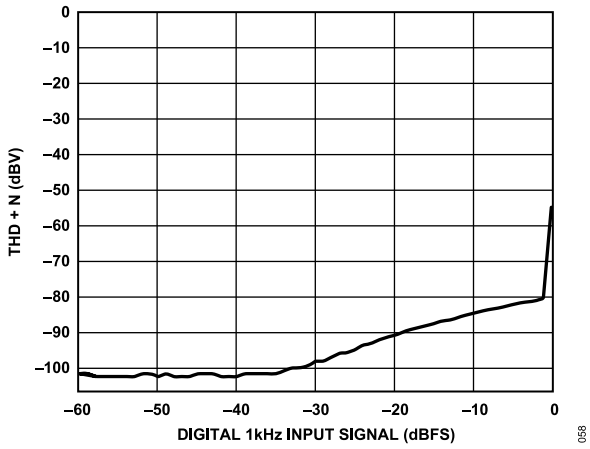


Figure 12. Headphone Amplifier THD + N vs. Input Level, 32 Ω Load

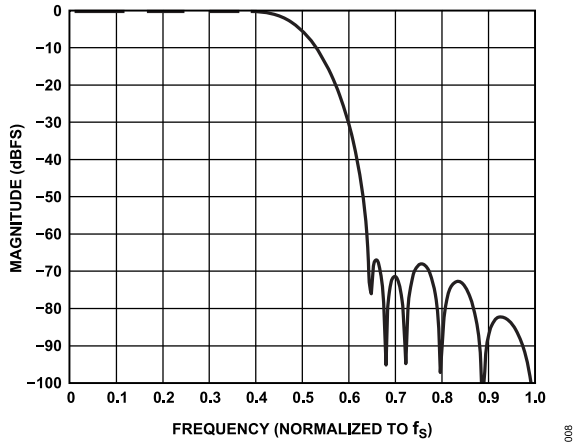


Figure 10. ADC Decimation Filter, 64× Oversampling, Normalized to f_S

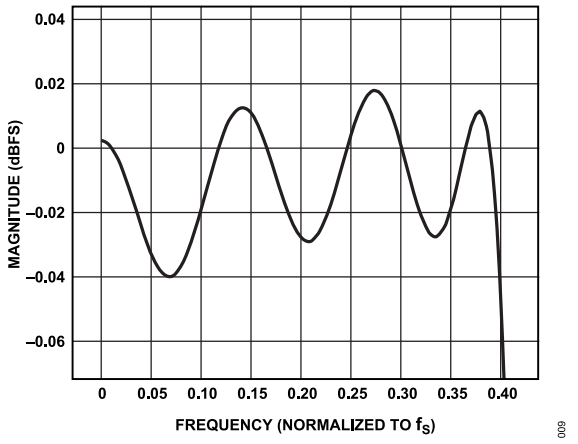


Figure 13. ADC Decimation Filter Pass-Band Ripple, 64× Oversampling, Normalized to f_S

TYPICAL PERFORMANCE CHARACTERISTICS

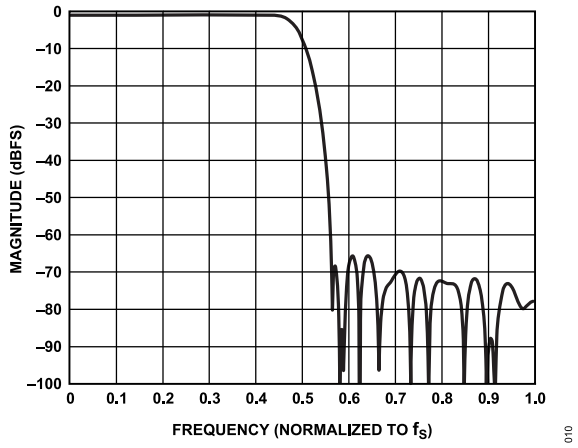


Figure 14. ADC Decimation Filter, 128× Oversampling, Normalized to f_S

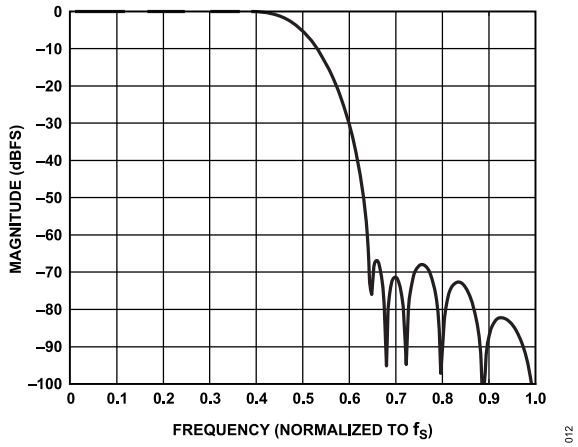


Figure 15. ADC Decimation Filter, 128× Oversampling, Double-Rate Mode, Normalized to f_S

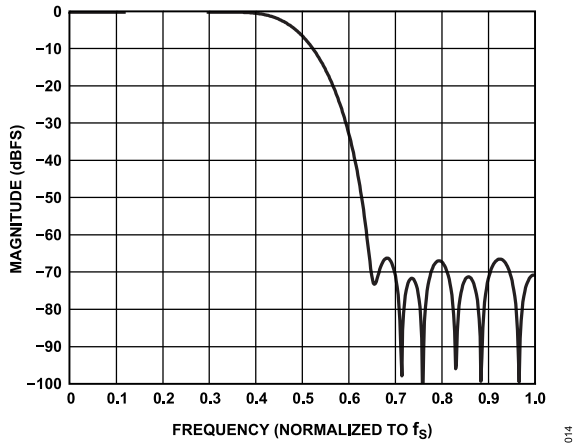


Figure 16. DAC Interpolation Filter, 64× Oversampling, Double-Rate Mode, Normalized to f_S

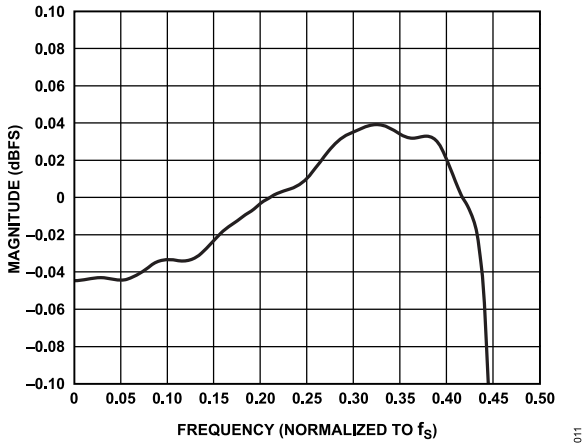


Figure 17. ADC Decimation Filter Pass-Band Ripple, 128× Oversampling, Normalized to f_S

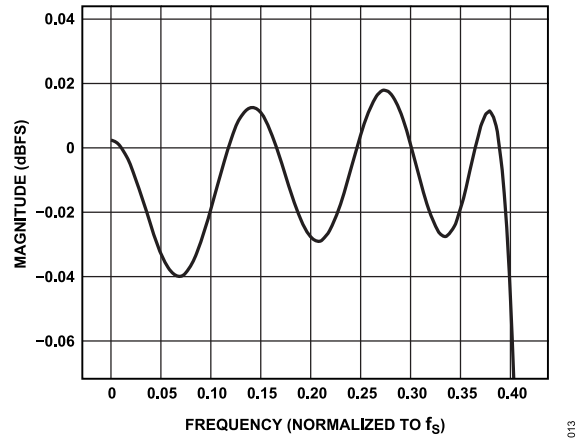


Figure 18. ADC Decimation Filter Pass-Band Ripple, 128× Oversampling, Double-Rate Mode, Normalized to f_S

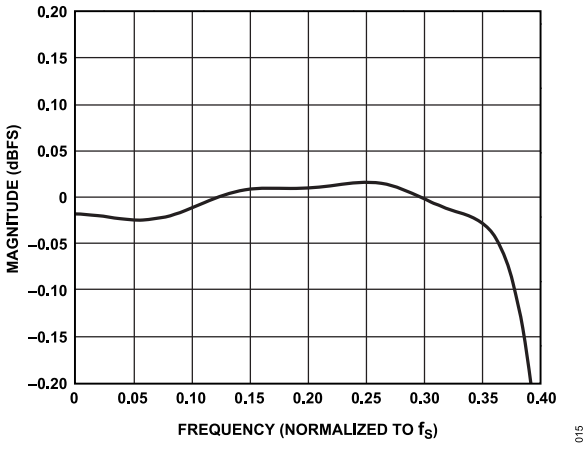


Figure 19. DAC Interpolation Filter Pass-Band Ripple, 64× Oversampling, Double-Rate Mode, Normalized to f_S

TYPICAL PERFORMANCE CHARACTERISTICS

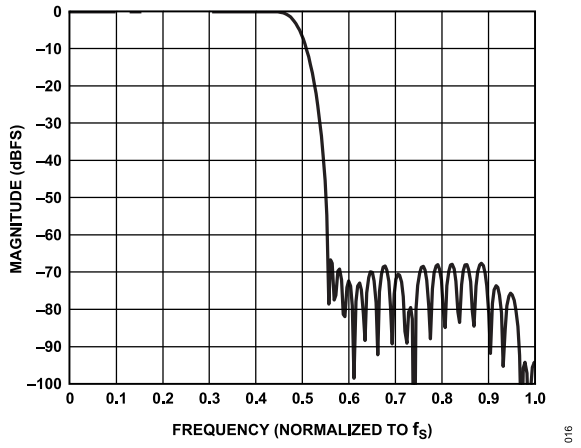


Figure 20. DAC Interpolation Filter, 128× Oversampling, Normalized to f_S

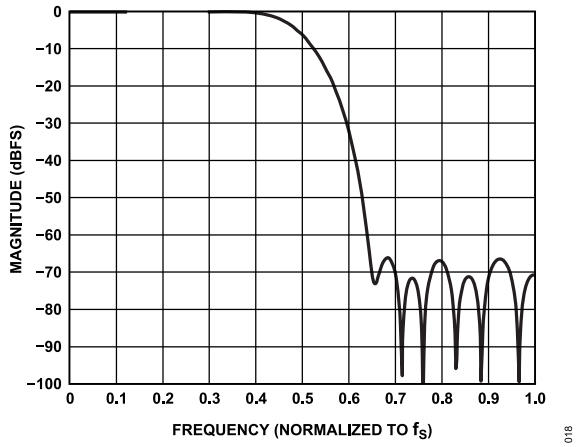


Figure 21. DAC Interpolation Filter, 128× Oversampling, Double-Rate Mode, Normalized to f_S

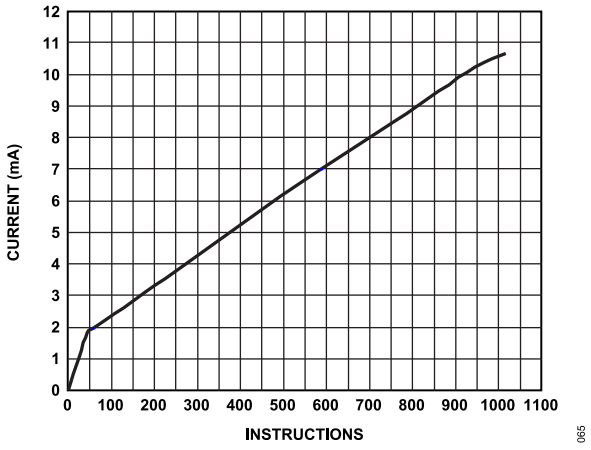


Figure 22. Typical DSP Current Draw

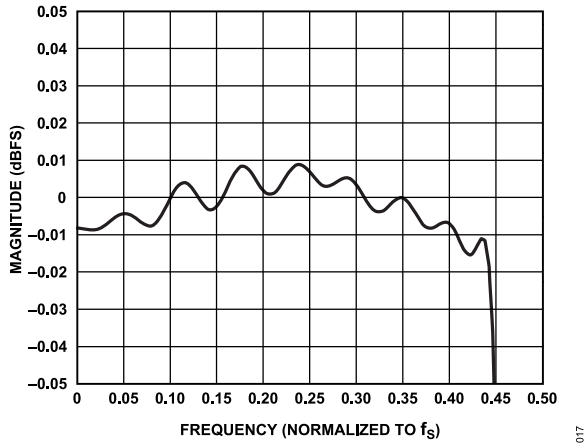


Figure 23. DAC Interpolation Filter Pass-Band Ripple, 128× Oversampling, Normalized to f_S

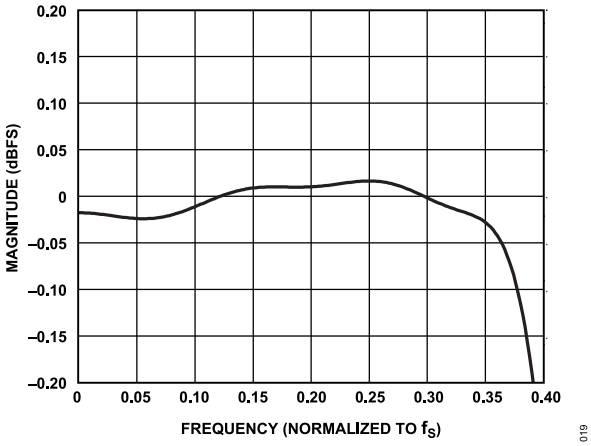


Figure 24. DAC Interpolation Filter Pass-Band Ripple, 128× Oversampling, Double-Rate Mode, Normalized to f_S

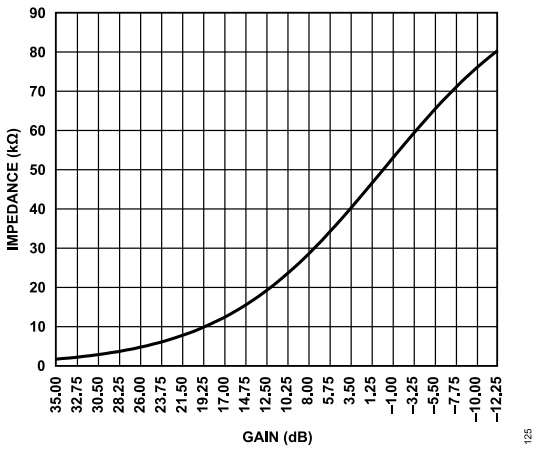


Figure 25. Input Impedance vs. Gain for Analog Inputs

SYSTEM BLOCK DIAGRAMS

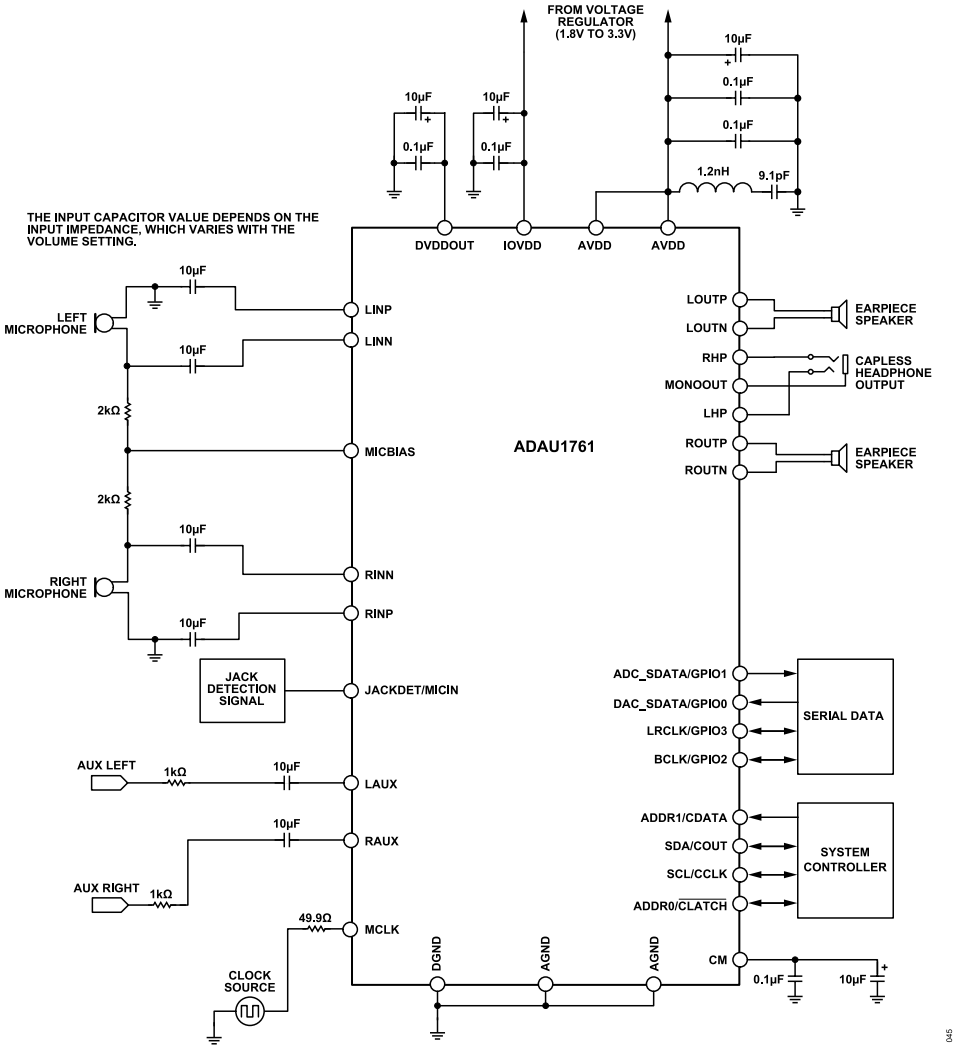


Figure 26. System Block Diagram

SYSTEM BLOCK DIAGRAMS

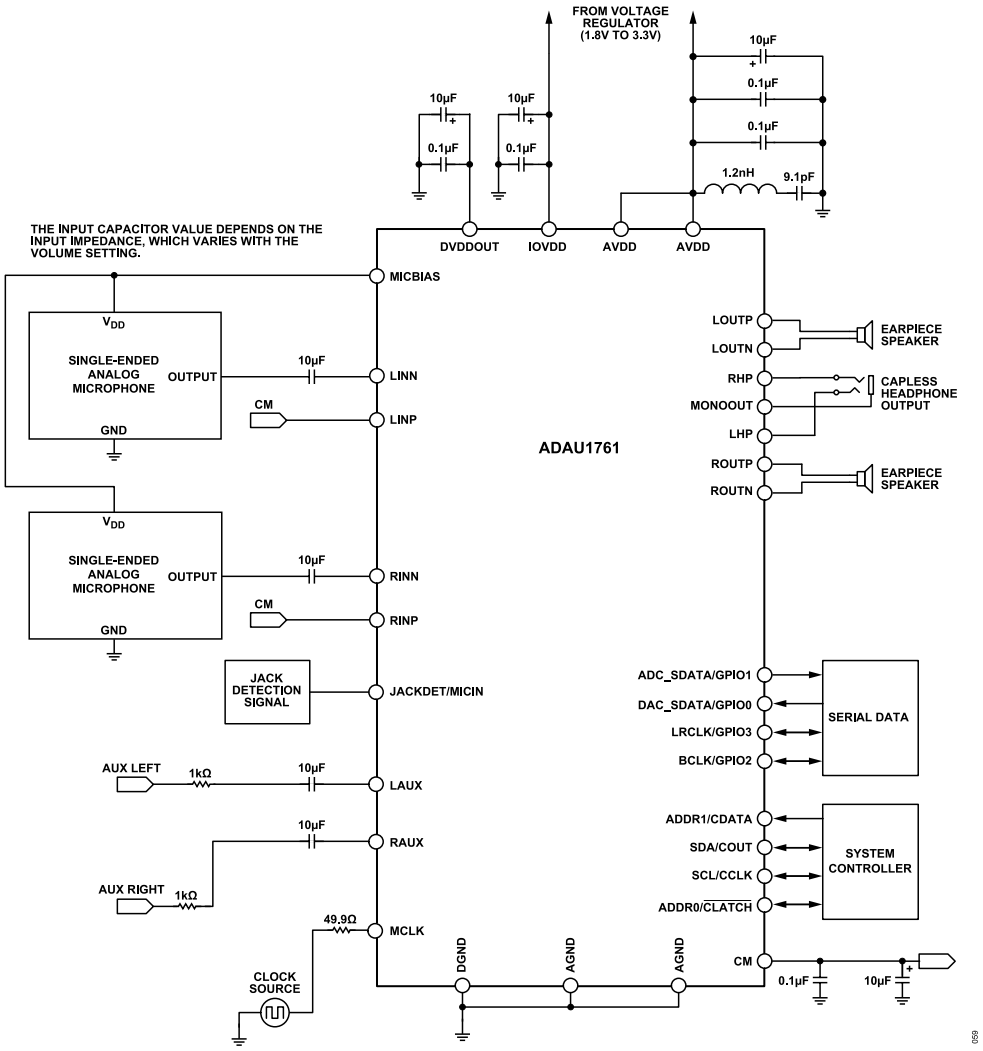


Figure 27. System Block Diagram with Analog Microphones

SYSTEM BLOCK DIAGRAMS

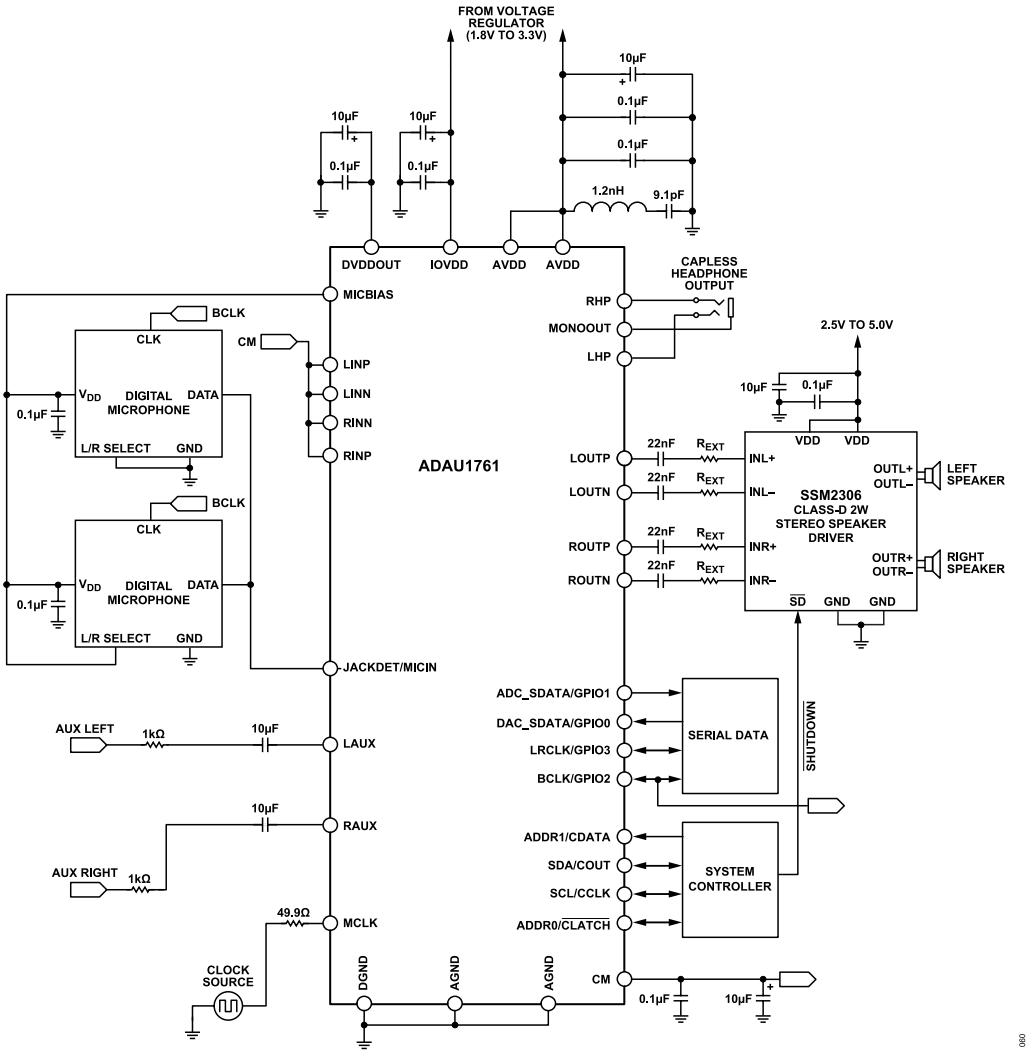


Figure 28. System Block Diagram with Digital Microphones and SSM2306 Class-D Speaker Driver

THEORY OF OPERATION

The ADAU1761 is a low power audio codec with an integrated stream-oriented DSP core, making it an all-in-one package that offers high quality audio, low power, small size, and many advanced features. The stereo ADC and stereo DAC each have an SNR of at least +98 dB and a THD + N of at least −90 dB. The serial data port is compatible with I²S, left-justified, right-justified, and TDM modes for interfacing to digital audio data. The operating voltage range is 1.8 V to 3.65 V, with an on-board regulator generating the internal digital supply voltage.

The record signal path includes very flexible input configurations that can accept differential and single-ended analog microphone inputs as well as a digital microphone input. A microphone bias pin provides seamless interfacing to electret microphones. Input configurations can accept up to six single-ended analog signals or variations of stereo differential or stereo single-ended signals with two additional auxiliary single-ended inputs. Each input signal has its own programmable gain amplifier (PGA) for volume adjustment and can be routed directly to the playback path output mixers, bypassing the ADCs. An automatic level control (ALC) can also be implemented to keep the recording volume constant.

The ADCs and DACs are high quality, 24-bit Σ-Δ converters that operate at selectable 64× or 128× oversampling ratios. The base sampling rate of the converters is set by the input clock rate and can be further scaled with the converter control register settings. The converters can operate at sampling frequencies from 8 kHz to 96 kHz. The ADCs and DACs also include very fine-step digital volume controls.

The playback path allows input signals and DAC outputs to be mixed into various output configurations. Headphone drivers are available for a stereo headphone output, and the other output pins are capable of differentially driving an earpiece speaker. Capless headphone outputs are possible with the use of the mono output as a virtual ground connection. The stereo line outputs can be used as either single-ended or differential outputs and as an optional mix-down mono output.

The DSP core introduces many features that make this codec unique and optimized for audio processing. The program and parameter RAMs can be loaded with custom audio processing signal flow built using the SigmaStudio graphical programming software from Analog Devices, Inc. The values stored in the parameter RAM control individual signal processing blocks, such as equalization filters, dynamics processors, audio delays, and mixer levels.

The SigmaStudio software is used to program and control the SigmaDSP through the control port. Along with designing and tuning a signal flow, the tools can be used to configure all of the DSP registers. The SigmaStudio graphical interface allows anyone with digital or analog audio processing knowledge to easily design DSP signal flow and port it to a target application. At the same time, it provides enough flexibility and programmability for an experienced DSP programmer to have in-depth control of the design. In SigmaStudio, the user can connect graphical blocks (such as biquad filters, dynamics processors, mixers, and delays), compile the design, and load the program and parameter files into the ADAU1761 memory through the control port. Signal processing blocks available in the provided libraries include the following:

- Enhanced stereo capture
- Single- and double-precision biquad filters
- FIR filters
- Dynamics processors with peak or rms detection for mono and multichannel dynamics
- Mixers and splitters
- Tone and noise generators
- Fixed and variable gain
- Loudness
- Delay
- Stereo enhancement
- Dynamic bass boost
- Noise and tone sources
- Level detectors

Additional processing blocks are always being developed. Analog Devices also provides proprietary and third-party algorithms for applications such as matrix decoding, bass enhancement, and surround virtualizers. Contact Analog Devices (www.analog.com) for information about licensing these algorithms.

The ADAU1761 can generate its internal clocks from a wide range of input clocks by using the on-board fractional PLL. The PLL accepts inputs from 8 MHz to 27 MHz.

The ADAU1761 is provided in a small, 32-lead, 5 mm × 5 mm LFCSP with an exposed bottom pad.

STARTUP, INITIALIZATION, AND POWER

This section describes the procedure for properly starting up the ADAU1761. The following sequence provides a high level approach to the proper initiation of the system.

1. Apply power to the ADAU1761.
2. Lock the PLL to the input clock (if using the PLL).
3. Enable the core clock.
4. Load the register settings.

See the Startup section for more information about the proper start-up sequence.

POWER-UP SEQUENCE

The ADAU1761 uses a power-on reset (POR) circuit to reset the registers upon power-up. The POR monitors the DVDDOUT pin and generates a reset signal whenever power is applied to the chip. During the reset, the ADAU1761 is set to the default values documented in the register map (see the Control Registers section). Typically, with a 10 µF capacitor on AVDD, the POR takes approximately 14 ms.

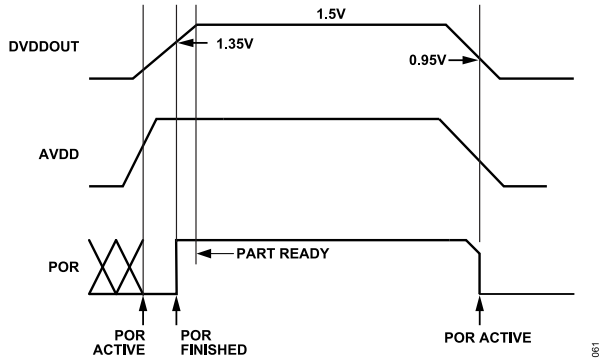


Figure 29. Power-On Reset Sequence

The PLL lock time is dependent on the MCLK rate. Typical lock times are provided in Table 11. The DSP can be enabled immediately after the PLL is locked.

Table 11. PLL Lock Times

PLL Mode	MCLK Frequency	Lock Time (Typical)
Fractional	8 MHz	3.5 ms
Fractional	12 MHz	3.0 ms
Integer	12.288 MHz	2.96 ms
Fractional	13 MHz	2.4 ms
Fractional	14.4 MHz	2.4 ms
Fractional	19.2 MHz	2.98 ms
Fractional	19.68 MHz	2.98 ms
Fractional	19.8 MHz	2.98 ms
Fractional	24 MHz	2.95 ms
Integer	24.576 MHz	2.96 ms
Fractional	26 MHz	2.4 ms
Fractional	27 MHz	2.4 ms

POWER REDUCTION MODES

Sections of the ADAU1761 chip can be turned on and off as needed to reduce power consumption. These include the ADCs, the DACs, the PLL, and the DSP core.

In addition, the control registers can be used to configure some functions for power saving, normal, or enhanced performance operation. See the Control Registers section for more information.

The digital filters of the ADCs and DACs can each be set to oversampling ratios of 64× or 128× (default). Setting the oversampling ratios to 64× for these filters lowers power consumption with a minimal impact on performance. See the Digital Filters section for specifications; see the Typical Performance Characteristics section for graphs of these filters.

DIGITAL POWER SUPPLY

The digital power supply for the ADAU1761 is generated from an internal regulator. This regulator generates a 1.5 V supply internally. The only external connection to this regulator is the DVDDOUT bypassing point. A 100 nF capacitor and a 10 µF capacitor should be connected between this pin and DGND.

INPUT/OUTPUT POWER SUPPLY

The power for the digital output pins is supplied from IOVDD, and this pin also sets the highest input voltage that should be seen on the digital input pins. IOVDD should be set between 1.8 V and 3.3 V; no digital input signal should be at a voltage level higher than the one on IOVDD. The current draw of this pin is variable because it depends on the loads of the digital outputs. IOVDD should be decoupled to DGND with a 100 nF capacitor and a 10 µF capacitor.

CLOCK GENERATION AND MANAGEMENT

The ADAU1761 uses a flexible clocking scheme that enables the use of many different input clock rates. The PLL can be bypassed or used, resulting in two different approaches to clock management. For more information about clocking schemes, PLL configuration, and sampling rates, see the Clocking and Sampling Rates section.

Case 1: PLL Is Bypassed

If the PLL is bypassed, the core clock is derived directly from the MCLK input. The rate of this clock must be set properly in Register R0 (clock control register, Address 0x4000) using the INFREQ[1:0] bits. When the PLL is bypassed, supported external clock rates are 256 × f_S, 512 × f_S, 768 × f_S, and 1024 × f_S, where f_S is the base sampling rate. The core clock of the chip is off until the core clock enable bit (COREN) is asserted. If a clock slower than 1024 × f_S is directly input to the ADAU1761 (bypassing the PLL), the number of available SigmaDSP processing cycles is reduced and the DSPSR bits in Register R57 (Address 0x40EB) should be adjusted accordingly.

STARTUP, INITIALIZATION, AND POWER

Case 2: PLL Is Used

The core clock to the entire chip is off during the PLL lock acquisition period. The user can poll the lock bit to determine when the PLL has locked. After lock is acquired, the ADAU1761 can be started by asserting the core clock enable bit (COREN) in Register R0 (clock control register, Address 0x4000). This bit enables the core clock to all the internal blocks of the ADAU1761.

PLL Lock Acquisition

During the lock acquisition period, only Register R0 (Address 0x4000) and Register R1 (Address 0x4002) are accessible through the control port. Because all other registers require a valid main clock for reading and writing, do not attempt to access any other register. Any read or write is prohibited until the core clock enable bit (COREN) and the lock bit are both asserted.

To program the PLL during initialization or reconfiguration of the clock setting, the following procedure must be followed:

1. Power down the PLL.
2. Reset the PLL control register.
3. Start the PLL.
4. Poll the lock bit.
5. Assert the core clock enable bit after the PLL lock is acquired.

The PLL control register (Register R1, Address 0x4002) is a 48-bit register where all bits must be written with a single continuous write to the control port.

CLOCKING AND SAMPLING RATES

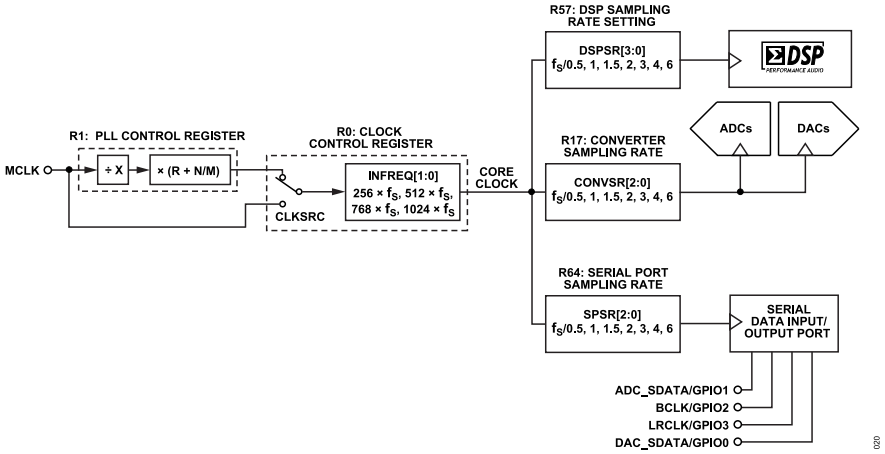


Figure 30. Clock Tree Diagram

CORE CLOCK

Clocks for the converters, the serial ports, and the DSP are derived from the core clock. The core clock can be derived directly from MCLK or it can be generated by the PLL. The CLKSRC bit (Bit 3 in Register R0, Address 0x4000) determines the clock source.

The INFREQ[1:0] bits should be set according to the expected input clock rate selected by CLKSRC; this value also determines the core clock rate and the base sampling frequency, f_S.

For example, if the input to CLKSRC = 49.152 MHz (from PLL), then

INFREQ[1:0] = 1024 × f_S

f_S = 49.152 MHz/1024 = 48 kHz

The PLL output clock rate is always 1024 × f_S, and the clock control register automatically sets the INFREQ[1:0] bits to 1024 × f_S when using the PLL. When using a direct clock, the INFREQ[1:0] frequency should be set according to the MCLK pin clock rate and the desired base sampling frequency.

To utilize the maximum amount of DSP instructions, the core clock should run at a rate of 1024 × f_S.

Table 12. Clock Control Register (Register R0, Address 0x4000)

Bits	Bit Name	Settings
3	CLKSRC	0: Direct from MCLK pin (default) 1: PLL clock
[2:1]	INFREQ[1:0]	00: 256 × f_S (default)01: 512 × f_S10: 768 × f_S11: 1024 × f_S
0	COREN	0: Core clock disabled (default)1: Core clock enabled

SAMPLING RATES

The ADCs, DACs, and serial port share a common sampling rate that is set in Register R17 (Converter Control 0 register, Address 0x4017). The CONVSR[2:0] bits set the sampling rate as a ratio of the base sampling frequency. The DSP sampling rate is set in Register R57 (DSP sampling rate setting register, Address 0x40EB) using the DSPSR[3:0] bits, and the serial port sampling rate is set in Register R64 (serial port sampling rate register, Address 0x40F8) using the SPSR[2:0] bits.

It is recommended that the sampling rates for the converters, serial ports, and DSP be set to the same value, unless appropriate compensation filtering is done within the DSP. Table 13 and Table 14 list the sampling rate divisions for common base sampling rates.

Table 13. 48 kHz Base Sampling Rate Divisions

Base Sampling Frequency	Sampling Rate Scaling	Sampling Rate
f_S = 48 kHz	f_S/1	48 kHz
	f_S/6	8 kHz
	f_S/4	12 kHz
	f_S/3	16 kHz
	f_S/2	24 kHz
	f_S/1.5	32 kHz
	f_S/0.5	96 kHz

Table 14. 44.1 kHz Base Sampling Rate Divisions

Base Sampling Frequency	Sampling Rate Scaling	Sampling Rate
f_S = 44.1 kHz	f_S/1	44.1 kHz
	f_S/6	7.35 kHz
	f_S/4	11.025 kHz
	f_S/3	14.7 kHz
	f_S/2	22.05 kHz
	f_S/1.5	29.4 kHz
	f_S/0.5	88.2 kHz

PLL

The PLL uses the MCLK as a reference to generate the core clock. PLL settings are set in Register R1 (PLL control register, Address 0x4002). Depending on the MCLK frequency, the PLL must be set for either integer or fractional mode. The PLL can accept input frequencies in the range of 8 MHz to 27 MHz.

CLOCKING AND SAMPLING RATES

All six bytes in the PLL control register must be written with a single continuous write to the control port.

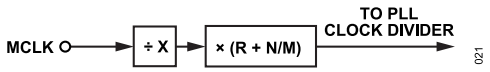


Figure 31. PLL Block Diagram

Integer Mode

Integer mode is used when the MCLK is an integer (R) multiple of the PLL output (1024 × f_S).

For example, if MCLK = 12.288 MHz and f_S = 48 kHz, then

PLL required output = 1024 × 48 kHz = 49.152 MHz

R = 49.152 MHz/12.288 MHz = 4

In integer mode, the values set for N and M are ignored.

Fractional Mode

Fractional mode is used when the MCLK is a fractional (R + (N/M)) multiple of the PLL output.

For example, if MCLK = 12 MHz and f_S = 48 kHz, then

PLL required output = 1024 × 48 kHz = 49.152 MHz

R + (*N*/*M*) = 49.152 MHz/12 MHz = 4 + (12/125)

Common fractional PLL parameter settings for 44.1 kHz and 48 kHz sampling rates can be found in Table 16 and Table 17.

The PLL outputs a clock in the range of 41 MHz to 54 MHz, which should be taken into account when calculating PLL values and MCLK frequencies.

Table 15. PLL Control Register (Register R1, Address 0x4002)

Bits	Bit Name	Description
[47:32]	M[15:0]	Denominator of the fractional PLL: 16-bit binary number0x00FD: M = 253 (default)
[31:16]	N[15:0]	Numerator of the fractional PLL: 16-bit binary number0x000C: N = 12 (default)
[14:11]	R[3:0]	Integer part of PLL: four bits, only values 2 to 8 are valid0010: R = 2 (default)0011: R = 30100: R = 40101: R = 50110: R = 60111: R = 71000: R = 8
[10:9]	X[1:0]	PLL input clock divider00: X = 1 (default)01: X = 210: X = 311: X = 4
8	Type	PLL operation mode0: Integer (default)1: Fractional
1	Lock	PLL lock (read-only bit)0: PLL unlocked (default)1: PLL locked
0	PLLEN	PLL enable0: PLL disabled (default)1: PLL enabled

Table 16. Fractional PLL Parameter Settings for f_S = 44.1 kHz (PLL Output = 45.1584 MHz = 1024 × f_S)

MCLK Input (MHz)	Input Divider (X)	Integer (R)	Denominator (M)	Numerator (N)	R2: PLL Control Setting (Hex)
8	1	5	625	403	0x0271 0193 2901
12	1	3	625	477	0x0271 01DD 1901
13	1	3	8125	3849	0x1FBD 0F09 1901
14.4	2	6	125	34	0x007D 0022 3301
19.2	2	4	125	88	0x007D 0058 2301
19.68	2	4	1025	604	0x0401 025C 2301
19.8	2	4	1375	772	0x055F 0304 2301
24	2	3	625	477	0x0271 01DD 1B01
26	2	3	8125	3849	0x1FBD 0F09 1B01
27	2	3	1875	647	0x0753 0287 1B01

Table 17. Fractional PLL Parameter Settings for f_S = 48 kHz (PLL Output = 49.152 MHz = 1024 × f_S)

MCLK Input (MHz)	Input Divider (X)	Integer (R)	Denominator (M)	Numerator (N)	R2: PLL Control Setting (Hex)
8	1	6	125	18	0x007D 0012 3101
12	1	4	125	12	0x007D 000C 2101
13	1	3	1625	1269	0x0659 04F5 1901
14.4	2	6	75	62	0x004B 003E 3301
19.2	2	5	25	3	0x0019 0003 2B01
19.68	2	4	205	204	0x00CD 00CC 2301
19.8	2	4	825	796	0x0339 031C 2301
24	2	4	125	12	0x007D 000C 2301

CLOCKING AND SAMPLING RATES

Table 17. Fractional PLL Parameter Settings for f_S = 48 kHz (PLL Output = 49.152 MHz = 1024 × f_S) (Continued)

MCLK Input (MHz)	Input Divider (X)	Integer (R)	Denominator (M)	Numerator (N)	R2: PLL Control Setting (Hex)
26	2	3	1625	1269	0x0659 04F5 1B01
27	2	3	1125	721	0x0465 02D1 1B01

Table 18. Integer PLL Parameter Settings for f_S = 48 kHz (PLL Output = 49.152 MHz = 1024 × f_S)

MCLK Input (MHz)	Input Divider (X)	Integer (R)	Denominator (M)	Numerator (N)	R2: PLL Control Setting (Hex)[1]
12.288	1	4	Don't care	Don't care	0xXXXX XXXX 2001
24.576	1	2	Don't care	Don't care	0xXXXX XXXX 1001

[1] X = don't care.

RECORD SIGNAL PATH

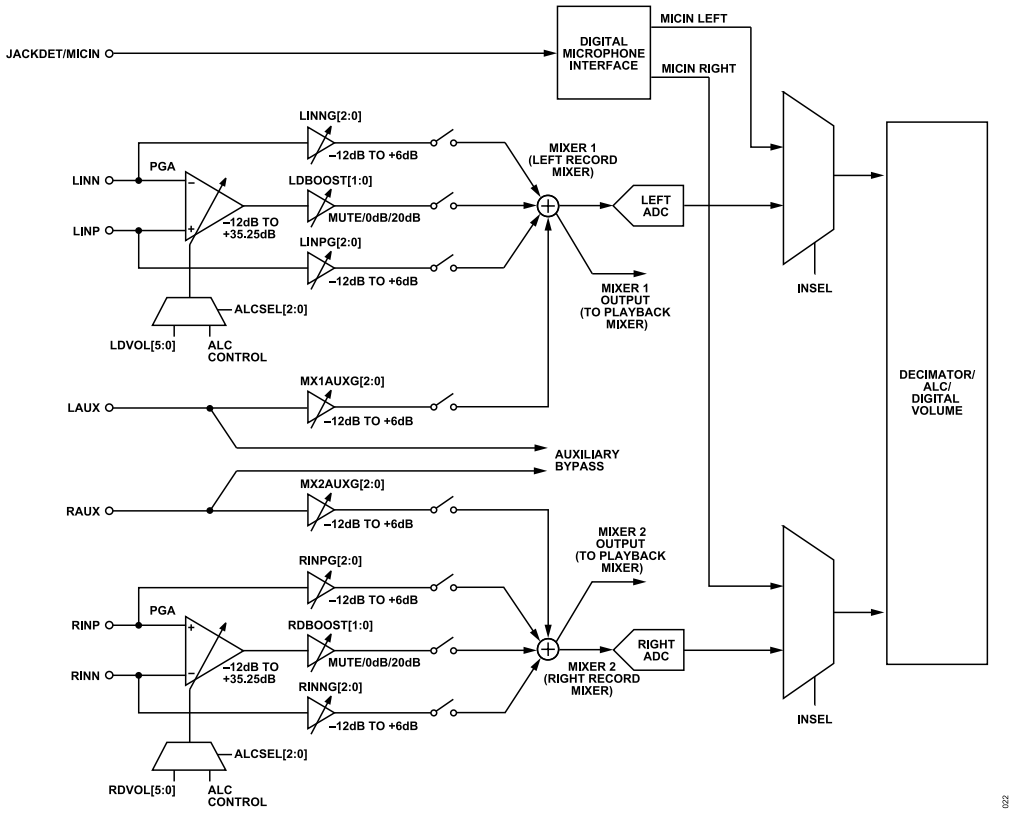


Figure 32. Record Signal Path

INPUT SIGNAL PATHS

The ADAU1761 can accept both line level and microphone inputs. The analog inputs can be configured in a single-ended or differential configuration. There is also an input for a digital microphone. The analog inputs are biased at AVDD/2. Unused input pins should be connected to CM.

Each of the six analog inputs has individual gain controls (boost or cut). The input signals are mixed and routed to an ADC. The mixed input signals can also bypass the ADCs and be routed directly to the playback mixers. Left channel inputs are mixed before the left ADC; however, it is possible to route the mixed analog signal around the ADC and output it into a left or right output channel. The same capabilities apply to the right channel and the right ADC.

Signals are inverted through the PGAs and the mixers. The result of this inversion is that differential signals input through the PGA are output from the ADCs at the same polarity as they are input. Single-ended inputs that pass through the mixer but not through the PGA are inverted. The ADCs are noninverting.

The input impedance of the analog inputs varies with the gain of the PGA. This impedance ranges from 1.7 kΩ at the 35.25 dB gain setting to 80.4 kΩ at the −12 dB setting. This range is shown in Figure 25.

Analog Microphone Inputs

For microphone inputs, configure the part in either stereo pseudo-differential mode or stereo full differential mode.

The LINN and LINP pins are the inverting and noninverting inputs for the left channel, respectively. The RINN and RINP pins are the inverting and noninverting inputs for the right channel, respectively.

For a differential microphone input, connect the positive signal to the noninverting input of the PGA and the negative signal to the inverting input of the PGA, as shown in Figure 33. The PGA settings are controlled with Register R8 (left differential input volume control register, Address 0x400E) and Register R9 (right differential input volume control register, Address 0x400F). The PGA must first be enabled by setting the RDEN and LDEN bits.

RECORD SIGNAL PATH

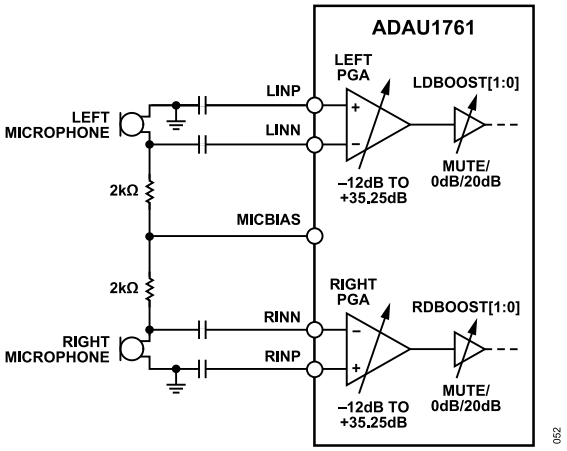


Figure 33. Stereo Differential Microphone Configuration

The PGA can also be used for single-ended microphone inputs. Connect LINP and/or RINP to the CM pin. In this configuration, the signal connects to the inverting input of the PGA, LINN and/or RINN, as shown in Figure 34.

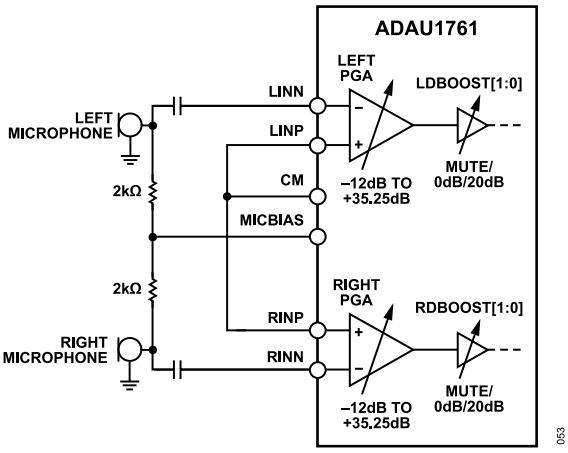


Figure 34. Stereo Single-Ended Microphone Configuration

Analog Line Inputs

Line input signals can be accepted by any analog input. It is possible to route signals on the RINN, RINP, LINN, and LINP pins around the differential amplifier to their own amplifier and to use these pins as single-ended line inputs by disabling the LDEN and RDEN bits (Bit 0 in Register R8, Address 0x400E, and Bit 0 in Register R9, Address 0x400F). Figure 35 depicts a stereo single-ended line input using the RINN and LINN pins.

The LAUX and RAUX pins are single-ended line inputs. They can be used together as a stereo single-ended auxiliary input, as shown in Figure 35. These inputs can bypass the input gain control, mixers, and ADCs to directly connect to the output playback mixers (see auxiliary bypass in Figure 32).

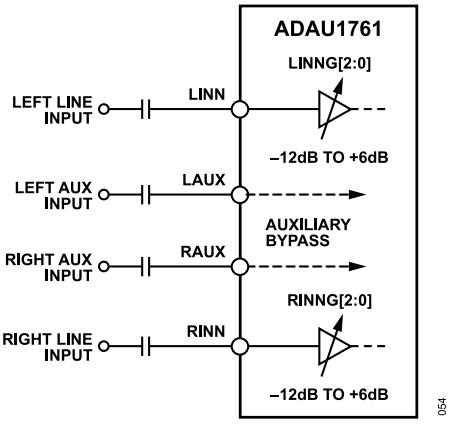


Figure 35. Stereo Single-Ended Line Input with Stereo Auxiliary Bypass

Digital Microphone Input

When using a digital microphone connected to the JACKDET/MICIN pin, the JDFUNC[1:0] bits in Register R2 (Address 0x4008) must be set to 10 to enable the microphone input and disable the jack detection function. The ADAU1761 must operate in main mode and source BCLK to the input clock of the digital microphone. The DSPRUN bit must also be asserted in Register R62 (DSP run register, Address 0x40F6) for digital microphone operation.

The digital microphone signal bypasses record path mixers and ADCs and is routed directly into the decimation filters. The digital microphone and ADCs share decimation filters and, therefore, both cannot be used simultaneously. The digital microphone input select bit, INSEL, can be set in Register R19 (ADC control register, Address 0x4019). Figure 36 depicts the digital microphone interface and signal routing.

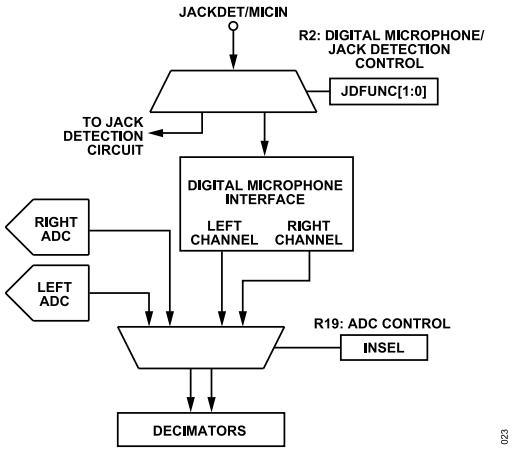


Figure 36. Digital Microphone Interface Block Diagram

Microphone Bias

The MICBIAS pin provides a voltage reference for electret analog microphones. The MICBIAS voltage is set in Register R10 (record microphone bias control register, Address 0x4010). In this register, the MICBIAS output can be enabled or disabled. Additional options

include high performance operation and a gain boost. The gain boost provides two different voltage biases: 0.65 × AVDD or 0.90 × AVDD. When enabled, the high performance bit increases supply current to the microphone bias circuit to decrease rms input noise.

The MICBIAS pin can also be used to cleanly supply voltage to digital microphones or analog microphones with separate power supply pins.

ANALOG-TO-DIGITAL CONVERTERS

The ADAU1761 uses two 24-bit Σ-Δ analog-to-digital converters (ADCs) with selectable oversampling ratios of 64× or 128× (selected by Bit 3 in Register R17, Address 0x4017).

ADC Full-Scale Level

The full-scale input to the ADCs (0 dBFS) depends on AVDD. At AVDD = 3.3 V, the full-scale input level is 1.0 V rms. This full-scale analog input will output a digital signal at −1.38 dBFS. This gain offset is built into the ADAU1761 to prevent clipping. The full-scale input level scales linearly with the level of AVDD.

For single-ended and pseudo-differential signals, the full-scale value corresponds to the signal level at the pins, 0 dBFS.

The full differential full-scale input level is measured after the differential amplifier, which corresponds to −6 dBFS at each pin.

Signal levels above the full-scale value cause the ADCs to clip.

Digital ADC Volume Control

The digital ADC volume can be attenuated before DSP processing using Register R20 (left input digital volume register, Address 0x401A) and Register R21 (right input digital volume register, Address 0x401B).

High-Pass Filter

By default, a high-pass filter is used in the ADC path to remove dc offsets; this filter can be enabled or disabled in Register R19 (ADC control register, Address 0x4019). At f_S = 48 kHz, the corner frequency of this high-pass filter is 2 Hz.

AUTOMATIC LEVEL CONTROL (ALC)

The ADAU1761 contains a hardware automatic level control (ALC). The ALC is designed to continuously adjust the PGA gain to keep the recording volume constant as the input level varies.

For optimal noise performance, the ALC uses the analog PGA to adjust the gain instead of using a digital method. This ensures that the ADC noise is not amplified at low signal levels. Extremely small gain step sizes are used to ensure high audio quality during gain changes.

To use the ALC function, the inputs must be applied either differentially or pseudo-differentially to input pins LINN and LINP, for the left channel, and RINN and RINP, for the right channel. The ALC function is not available for the auxiliary line input pins, LAUX and RAUX.

A block diagram of the ALC block is shown in Figure 37. The ALC logic receives the ADC output signals and analyzes these digital signals to set the PGA gain. The ALC control registers are used to control the time constants and output levels, as described in this section.

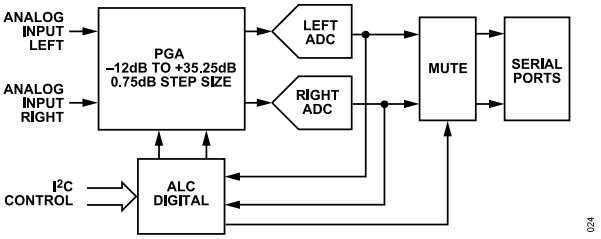


Figure 37. ALC Architecture

ALC PARAMETERS

The ALC function is controlled with the ALC control registers (Address 0x4011 through Address 0x4014) using the following parameters:

- ALCSEL[2:0]: The ALC select bits are used to enable the ALC and set the mode to left only, right only, stereo, or DSP. In stereo mode, the greater of the left or right inputs is used to calculate the gain, and the same gain is then applied to both the left and right channels. In DSP mode, the PGA gain is controlled by the SigmaDSP core.
- ALCTARG[3:0]: The ALC target is the desired input recording level that the ALC attempts to achieve.
- ALCATCK[3:0]: The ALC attack time sets how fast the ALC starts attenuating after a sudden increase in input level above the ALC target. Although it may seem that the attack time should be set as fast as possible to avoid clipping on transients, using a moderate value results in better overall sound quality. If the value is too fast, the ALC overreacts to very short transients, causing audible gain-pumping effects, which sounds worse than using a moderate value that allows brief periods of clipping on transients. A typical setting for music recording is 384 ms. A typical setting for voice recording is 24 ms.
- ALCHOLD[3:0]: These bits set the ALC hold time. When the output signal falls below the target output level, the gain is not increased unless the output remains below the target level for the period of time set by the hold time bits. The hold time is used to prevent the gain from modulating on a steady low frequency sine wave signal, which would cause distortion.
- ALCDEC[3:0]: The ALC decay time sets how fast the ALC increases the PGA gain after a sudden decrease in input level below the ALC target. A very slow setting can be used if the main function of the ALC is to set a music recording level. A faster setting can be used if the function of the ALC is to compress the dynamic range of a voice recording. Using a very fast decay time can cause audible artifacts such as noise pumping or distortion. A typical setting for music recording is 24.58 sec. A typical setting for voice recording is 1.54 sec.
- ALCMAX[2:0]: The maximum ALC gain bits are used to limit the maximum gain that can be programmed into the ALC. This can be used to prevent excessive noise in the recording for small input signals. Note that setting this register to a low value may prevent the ALC from reaching its target output level, but this behavior is often desirable to achieve the best overall sound.

Figure 38 shows the dynamic behavior of the PGA gain for a tone-burst input. The target output is achieved for three different input levels, with the effect of attack, hold, and decay shown in the figure. Note that for very small signals, the maximum PGA gain may prevent the ALC from achieving its target level; in the same way, for very large inputs, the minimum PGA gain may prevent the ALC from achieving its target level (assuming that the target output level is set to a very low value). The effects of the PGA gain limit are shown in the input/output graph of Figure 39.

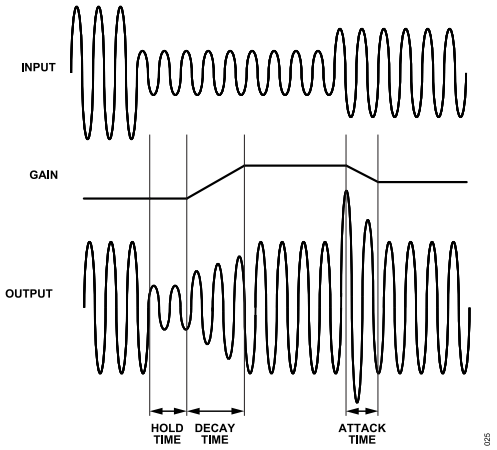


Figure 38. Basic ALC Operation

AUTOMATIC LEVEL CONTROL (ALC)

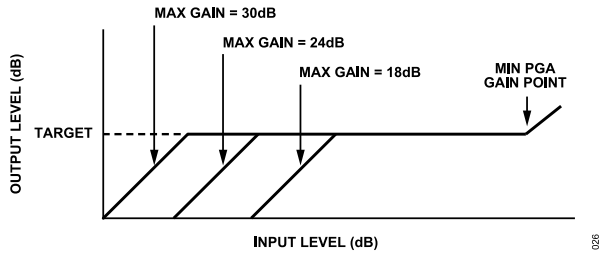


Figure 39. Effect of Varying the Maximum Gain Parameter

NOISE GATE FUNCTION

When using the ALC, one potential problem is that for small input signals, the PGA gain can become very large. A side effect of this is that the noise is amplified along with the signal of interest. To avoid this situation, the ADAU1761 noise gate can be used. The noise gate cuts off the ADC output when its signal level is below a set threshold. The noise gate is controlled using the following parameters in the ALC Control 3 register (Address 0x4014):

- NGTYP[1:0]: The noise gate type is set to one of four modes by writing to the NGTYP[1:0] bits.
- NGEN: The noise gate function is enabled by writing to the NGEN bit.
- NGTHR[4:0]: The threshold for muting the output is set by writing to the NGTHR[4:0] bits.

One common problem with noise gate functions is chatter, where a small signal that is close to the noise gate threshold varies in amplitude, causing the noise gate function to open and close rapidly. This causes an unpleasant sound.

To reduce this effect, the noise gate in the ADAU1761 uses a combination of a timeout period and hysteresis. The timeout period is set to 250 ms, so the signal must consistently be below the threshold for 250 ms before the noise gate operates. Hysteresis is used so that the threshold for coming out of the mute state is 6 dB higher than the threshold for going into the mute state. There are four operating modes for the noise gate.

Noise Gate Mode 0 (see Figure 40) is selected by setting the NGTYP[1:0] bits to 00. In this mode, the current state of the PGA gain is held at its current state when the noise gate logic is activated. This prevents a large increase in background noise during periods of silence. When using this mode, it is advisable to use a relatively slow decay time. This is because the noise gate takes at least 250 ms to activate, and if the PGA gain has already increased to a large value during this time, the value at which the gain is held will be large.

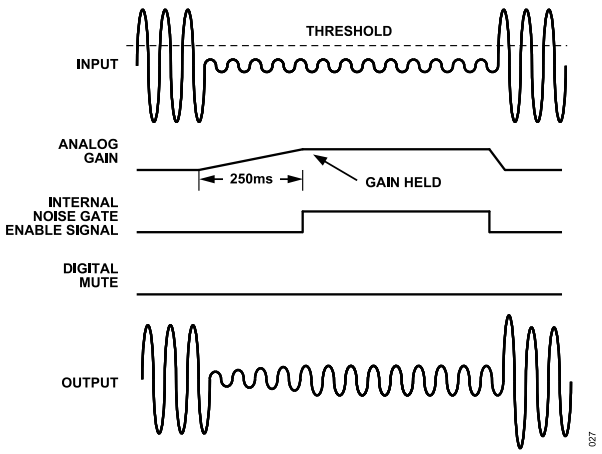


Figure 40. Noise Gate Mode 0 (PGA Gain Hold)

Noise Gate Mode 1 (see Figure 41) is selected by setting the NGTYP[1:0] bits to 01. In this mode, the ADAU1761 does a simple digital mute of the ADC output. Although this mode completely eliminates any background noise, the effect of an abrupt mute may not be pleasant to the ear.

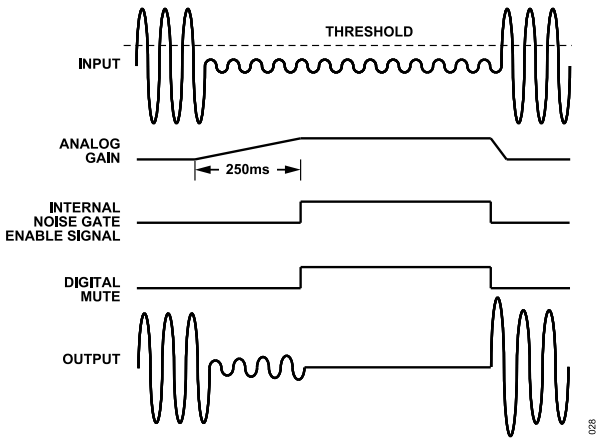


Figure 41. Noise Gate Mode 1 (Digital Mute)

Noise Gate Mode 2 (see Figure 42) is selected by setting the NGTYP[1:0] bits to 10. In this mode, the ADAU1761 improves the sound of the noise gate operation by first fading the PGA gain over a period of about 100 ms to the minimum PGA gain value. The ADAU1761 does not do a hard mute after the fade is complete, so some small background noise will still exist.

AUTOMATIC LEVEL CONTROL (ALC)

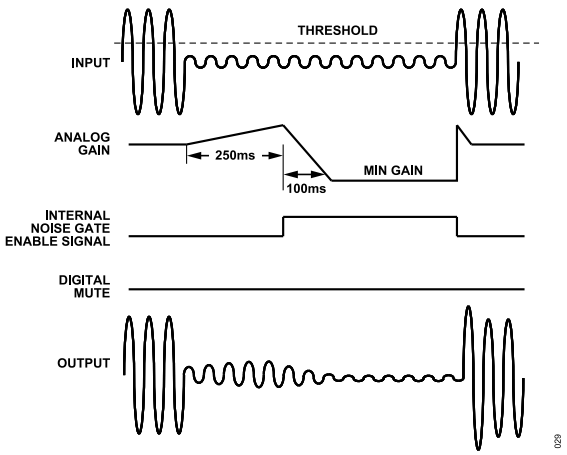


Figure 42. Noise Gate Mode 2 (Analog Fade)

Noise Gate Mode 3 (see Figure 43) is selected by setting the NGTYP[1:0] bits to 11. This mode is the same as Mode 2 except that at the end of the PGA fade gain interval, a digital mute is performed. In general, this mode is the best-sounding mode, because the audible effect of the digital hard mute is reduced by the fact that the gain has already faded to a low level before the mute occurs.

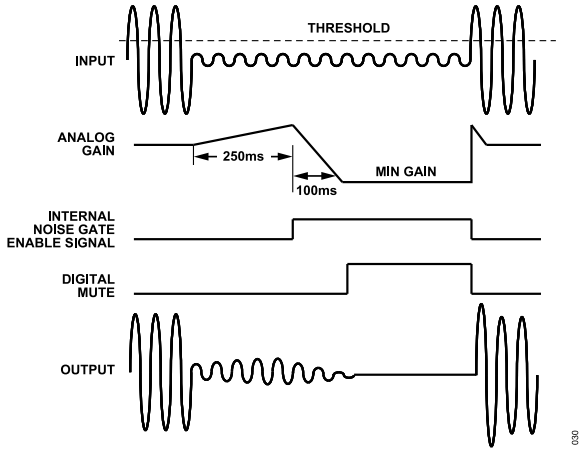


Figure 43. Noise Gate Mode 3 (Analog Fade/Digital Mute)

PLAYBACK SIGNAL PATH

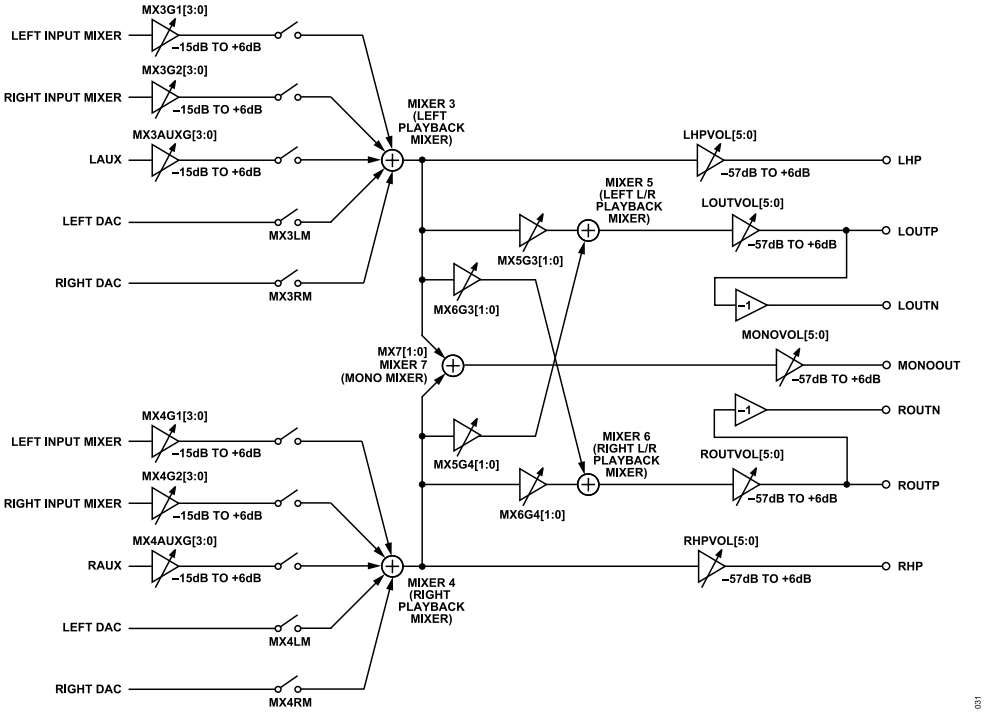


Figure 44. Playback Signal Path

OUTPUT SIGNAL PATHS

The outputs of the ADAU1761 can be configured as a variety of differential or single-ended outputs. All analog output pins are capable of driving headphone or earpiece speakers. There are selectable output paths for stereo signals or a downmixed mono output. The line outputs can drive a load of at least 10 kΩ or can be put into HP mode to drive headphones or earpiece speakers. The analog output pins are biased at AVDD/2.

With a 0 dBFS digital input and AVDD = 1.8 V, the full-scale output level is 500 mV rms; when AVDD = 3.3 V, the full-scale output level is 920 mV rms.

Signals are inverted through the mixers and volume controls. The result of this inversion is that the polarity of the differential outputs and the headphone outputs is preserved. The single-ended mono output is inverted. The DACs are noninverting.

Routing Flexibility

The playback path contains five mixers (Mixer 3 to Mixer 7) that perform the following functions:

- Mix signals from the record path and the DACs.
- Mix or swap the left and right channels.
- Mix a mono signal or generate a common-mode output.

Mixer 3 and Mixer 4 are dedicated to mixing signals from the record path and the DACs. Each of these two mixers can accept signals from the left and right DACs, the left and right input mixers, and the dedicated channel auxiliary input. Signals coming from the record path can be boosted or cut before the playback mixer.

For example, the MX4G2[3:0] bits set the gain from the output of Mixer 2 (right record channel) to the input of Mixer 4, hence the naming convention.

Signals coming from the DACs have digital volume attenuation controls set in Register R20 (left input digital volume register, Address 0x401A) and Register R21 (right input digital volume register, Address 0x401B).

HEADPHONE OUTPUT

The LHP and RHP pins can be driven by either a line output driver or a headphone driver by setting the HPMODE bit in Register R30 (playback headphone right volume control register, Address 0x4024). The headphone outputs can drive a load of at least 16 Ω.

Separate volume controls for the left and right channels range from −57 dB to +6 dB. Slew can be applied to all the playback volume controls using the ASLEW[1:0] bits in Register R34 (playback pop/click suppression register, Address 0x4028).

Capless Headphone Configuration

The headphone outputs can be configured in a capless output configuration with the MONOOUT pin used as a dc virtual ground reference. Figure 45 depicts a typical playback path in a capless headphone configuration. Table 19 lists the register settings for this configuration. As shown in this table, the MONOOUT pin outputs

PLAYBACK SIGNAL PATH

common mode (AVDD/2), which is used as the virtual headphone reference.

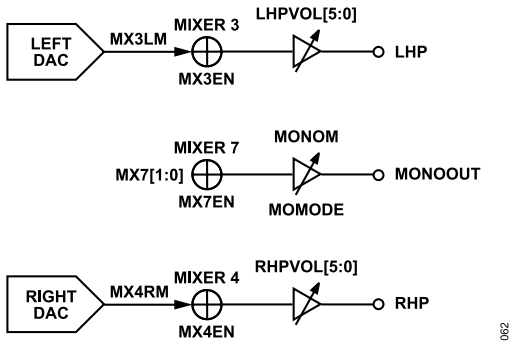


Figure 45. Capless Headphone Configuration Diagram

Table 19. Capless Headphone Register Settings

Register	Bit Name	Setting
R36	DACEN[1:0]	11 = both DACs on
R22	MX3EN	1 = enable Mixer 3
	MX3LM	1 = unmute left DAC input
R24	MX4EN	1 = enable Mixer 4
	MX4RM	1 = unmute right DAC input
R28	MX7EN	1 = enable Mixer 7
	MX7[1:0]	00 = common-mode output
R33	MONOM	1 = unmute mono output
	MOMODE	1 = headphone output
R29	LHPVOL[5:0]	Desired volume for LHP output
	LHPM	1 = unmute left headphone output
R30	HPMODE	1 = headphone output
	RHPVOL[5:0]	Desired volume for RHP output
	RHPM	1 = unmute right headphone output

Headphone Output Power-Up/Power-Down Sequencing

To prevent pops when turning on the headphone outputs, the user must wait at least 4 ms to unmute these outputs after enabling the headphone output with the HPMODE bit. This is because of an internal capacitor that must charge before these outputs can be used. Figure 46 and Figure 47 illustrate the headphone power-up/power-down sequencing.

For capless headphones, configure the MONOOUT pin before unmuting the headphone outputs.

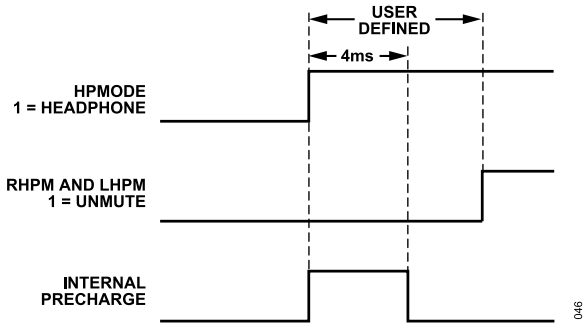


Figure 46. Headphone Output Power-Up Timing

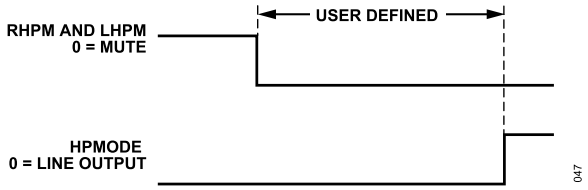


Figure 47. Headphone Output Power-Down Timing

Ground-Centered Headphone Configuration

The headphone outputs can also be configured as ground-centered outputs by placing coupling capacitors on the LHP and RHP pins. Ground-centered headphones should use the AGND pin as the ground reference.

When the headphone outputs are configured in this manner, the capacitors create a high-pass filter on the outputs. The corner frequency of this filter, at which point its attenuation is 3 dB, is calculated by the following formula:

$$f_{3dB} = 1/(2\pi \times R \times C) \quad (1)$$

where:
C is the capacitor value.
R is the impedance of the headphones.

For a typical headphone impedance of 16 Ω and a 47 µF capacitor, the corner frequency is 211 Hz.

Jack Detection

When the JACKDET/MICIN pin is set to the jack detect function, a flag on this pin can be used to mute the line outputs when headphones are plugged into the jack. This pin can be configured in Register R2 (digital microphone/jack detection control register, Address 0x4008). The JDFUNC[1:0] bits set the functionality of the JACKDET/MICIN pin.

Additional settings for jack detection include debounce time (JDDB[1:0] bits) and detection polarity (JDPOL bit). Because the jack detection and digital microphone share a pin, both functions cannot be used simultaneously.

PLAYBACK SIGNAL PATH

POP-AND-CLICK SUPPRESSION

Upon power-up, precharge circuitry is enabled to suppress pops and clicks. After power-up, the precharge circuitry can be put into a low power mode using the POPMODE bit in Register R34 (playback pop/click suppression register, Address 0x4028).

The precharge time depends on the capacitor value on the CM pin and the RC time constant of the load. For a typical line output load, the precharge time is between 2 ms and 3 ms. After this precharge time, the POPMODE bit can be set to low power mode.

Changing any register settings that affect the signal path can cause pops and clicks on the analog outputs. To avoid these pops and clicks, mute the appropriate outputs using Register R29 to Register R32 (Address 0x4023 to Address 0x4026). Unmute the analog outputs after the changes are made.

LINE OUTPUTS

The line output pins (LOUTP, LOUTN, ROUTP, and ROUTN) can be used to drive both differential and single-ended loads. In their default settings, these pins can drive typical line loads of 10 kΩ or greater, but they can also be put into headphone mode by setting the LOMODE bit in Register R31 (playback line output left volume control register, Address 0x4025) and the ROMODE bit in Register R32 (playback line output right volume control register, Address 0x4026). In headphone mode, the line output pins are capable of driving headphone and earpiece speakers of 16 Ω or greater. The output impedance of the line outputs is approximately 1 kΩ.

When the line output pins are used in single-ended mode, LOUTP and ROUTP should be used to output the signals, and LOUTN and ROUTN should be left unconnected.

The volume controls for these outputs range from −57 dB to +6 dB. Slew can be applied to all the playback volume controls using the ASLEW[1:0] bits in Register R34 (playback pop/click suppression register, Address 0x4028).

The MX5G4[1:0], MX5G3[1:0], MX6G3[1:0], and MX6G4[1:0] bits can all provide a 6 dB gain boost to the line outputs. This gain boost allows single-ended output signals to achieve 0 dBV (1.0 V rms) and differential output signals to achieve up to 6 dBV (2.0 V rms). For more information, see Register R26 (playback L/R mixer left (Mixer 5) line output control register, Address 0x4020) and Register R27 (playback L/R mixer right (Mixer 6) line output control register, Address 0x4021).

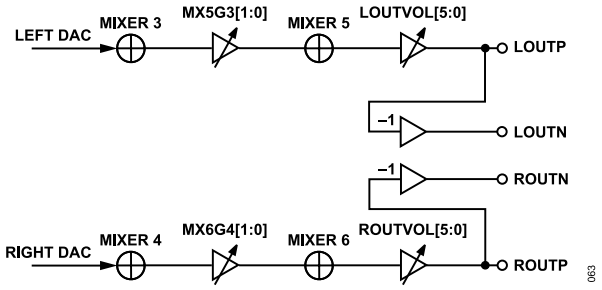


Figure 48. Differential Line Output Configuration

CONTROL PORTS

The ADAU1761 can operate in one of two control modes:

- I²C control
- SPI control

The ADAU1761 has both a 4-wire SPI control port and a 2-wire I²C bus control port. Both ports can be used to set the registers. The part defaults to I²C mode, but it can be put into SPI control mode by pulling the $\overline{\text{CLATCH}}$ pin low three times.

The control port is capable of full read/write operation for all addressable registers. The ADAU1761 must have a valid main clock in order to write to all registers except for Register R0 (Address 0x4000) and Register R1 (Address 0x4002).

All addresses can be accessed in both a single-address mode or a burst mode. The first byte (Byte 0) of a control port write contains the 7-bit chip address plus the R/$\overline{\text{W}}$ bit. The next two bytes (Byte 1 and Byte 2) together form the subaddress of the register location within the ADAU1761. This subaddress must be two bytes long because the memory locations within the ADAU1761 are directly addressable and their sizes exceed the range of single-byte addressing. All subsequent bytes (starting with Byte 3) contain the data, such as control port data, program data, or parameter data. The number of bytes per word depends on the type of data that is being written.

The ADAU1761 has several mechanisms for updating signal processing parameters in real time without causing pops or clicks. If large blocks of data need to be downloaded, the output of the DSP core can be halted (using the DSPRUN bit in the DSP run register, Address 0x40F6), new data can be loaded, and the device can be restarted. This is typically done during the booting sequence at start-up or when loading a new program into RAM.

The control port pins are multifunctional, depending on the mode in which the part is operating. Table 20 describes these multiple functions.

Table 20. Control Port Pin Functions

Pin Name	I²C Mode	SPI Mode
SCL/CCLK	SCL: input clock	CCLK: input clock
SDA/COUT	SDA: open-collector input/ output	COUT: output
ADDR1/CDATA	I²C Address Bit 1: input	CDATA: input
ADDR0/$\overline{\text{CLATCH}}$	I²C Address Bit 0: input	$\overline{\text{CLATCH}}$: input

BURST MODE WRITING AND READING

Burst mode addressing, where the subaddresses are automatically incremented at word boundaries, can be used for writing large amounts of data to contiguous registers. This increment happens automatically after a single-word write or read unless a stop condition is encountered (I²C) or $\overline{\text{CLATCH}}$ is brought high (SPI). A burst write starts like a single-word write, but following the first data-word, the data-word for the next immediate address can be written immediately without sending its two-byte address.

The registers in the ADAU1761 are one byte wide with the exception of the PLL control register, which is six bytes wide. The autoincrement feature knows the word length at each subaddress, so the subaddress does not need to be specified manually for each address in a burst write.

The subaddresses are autoincremented by 1 following each read or write of a data-word, regardless of whether there is a valid register or RAM word at that address. Address holes in the register map can be written to or read from without consequence. In the ADAU1761, these address holes exist at Address 0x4001, Address 0x4003 to Address 0x4007, Address 0x402E, Address 0x4032 to Address 0x4035, Address 0x4037 to Address 0x40BF, Address 0x40C5, Address 0x40CA to Address 0x40CF, Address 0x40D5 to Address 0x40EA, and Address 0x40EC to Address 0x40F1. A single-byte write to these registers is ignored by the ADAU1761, and a read returns a single byte 0x00.

I²C PORT

The ADAU1761 supports a 2-wire serial (I²C-compatible) microprocessor bus driving multiple peripherals. Two pins, serial data (SDA) and serial clock (SCL), carry information between the ADAU1761 and the system I²C main controller. In I²C mode, the ADAU1761 is always a subordinate on the bus, meaning that it cannot initiate a data transfer. Each subordinate device is recognized by a unique address. The address and R/$\overline{\text{W}}$ byte format is shown in Table 21. The address resides in the first seven bits of the I²C write. Bits[5:6] of the I²C address for the ADAU1761 are set by the levels on the ADDR1 and ADDR0 pins. The LSB of the address—the R/$\overline{\text{W}}$ bit—specifies either a read or write operation. Logic Level 1 corresponds to a read operation, and Logic Level 0 corresponds to a write operation.

Table 21. ADAU1761 I²C Address and Read/$\overline{\text{Write}}$ Byte Format

Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7
0	1	1	1	0	ADDR1	ADDR0	R/$\overline{\text{W}}$

The SDA and SCL pins should each have a 2 kΩ pull-up resistor on the line connected to it. The voltage on these signal lines should not be higher than IOVDD (1.8 V to 3.3 V).

Addressing

Initially, each device on the I²C bus is in an idle state and monitors the SDA and SCL lines for a start condition and the proper address. The I²C main initiates a data transfer by establishing a start condition, defined by a high-to-low transition on SDA while SCL remains high. This indicates that an address/ data stream follows. All devices on the bus respond to the start condition and shift the next eight bits (the 7-bit address plus the R/$\overline{\text{W}}$ bit) MSB first. The device that recognizes the transmitted address responds by pulling the data line low during the ninth clock pulse. This ninth bit is known as an acknowledge bit. All other devices withdraw from the bus at this point and return to the idle condition.

CONTROL PORTS

The R/$\overline{W}$ bit determines the direction of the data. A Logic 0 on the LSB of the first byte means that the main will write information to the peripheral, whereas a Logic 1 means that the main will read information from the peripheral after writing the subaddress and repeating the start address. A data transfer takes place until a stop condition is encountered. A stop condition occurs when SDA transitions from low to high while SCL is held high. Figure 49 shows the timing of an I^{2}C write, and Figure 50 shows an I^{2}C read.

Stop and start conditions can be detected at any stage during the data transfer. If these conditions are asserted out of sequence with normal read and write operations, the ADAU1761 immediately jumps to the idle condition. During a given SCL high period, the user should only issue one start condition, one stop condition, or a single stop condition followed by a single start condition. If an invalid subaddress is issued by the user, the ADAU1761 does not issue an acknowledge and returns to the idle condition.

If the user exceeds the highest subaddress while in autoincrement mode, one of two actions is taken. In read mode, the ADAU1761 outputs the highest subaddress register contents until the main device issues a no acknowledge, indicating the end of a read. A no acknowledge condition is where the SDA line is not pulled low on the ninth clock pulse on SCL. If the highest subaddress location is reached while in write mode, the data for the invalid byte is not loaded into any subaddress register, a no acknowledge is issued by the ADAU1761, and the part returns to the idle condition.

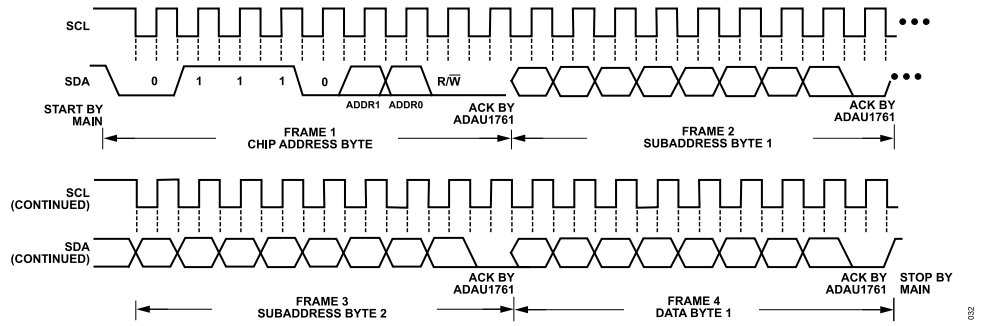


Figure 49. I^{2}C Write to ADAU1761 Clocking

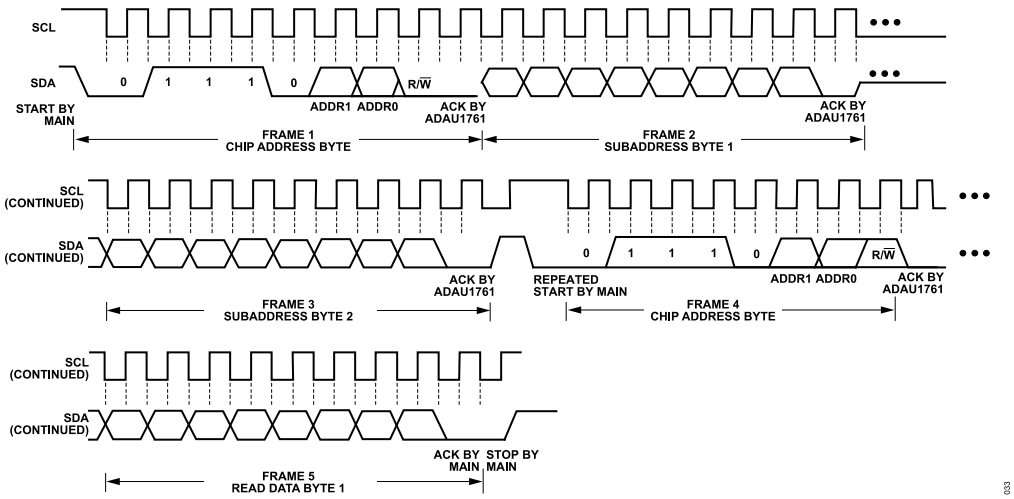


Figure 50. I^{2}C Read from ADAU1761 Clocking

CONTROL PORTS

I²C Read and Write Operations

Figure 51 shows the format of a single-word write operation. Every ninth clock pulse, the ADAU1761 issues an acknowledge by pulling SDA low.

Figure 52 shows the format of a burst mode write sequence. This figure shows an example of a write to sequential single-byte registers. The ADAU1761 increments its subaddress register after every byte because the requested subaddress corresponds to a register or memory area with a 1-byte word length.

Figure 53 shows the format of a single-word read operation. Note that the first R/$\overline{W}$ bit is 0, indicating a write operation. This is because the subaddress still needs to be written to set up the internal address. After the ADAU1761 acknowledges the receipt of the subaddress, the main must issue a repeated start command followed by the chip address byte with the R/$\overline{W}$ bit set to 1 (read). This causes the ADAU1761 SDA to reverse and begin driving data back to the main. The main then responds every ninth pulse with an acknowledge pulse to the ADAU1761.

Figure 54 shows the format of a burst mode read sequence. This figure shows an example of a read from sequential single-byte registers. The ADAU1761 increments its subaddress register after every byte because the requested subaddress corresponds to a register or memory area with a 1-byte word length. The ADAU1761 always decodes the subaddress and sets the auto-increment circuit so that the address increments after the appropriate number of bytes.

Figure 51 to Figure 54 use the following abbreviations:

S = start bit

P = stop bit

AM = acknowledge by main

AS = acknowledge by subordinate

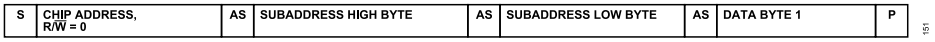


Figure 51. Single-Word I²C Write Format

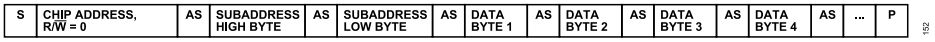


Figure 52. Burst Mode I²C Write Format



Figure 53. Single-Word I²C Read Format

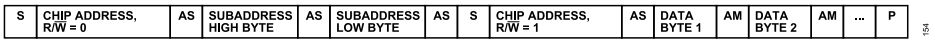


Figure 54. Burst Mode I²C Read Format

CONTROL PORTS

SPI PORT

By default, the ADAU1761 is in I²C mode, but it can be put into SPI control mode by pulling $\overline{\text{CLATCH}}$ low three times. This is done by performing three dummy writes to the SPI port (the ADAU1761 does not acknowledge these three writes). Beginning with the fourth SPI write, data can be written to or read from the IC. The ADAU1761 can be taken out of SPI mode only by a full reset initiated by power cycling the IC.

The SPI port uses a 4-wire interface, consisting of the $\overline{\text{CLATCH}}$, CCLK, CDATA, and COUT signals, and it is always a subordinate port. The $\overline{\text{CLATCH}}$ signal should go low at the beginning of a trans-action and high at the end of a transaction. The CCLK signal latches CDATA on a low-to-high transition. COUT data is shifted out of the ADAU1761 on the falling edge of CCLK and should be clocked into a receiving device, such as a microcontroller, on the CCLK rising edge. The CDATA signal carries the serial input data, and the COUT signal carries the serial output data. The COUT signal remains three-state until a read operation is requested. This allows other SPI-compatible peripherals to share the same readback line. All SPI transactions have the same basic format shown in Table 23. A timing diagram is shown in Figure 4. All data should be written MSB first.

Chip Address R/$\overline{\text{W}}$

The LSB of the first byte of an SPI transaction is an R/$\overline{\text{W}}$ bit. This bit determines whether the communication is a read (Logic Level 1) or a write (Logic Level 0). This format is shown in Table 22.

Table 22. ADAU1761 SPI Address and Read/$\overline{\text{Write}}$ Byte Format

Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7
0	0	0	0	0	0	0	R/$\overline{\text{W}}$

Subaddress

The 16-bit subaddress word is decoded into a location in one of the registers. This subaddress is the location of the appropriate register. The MSBs of the subaddress are zero-padded to bring the word to a full 2-byte length.

Data Bytes

The number of data bytes varies according to the register being accessed. During a burst mode write, an initial subaddress is written followed by a continuous sequence of data for consecutive register locations.

A sample timing diagram for a single-word SPI write operation to a register is shown in Figure 55. A sample timing diagram of a single-word SPI read operation is shown in Figure 56. The COUT pin goes from being three-state to being driven at the beginning of Byte 3. In this example, Byte 0 to Byte 2 contain the addresses and R/$\overline{\text{W}}$ bit, and subsequent bytes carry the data.

Table 23. Generic Control Word Format

Byte 0	Byte 1	Byte 2	Byte 3	Byte 4[1]
chip_adr[6:0], R/$\overline{\text{W}}$	subaddr[15:8]	subaddr[7:0]	data	data

[1] Continues to end of data.

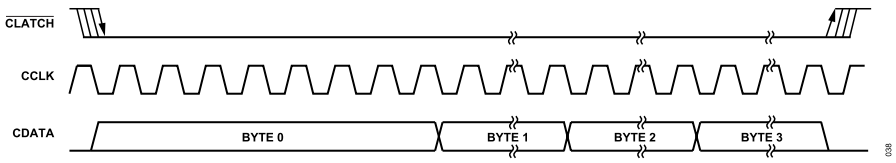


Figure 55. SPI Write to ADAU1761 Clocking (Single-Word Write Mode)

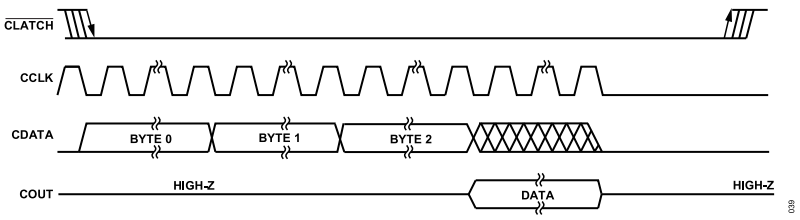


Figure 56. SPI Read from ADAU1761 Clocking (Single-Word Read Mode)

SERIAL DATA INPUT/OUTPUT PORTS

The flexible serial data input and output ports of the ADAU1761 can be set to accept or transmit data in 2-channel format or in a 4-channel or 8-channel TDM stream to interface to external ADCs or DACs. Data is processed in twos complement, MSB first format. The left channel data field always precedes the right channel data field in 2-channel streams. In TDM mode, Slot 0 to Slot 3 are in the first half of the audio frame, and Slot 4 to Slot 7 are in the second half of the frame. The serial modes and the position of the data in the frame are set in Register R15 to Register R18 (serial port and converter control registers, Address 0x4015 to Address 0x4018).

If the PLL of the ADAU1761 is not used, the serial data clocks must be synchronous with the ADAU1761 main clock input. The LRCLK and BCLK pins are used to clock both the serial input and output ports. The ADAU1761 can be set as the main or the subordinate in a system. Because there is only one set of serial data clocks, the input and output ports must always be both main or both subordinate.

Register R15 and Register R16 (serial port control registers, Address 0x4015 and Address 0x4016) allow control of clock polarity and data input modes. The valid data formats are I²S, left-justified, right-justified (24-/20-/18-/16-bit), and TDM. In all modes except for the right-justified modes, the serial port inputs an arbitrary number of bits up to a limit of 24. Extra bits do not cause an error, but they are truncated internally.

The serial port can operate with an arbitrary number of BCLK transitions in each LRCLK frame. The LRCLK in TDM mode can be input to the ADAU1761 either as a 50% duty cycle clock or as a bit-wide pulse.

When the LRCLK is set as a pulse, a 47 pF capacitor should be connected between the LRCLK pin and ground (see Figure 57). This capacitor is necessary in both main and subordinate modes to properly align the LRCLK signal to the serial data stream.

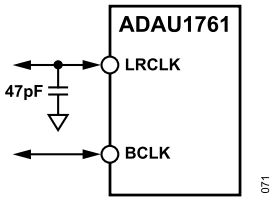


Figure 57. LRCLK Capacitor Alignment, TDM Pulse Mode

In TDM 8 mode, the ADAU1761 can be a main for f_S up to 48 kHz. Table 24 lists the modes in which the serial output port can function.

Table 24. Serial Output Port Main/Subordinate Mode Capabilities

f_S	2-Channel Modes (I²S, Left-Justified, Right-Justified)	8-Channel TDM
48 kHz	main and subordinate	main and subordinate
96 kHz	main and subordinate	Subordinate

Table 25 describes the proper configurations for standard audio data formats.

Table 25. Data Format Configurations

Format	LRCLK Polarity (LRPOL)	LRCLK Mode (LRMOD)	BCLK Polarity (BPOL)	BCLK Cycles/Audio Frame (BPF[2:0])	Data Delay from LRCLK Edge (LRDEL[1:0])
I²S (see Figure 58)	Frame begins on falling edge	50% duty cycle	Data changes on falling edge	32 to 64	Delayed from LRCLK edge by 1 BCLK
Left-Justified (see Figure 59)	Frame begins on rising edge	50% duty cycle	Data changes on falling edge	32 to 64	Aligned with LRCLK edge
Right-Justified (see Figure 60)	Frame begins on rising edge	50% duty cycle	Data changes on falling edge	32 to 64	Delayed from LRCLK edge by 8 or 16 BCLKs
TDM with Clock (see Figure 61)	Frame begins on falling edge	50% duty cycle	Data changes on falling edge	64 to 256	Delayed from start of word clock by 1 BCLK
TDM with Pulse (see Figure 62)	Frame begins on rising edge	Pulse	Data changes on falling edge	64 to 256	Delayed from start of word clock by 1 BCLK

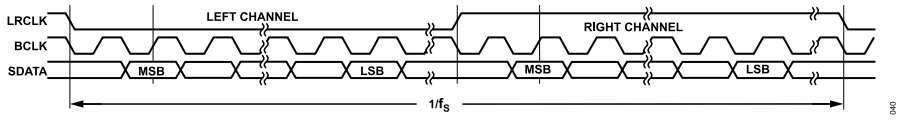


Figure 58. I²S Mode—16 Bits to 24 Bits per Channel

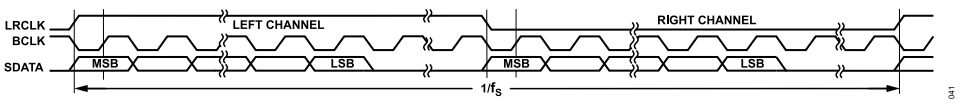


Figure 59. Left-Justified Mode—16 Bits to 24 Bits per Channel

SERIAL DATA INPUT/OUTPUT PORTS

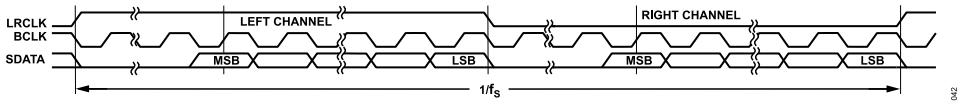


Figure 60. Right-Justified Mode—16 Bits to 24 Bits per Channel

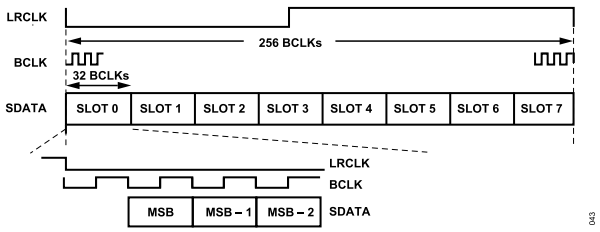


Figure 61. TDM 8 Mode

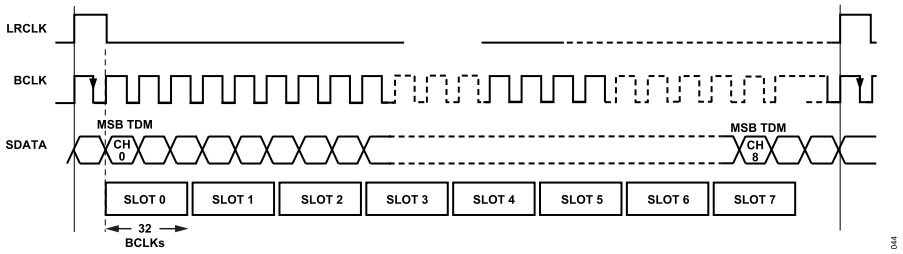


Figure 62. TDM 8 Mode with Pulse Word Clock

APPLICATIONS INFORMATION

POWER SUPPLY BYPASS CAPACITORS

Each analog and digital power supply pin should be bypassed to its nearest appropriate ground pin with a single 100 nF capacitor. The connections to each side of the capacitor should be as short as possible, and the trace should stay on a single layer with no vias. For maximum effectiveness, locate the capacitor equi-distant from the power and ground pins or, when equidistant placement is not possible, slightly closer to the power pin. Thermal connections to the ground planes should be made on the far side of the capacitor.

Each supply signal on the board should also be bypassed with a single bulk capacitor (10 µF to 47 µF).

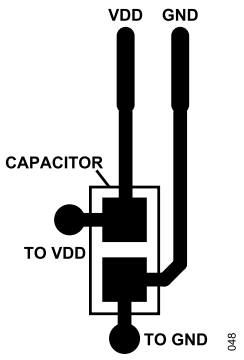


Figure 63. Recommended Power Supply Bypass Capacitor Layout

GSM NOISE FILTER

In mobile phone applications, excessive 217 Hz GSM noise on the analog supply pins can degrade the audio quality. To avoid this problem, it is recommended that an L-C filter be used in series with the bypass capacitors for the AVDD pins. This filter should consist of a 1.2 nH inductor and a 9.1 pF capacitor in series between AVDD and ground, as shown in Figure 64.

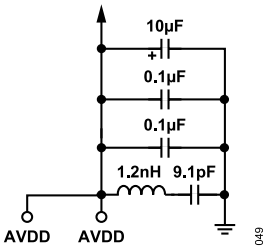


Figure 64. GSM Filter on the Analog Supply Pins

GROUNDING

A single ground plane should be used in the application layout. Components in an analog signal path should be placed away from digital signals.

EXPOSED PAD PCB DESIGN

The ADAU1761 has an exposed pad on the underside of the LFCSP. This pad is used to couple the package to the PCB for heat dissipation when using the outputs to drive earpiece or headphone loads. When designing a board for the ADAU1761, special consideration should be given to the following:

- A copper layer equal in size to the exposed pad should be on all layers of the board, from top to bottom, and should connect somewhere to a dedicated copper board layer (see Figure 65).
- Vias should be placed to connect all layers of copper, allowing for efficient heat and energy conductivity. For an example, see Figure 66, which has nine vias arranged in a 3 inch × 3 inch grid in the pad area.

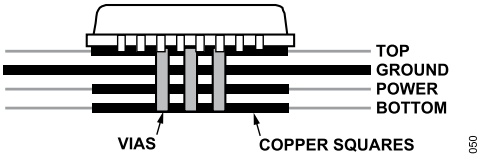


Figure 65. Exposed Pad Layout Example, Side View

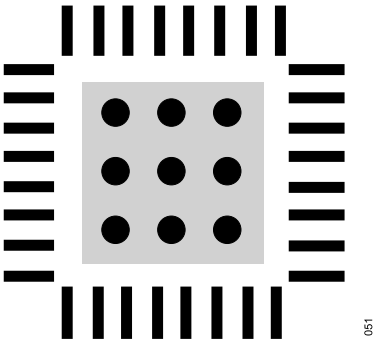


Figure 66. Exposed Pad Layout Example, Top View

DSP CORE

SIGNAL PROCESSING

The ADAU1761 is designed to provide all audio signal processing functions commonly used in stereo or mono low power record and playback systems. The signal processing flow is designed using the SigmaStudio software, which allows graphical entry and real-time control of all signal processing functions.

Many of the signal processing functions are coded using full, 56-bit, double-precision arithmetic data. The input and output word lengths of the DSP core are 24 bits. Four extra headroom bits are used in the processor to allow internal gains of up to 24 dB without clipping. Additional gains can be achieved by initially scaling down the input signal in the DSP signal flow.

ARCHITECTURE

The DSP core consists of a simple 28-/56-bit multiply-accumulate (MAC) unit with two sources: a data source and a coefficient source. The data source can come from the data RAM, a ROM table of commonly used constant values, or the audio inputs to the core. The coefficient source can come from the parameter RAM or from a ROM table of commonly used constant values.

The two sources are multiplied in a 28-bit fixed-point multiplier and then the signal is input to the 56-bit adder; the result is usually stored in one of three 56-bit accumulator registers. The accumulators can be output from the core (in 28-bit format) or can optionally be written back into the data or parameter RAMs.

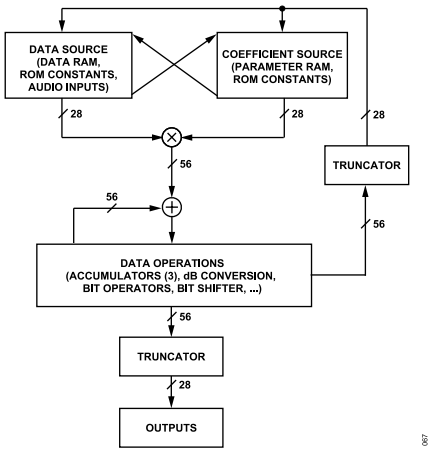


Figure 67. Simplified DSP Core Architecture

PROGRAM COUNTER

The execution of instructions in the core is governed by a program counter, which sequentially steps through the addresses of the program RAM. The program counter starts every time that a new audio frame is clocked into the core. SigmaStudio inserts a jump-to-start command at the end of every program. The program counter increments sequentially until it reaches this command and then jumps to the program start address and waits for the next audio frame to clock into the core.

FEATURES

The SigmaDSP core was designed specifically for audio processing and therefore includes several features intended for maximizing efficiency. These include hardware decibel conversion and audio-specific ROM constants.

STARTUP

Before the DSPRUN bit is set or any settings are written to the parameter RAM, the DSP core must be enabled by setting the DSPEN bit in Register R61 (Address 0x40F5).

The following steps should be performed every time that a new program is loaded to the SigmaDSP core, or any time that the DSPRUN bit is disabled and reenabled.

1. Set the DSPSR[3:0] bits in Register R57 (Address 0x40EB) to 1111 (none).
2. Set the DSPRUN bit in Register R62 (Address 0x40F6) to 0.
3. Download the rest of the registers, the program RAM, and the parameter RAM.
4. Set the DSPRUN bit in Register R62 to 1.
5. Set the DSPSR[3:0] bits in Register R57 to the operational setting (default value is 0001).

Changing any register setting or RAM can cause pops and clicks on the analog outputs. To avoid these pops and clicks, mute the appropriate outputs using Register R29 to Register R32 (Address 0x4023 to Address 0x4026). Unmute the analog out-puts after the startup procedure is completed.

NUMERIC FORMATS

DSP systems commonly use a standard numeric format. Fractional numeric systems are specified by an A.B format, where A is the number of bits to the left of the decimal point and B is the number of bits to the right of the decimal point.

The ADAU1761 uses numeric format 5.23 for both the parameter and data values.

Numeric Format 5.23

Linear range: −16.0 to (+16.0 − 1 LSB)

Examples:

1000 0000 0000 0000 0000 0000 0000 = −16.0

1110 0000 0000 0000 0000 0000 0000 = −4.0

1111 1000 0000 0000 0000 0000 0000 = −1.0

1111 1110 0000 0000 0000 0000 0000 = −0.25

1111 1111 0011 0011 0011 0011 0011 = −0.1

1111 1111 1111 1111 1111 1111 1111 = (1 LSB below 0)

0000 0000 0000 0000 0000 0000 0000 = 0

DSP CORE

0000 0000 1100 1100 1100 1100 1101 = 0.1

0000 0010 0000 0000 0000 0000 0000 = 0.25

0000 1000 0000 0000 0000 0000 0000 = 1.0

0010 0000 0000 0000 0000 0000 0000 = 4.0

0111 1111 1111 1111 1111 1111 1111 = (16.0 − 1 LSB)

The serial port accepts up to 24 bits on the input and is sign-extended to the full 28 bits of the DSP core. This allows internal gains of up to 24 dB without internal clipping.

A digital clipper circuit is used between the output of the DSP core and the DACs or serial port outputs (see Figure 68). This circuit clips the top four bits of the signal to produce a 24-bit output with a range of 1.0 (minus 1 LSB) to −1.0. Figure 68 shows the maximum signal levels at each point in the data flow in both binary and decibel levels.

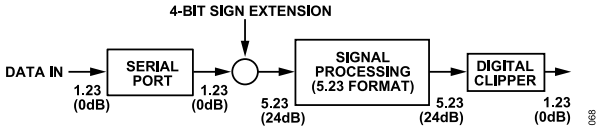


Figure 68. Numeric Precision and Clipping Structure

PROGRAMMING

On power-up, the ADAU1761 must be configured with a clocking scheme and then loaded with register settings. After the codec signal path is set up, the DSP core can be programmed. There are 1024 instruction cycles per audio sample, resulting in an internal clock rate of 49.152 MHz when f_S = 48 kHz.

The part can be programmed easily using SigmaStudio, a graphical tool provided by Analog Devices (see Figure 69). No knowledge of writing line-level DSP code is required. More information about SigmaStudio can be found at www.analog.com.

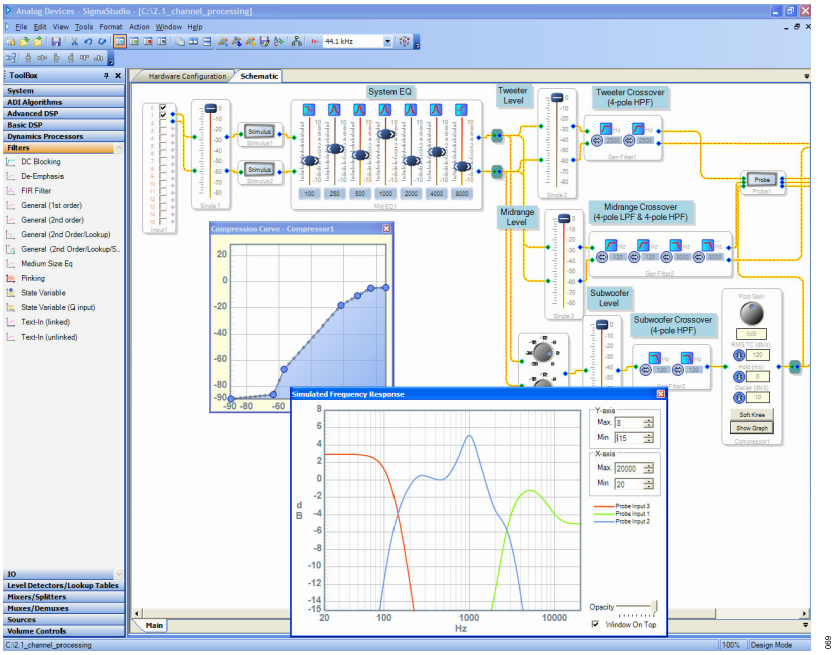


Figure 69. SigmaStudio Screen Shot

PROGRAM RAM, PARAMETER RAM, AND DATA RAM

Table 26. RAM Map and Read/Write Modes

Memory	Size	Address Range	Read	Write	Write Modes
Parameter RAM	1024 × 32	0 to 1023 (0x0000 to 0x03FF)	Yes	Yes	Direct, safeload
Program RAM	1024 × 40	2048 to 3071 (0x0800 to 0x0BFF)	Yes	Yes	Direct

Table 26 shows the RAM map (the ADAU1761 register map is provided in the Control Registers section). The address space encompasses a set of registers and three RAMs: program, parameter, and data. The program RAM and parameter RAM are not initialized on power-up and are in an unknown state until written to.

PROGRAM RAM

The program RAM contains the 40-bit operation codes that are executed by the core. The SigmaStudio compiler calculates maximum instructions per frame for a project and generates an error when the value exceeds the maximum allowable instructions per frame based on the sample rate of the signals in the core.

Because the end of a program contains a jump-to-start command, the unused program RAM space does not need to be filled with no-operation (NOP) commands.

PARAMETER RAM

The parameter RAM is 32 bits wide and occupies Address 0 to Address 1023. Each parameter is padded with four 0s before the MSB to extend the 28-bit word to a full 4-byte width. The data format of the parameter RAM is twos complement, 5.23. This means that the coefficients can range from +16.0 (minus 1 LSB) to −16.0, with 1.0 represented by the binary word 0000 1000 0000 0000 0000 0000 0000 or by the hexadecimal word 0x00 0x80 0x00 0x00.

The parameter RAM can be written to directly or with a safe-load write. The direct write mode of operation is typically used during a complete new loading of the RAM using burst mode addressing to avoid any clicks or pops in the outputs. Note that this mode can be used during live program execution, but because there is no handshaking between the core and the control port, the parameter RAM is unavailable to the DSP core during control writes, resulting in pops and clicks in the audio stream.

SigmaStudio automatically assigns the first eight positions to safe-load parameters; therefore, project-specific parameters start at Address 0x0008.

The parameter RAM should not be written to until the DSPEN bit has been set in Register R61 (Address 0x40F5).

DATA RAM

The ADAU1761 data RAM is used to store audio data-words for processing, as well as certain run-time parameters. SigmaStudio provides the data and address information for writing to and reading from the data RAM.

When implementing blocks, such as delays, that require large amounts of data RAM space, data RAM utilization should be taken into account. The SigmaDSP core processes delay times in one-sample increments; therefore, the total pool of delay available to the user equals 4096 multiplied by the sample period. For a $f_{S,DSP}$ of 48 kHz, the pool of available delay is a maximum of about 86 ms, where $f_{S,DSP}$ is the DSP core sampling rate. In practice, this much data memory is not available to the user because every block in a design uses a few data memory locations for its processing. In most DSP programs, this does not significantly affect the total delay time. The SigmaStudio compiler manages the data RAM and indicates whether the number of addresses needed in the design exceeds the maximum number available.

READ/WRITE DATA FORMATS

The read/write formats of the control port are designed to be byte oriented to allow for easy programming of common micro-controller chips. To fit into a byte-oriented format, 0s are added to the data fields before the MSB to extend the data-word to eight bits. For example, 28-bit words written to the parameter RAM are preceded by four leading 0s to equal 32 bits (four bytes); 40-bit words written to the program RAM are not preceded by 0s because they are already a full five bytes. These zero-padded data fields are appended to a 3-byte field consisting of a 7-bit chip address, a read/write bit, and a 16-bit RAM/register address. The control port knows how many data bytes to expect based on the address given in the first three bytes.

The total number of bytes for a single-location write command can vary from one byte (for a control register write) to five bytes (for a program RAM write). Burst mode can be used to fill contiguous register or RAM locations. A burst mode write begins by writing the address and data of the first RAM or register location to be written. Rather than ending the control port transaction (by issuing a stop command in I^2C mode or by bringing the $\overline{CLATCH}$ signal high in SPI mode after the data-word), as would be done in a single-address write, the next data-word can be written immediately without specifying its address. The ADAU1761 control port auto-increments the address of each write even across the boundaries of the different RAMs and registers. Table 28 and Table 30 show examples of burst mode writes.

PROGRAM RAM, PARAMETER RAM, AND DATA RAM

Table 27. Parameter RAM Read/Write Format (Single Address)

Byte 0	Byte 1	Byte 2	Byte 3	Bytes[4:6]
chip_adr[6:0], R/$\overline{W}$	param_adr[15:8]	param_adr[7:0]	0000, param[27:24]	param[23:0]

Table 28. Parameter RAM Block Read/Write Format (Burst Mode)

Byte 0	Byte 1	Byte 2	Byte 3	Bytes[4:6]	Bytes[7:10]	Bytes[11:14]
chip_adr[6:0], R/$\overline{W}$	param_adr[15:8]	param_adr[7:0]	0000, param[27:24]	param[23:0]		
			<—param_adr—>		param_adr + 1	param_adr + 2

Table 29. Program RAM Read/Write Format (Single Address)

Byte 0	Byte 1	Byte 2	Bytes[3:7]
chip_adr[6:0], R/$\overline{W}$	prog_adr[15:8]	prog_adr[7:0]	prog[39:0]

Table 30. Program RAM Block Read/Write Format (Burst Mode)

Byte 0	Byte 1	Byte 2	Bytes[3:7]	Bytes[8:12]	Bytes[13:17]
chip_adr[6:0], R/$\overline{W}$	prog_adr[15:8]	prog_adr[7:0]	prog[39:0]		
			<—prog_adr—>	prog_adr + 1	prog_adr + 2

SOFTWARE SAFELOAD

To update parameters in real time while avoiding pop and click noises on the output, the ADAU1761 uses a software safeload mechanism. The software safeload mechanism enables the SigmaDSP core to load new parameters into RAM while guaranteeing that the parameters are not in use. This prevents an undesirable condition where an instruction could execute with a mix of old and new parameters.

SigmaStudio sets up the necessary code and parameters automatically for new projects. The safeload code, along with other initialization code, fills the first 39 locations in program RAM. The first eight parameter RAM locations (Address 0x0000 to Address 0x0007) are configured by default in SigmaStudio as described in Table 31.

Table 31. Software Safeload Parameter RAM Defaults

Address (Hex)	Function
0x0000	Modulo RAM size
0x0001	Safeload Data 1
0x0002	Safeload Data 2
0x0003	Safeload Data 3
0x0004	Safeload Data 4
0x0005	Safeload Data 5
0x0006	Safeload target address (offset of −1)
0x0007	Number of words to write/safeload trigger

Address 0x0000, which controls the modulo RAM size, is set by SigmaStudio and is based on the dynamic address generator mode of the project.

Parameter RAM Address 0x0001 to Address 0x0005 are the five data slots for storing the data to be safeloaded. The safeload parameter space contains five data slots by default because most standard signal processing algorithms have five parameters or less.

Address 0x0006 is the target address in parameter RAM (with an offset of −1). This designates the first address to be written. If more than one word is written, the address increments automatically for each data-word. Up to five sequential parameter RAM locations can be updated with safeload during each audio frame. The target address offset of −1 is used because the write address is calculated relative to the address of the data, which starts at Address 0x0001. Therefore, to update a parameter at Address 0x000A, the target address is 0x0009.

Address 0x0007 designates the number of words to be written into the parameter RAM during the safeload. A biquad filter uses all five safeload data addresses. A simple mono gain cell uses only one safeload data address. Writing to Address 0x0007 also triggers the safeload write to occur in the next audio frame.

The safeload mechanism is software based and executes once per audio frame. Therefore, system designers must take care when designing the communication protocol. A delay equal to or greater than the sampling period (the inverse of sampling frequency) is required between each safeload write. A sample rate of 48 kHz equates to a delay of at least 21 µs. If this delay is not observed, the downloaded data is corrupted.

SOFTWARE SLEW

When the values of signal processing parameters are changed abruptly in real time, they sometimes cause pop and click sounds to appear on the audio outputs. To avoid pops and clicks, some algorithms in SigmaStudio implement a software slew functionality. Algorithms using software slew set a target value for a parameter and continuously update the value of that parameter until it reaches the target.

The target value takes an additional space in parameter RAM, and the current value of the parameter is updated in the non-modulo section of data RAM. Assignment of parameters and nonmodulo

PROGRAM RAM, PARAMETER RAM, AND DATA RAM

data RAM is handled by the SigmaStudio compiler and does not need to be programmed manually.

Slew parameters can follow several different curves, including an RC-type curve and a linear curve. These curve types are coded into each algorithm and cannot be modified by the user.

Because algorithms that use software slew generally require more RAM than their nonslew equivalents, they should be used only in situations where a parameter will change during operation of the device.

Figure 70 shows an example of volume slew applied to a sine wave.

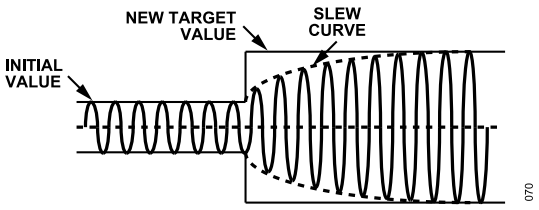


Figure 70. Example of Volume Slew

GENERAL-PURPOSE INPUT/OUTPUT

The serial data input/output pins (Pin 26 to Pin 29) are shared with the general-purpose input/output function. Each of these four pins can be set to only one of these functions. The function of these pins is set in the serial data/GPIO pin configuration register (Register R60, Address 0x40F4).

The GPIOx pins can be used as inputs or outputs. These pins are readable and can be set through the control port or directly by the SigmaDSP core. When configured as inputs, the GPIOx pins can be used with push-button switches or rotary encoders to control DSP program settings. These pins can also be used with digital outputs to drive LEDs or external logic to indicate the status of internal signals and control other devices. Examples of this use include indicating signal overload, signal present, and button press confirmation.

When configured as an output, each GPIO pin can typically drive 2 mA, which is enough current to directly drive some high efficiency LEDs. Standard LEDs require about 20 mA of current and can be driven from a GPIO output with an external transistor or buffer. Because of problems that can arise from simultaneously driving or sinking a large amount of current on many pins, avoid connecting high efficiency LEDs directly to many or all of the GPIO pins when designing the application.

If many LEDs are required, use an external driver. When the GPIO pins are configured as open-collector outputs, they should be pulled up to a maximum voltage equal to the voltage set on IOVDD.

The configuration of the GPIO functions is set up in the GPIO pin control registers (Register R48 to Register R51, Address 0x40C6 to Address 0x40C9).

GPIO PINS SET FROM THE CONTROL PORT

The GPIO pins can also be configured to be directly controlled from the I^2C/SPI control port. When the pins are set to this mode, four memory locations are enabled for the GPIO pin settings. The physical settings on the GPIO pins mirror the settings of the LSB of these 4-byte-wide memory locations.

Table 32. GPIOx Pin Memory Settings (Set from Control Port)

Memory Location			
Decimal	**Hex**	**Bits[31:1]**	**Bit 0**
1568	0x0620	Reserved	GPIO0SET
1569	0x0621	Reserved	GPIO1SET
1570	0x0622	Reserved	GPIO2SET
1571	0x0623	Reserved	GPIO3SET

CONTROL REGISTERS

Table 33. Register Map

Reg	Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default
R0	0x4000	Clock control	Reserved				CLKSRC	INFREQ[1:0]		COREN	00000000
R1	0x4002	PLL control	M[15:8]								00000000
			M[7:0]								11111101
			N[15:8]								00000000
			N[7:0]								00001100
			Reserved	R[3:0]				X[1:0]		Type	00010000
			Reserved						Lock	PLLEN	00000000
R2	0x4008	Dig mic/jack detect	JDDB[1:0]		JDFUNC[1:0]		Reserved			JDPOL	00000000
R3	0x4009	Rec power mgmt	Reserved	MXBIAS[1:0]		ADCBIAS[1:0]		RBIAS[1:0]		Reserved	00000000
R4	0x400A	Rec Mixer Left 0	Reserved	LINPG[2:0]			LINNG[2:0]			MX1EN	00000000
R5	0x400B	Rec Mixer Left 1	Reserved			LDBOOST[1:0]		MX1AUXG[2:0]			00000000
R6	0x400C	Rec Mixer Right 0	Reserved	RINPG[2:0]			RINNG[2:0]			MX2EN	00000000
R7	0x400D	Rec Mixer Right 1	Reserved			RDBOOST[1:0]		MX2AUXG[2:0]			00000000
R8	0x400E	Left diff input vol	LDVOL[5:0]						LDMUTE	LDEN	00000000
R9	0x400F	Right diff input vol	RDVOL[5:0]						RDMUTE	RDEN	00000000
R10	0x4010	Record mic bias	Reserved				MPERF	MBI	Reserved	MBIEN	00000000
R11	0x4011	ALC 0	PGASLEW[1:0]		ALCMAX[2:0]			ALCSEL[2:0]			00000000
R12	0x4012	ALC 1	ALCHOLD[3:0]				ALCTARG[3:0]				00000000
R13	0x4013	ALC 2	ALCATCK[3:0]				ALCDEC[3:0]				00000000
R14	0x4014	ALC 3	NGTYP[1:0]		NGEN	NGTHR[4:0]					00000000
R15	0x4015	Serial Port 0	Reserved	SPSRS	LRMOD	BPOL	LRPOL	CHPF[1:0]		MS	00000000
R16	0x4016	Serial Port 1	BPF[2:0]			ADTDM	DATDM	MSBP	LRDEL[1:0]		00000000
R17	0x4017	Converter 0	Reserved	DAPAIR[1:0]		DAOSR	ADOSR	CONVSR[2:0]			00000000
R18	0x4018	Converter 1	Reserved						ADPAIR[1:0]		00000000
R19	0x4019	ADC control	Reserved	ADCPOL	HPF	DMPOL	DMSW	INSEL	ADCEN[1:0]		00010000
R20	0x401A	Left digital vol	LADVOL[7:0]								00000000
R21	0x401B	Right digital vol	RADVOL[7:0]								00000000
R22	0x401C	Play Mixer Left 0	Reserved	MX3RM	MX3LM	MX3AUXG[3:0]				MX3EN	00000000
R23	0x401D	Play Mixer Left 1	MX3G2[3:0]				MX3G1[3:0]				00000000
R24	0x401E	Play Mixer Right 0	Reserved	MX4RM	MX4LM	MX4AUXG[3:0]				MX4EN	00000000
R25	0x401F	Play Mixer Right 1	MX4G2[3:0]				MX4G1[3:0]				00000000
R26	0x4020	Play L/R mixer left	Reserved			MX5G4[1:0]		MX5G3[1:0]		MX5EN	00000000
R27	0x4021	Play L/R mixer right	Reserved			MX6G4[1:0]		MX6G3[1:0]		MX6EN	00000000
R28	0x4022	Play L/R mixer mono	Reserved					MX7[1:0]		MX7EN	00000000
R29	0x4023	Play HP left vol	LHPVOL[5:0]						LHPM	HPEN	00000010
R30	0x4024	Play HP right vol	RHPVOL[5:0]						RHPM	HPMODE	00000010
R31	0x4025	Line output left vol	LOUTVOL[5:0]						LOUTM	LOMODE	00000010
R32	0x4026	Line output right vol	ROUTVOL[5:0]						ROUTM	ROMODE	00000010
R33	0x4027	Play mono output	MONOVOL[5:0]						MONOM	MOMODE	00000010
R34	0x4028	Pop/click suppress	Reserved			POPMODE	POPLESS	ASLEW[1:0]		Reserved	00000000
R35	0x4029	Play power mgmt	HPBIAS[1:0]		DACBIAS[1:0]		PBIAS[1:0]		PREN	PLEN	00000000
R36	0x402A	DAC Control 0	DACMONO[1:0]		DACPOL	Reserved		DEMPH	DACEN[1:0]		00000000
R37	0x402B	DAC Control 1	LDAVOL[7:0]								00000000
R38	0x402C	DAC Control 2	RDAVOL[7:0]								00000000
R39	0x402D	Serial port pad	ADCSDP[1:0]		DACSDP[1:0]		LRCLKP[1:0]		BCLKP[1:0]		10101010
R40	0x402F	Control Port Pad 0	CDATP[1:0]		CLCHP[1:0]		SCLP[1:0]		SDAP[1:0]		10101010

CONTROL REGISTERS

Table 33. Register Map (Continued)

Reg	Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default
R41	0x4030	Control Port Pad 1	Reserved							SDASTR	00000000
R42	0x4031	Jack detect pin	Reserved		JDSTR	Reserved	JDP[1:0]		Reserved		00001000
R67	0x4036	Dejitter control	DEJIT[7:0]								00000011
R43	0x40C0	Cyclic redundancy check	CRC[31:24]								00000000
R44	0x40C1		CRC[23:16]								00000000
R45	0x40C2		CRC[15:8]								00000000
R46	0x40C3		CRC[7:0]								00000000
R47	0x40C4	CRC enable	Reserved							CRCEN	00000000
R48	0x40C6	GPIO0 pin control	Reserved				GPIO0[3:0]				00000000
R49	0x40C7	GPIO1 pin control	Reserved				GPIO1[3:0]				00000000
R50	0x40C8	GPIO2 pin control	Reserved				GPIO2[3:0]				00000000
R51	0x40C9	GPIO3 pin control	Reserved				GPIO3[3:0]				00000000
R52	0x40D0	Watchdog enable	Reserved							DOGEN	00000000
R53	0x40D1	Watchdog value	DOG[23:16]								00000000
R54	0x40D2		DOG[15:8]								00000000
R55	0x40D3		DOG[7:0]								00000000
R56	0x40D4	Watchdog error	Reserved							DOGER	00000000
R57	0x40EB	DSP sampling rate setting	Reserved				DSPSR[3:0]				00000001
R58	0x40F2	Serial input route control	Reserved				SINRT[3:0]				00000000
R59	0x40F3	Serial output route control	Reserved				SOUTRT[3:0]				00000000
R60	0x40F4	Serial data/GPIO pin configuration	Reserved				LRGP3	BGP2	SDOGP1	SDIGP0	00000000
R61	0x40F5	DSP enable	Reserved							DSPEN	00000000
R62	0x40F6	DSP run	Reserved							DSPRUN	00000000
R63	0x40F7	DSP slew modes	Reserved			MOSLW	ROSLW	LOSLW	RHPSLW	LHPSLW	00000000
R64	0x40F8	Serial port sampling rate	Reserved					SPSR[2:0]			00000000
R65	0x40F9	Clock Enable 0	Reserved	SLEWPD	ALCPD	DECPD	SOUTPD	INTPD	SINPD	SPPD	00000000
R66	0x40FA	Clock Enable 1	Reserved						CLK1	CLK0	00000000

CONTROL REGISTER DETAILS

All registers except for the PLL control register are 1-byte write and read registers.

R0: Clock Control, 16,384 (0x4000)

Table 34.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved				CLKSRC	INFREQ[1:0]		COREN

Table 35. Clock Control Register

Bits	Bit Name	Description
3	CLKSRC	Clock source select. 0 = direct from MCLK pin (default). 1 = PLL clock.
[2:1]	INFREQ[1:0]	Input clock frequency. Sets the core clock rate that generates the core clock. If the PLL is used, this value is automatically set to 1024 × f_S.

CONTROL REGISTERS

Table 35. Clock Control Register (Continued)

Bits	Bit Name	Description	
		Setting	**Input Clock Frequency**
		00	256 × f_S (default)
		01	512 × f_S
		10	768 × f_S
		11	1024 × f_S
0	COREN	Core clock enable. Only the R0 and R1 registers can be accessed when this bit is set to 0 (core clock disabled). 0 = core clock disabled (default). 1 = core clock enabled.	

R1: PLL Control, 16,386 (0x4002)

Table 36.

Byte	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	M[15:8]							
1	M[7:0]							
2	N[15:8]							
3	N[7:0]							
4	Reserved	R[3:0]				X[1:0]		Type
5	Reserved						Lock	PLLEN

Table 37. PLL Control Register

Byte	Bits	Bit Name	Description		
0	[7:0]	M[15:8]	PLL denominator MSB. This value is concatenated with M[7:0] to make up a 16-bit number.		
1	[7:0]	M[7:0]	PLL denominator LSB. This value is concatenated with M[15:8] to make up a 16-bit number.		
			M[15:8] (MSB)	**M[7:0] (LSB)**	**Value of M**
			00000000	00000000	0
			…	…	…
			00000000	11111101	253 (default)
			…	…	…
			11111111	11111111	65,535
2	[7:0]	N[15:8]	PLL numerator MSB. This value is concatenated with N[7:0] to make up a 16-bit number.		
3	[7:0]	N[7:0]	PLL numerator LSB. This value is concatenated with N[15:8] to make up a 16-bit number.		
			N[15:8] (MSB)	**N[7:0] (LSB)**	**Value of N**
			00000000	00000000	0
			…	…	…
			00000000	00001100	12 (default)
			…	…	…
			11111111	11111111	65,535
4	[6:3]	R[3:0]	PLL integer setting.		
			Setting	**Value of R**	
			0010	2 (default)	
			0011	3	
			0100	4	
			0101	5	
			0110	6	
			0111	7	
			1000	8	
4	[2:1]	X[1:0]	PLL input clock divider.		

CONTROL REGISTERS

Table 37. PLL Control Register (Continued)

Byte	Bits	Bit Name	Description	
			Setting	**Value of X**
			00	1 (default)
			01	2
			10	3
			11	4
4	0	Type	Type of PLL. When set to integer mode, the values of M and N are ignored. 0 = integer (default). 1 = fractional.	
5	1	Lock	PLL lock. This read-only bit is flagged when the PLL has finished locking. 0 = PLL unlocked (default). 1 = PLL locked.	
5	0	PLLEN	PLL enable. 0 = PLL disabled (default). 1 = PLL enabled.	

R2: Digital Microphone/Jack Detection Control, 16,392 (0x4008)

Table 38.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
JDDB[1:0]		JDFUNC[1:0]		Reserved			JDPOL

Table 39. Digital Microphone/Jack Detection Control Register

Bits	Bit Name	Description	
[7:6]	JDDB[1:0]	Jack detect debounce time.	
		Setting	**Debounce Time**
		00	5 ms (default)
		01	10 ms
		10	20 ms
		11	40 ms
[5:4]	JDFUNC[1:0]	JACKDET/MICIN pin function. Enables or disables the jack detect function or configures the pin for a digital microphone input.	
		Setting	**Pin Function**
		00	Jack detect off (default)
		01	Jack detect on
		10	Digital microphone input
		11	Reserved
0	JDPOL	Jack detect polarity. Detects high or low signal. 0 = detect high signal (default). 1 = detect low signal.	

R3: Record Power Management, 16,393 (0x4009)

This register manages the power consumption for the record path. In particular, the current distribution for the mixer boosts, ADCs, record path mixers, and PGAs can be set to one of four modes. These settings are normal operation, power saving mode, enhanced performance mode, and extreme power saving mode. Each of these modes draws current from a central bias. Enhanced performance mode offers the highest performance with the trade-off of higher power consumption.

Table 40.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	MXBIAS[1:0]		ADCBIAS[1:0]		RBIAS[1:0]		Reserved

CONTROL REGISTERS

Table 41. Record Power Management Register

Bits	Bit Name	Description	
[6:5]	MXBIAS[1:0]	Mixer amplifier bias boost. Sets the boost level for the bias current of the record path mixers. In some cases, the boost level enhances the THD + N performance.	
		Setting	**Boost Level**
		00	Normal operation (default)
		01	Boost Level 1
		10	Boost Level 2
		11	Boost Level 3
[4:3]	ADCBIAS[1:0]	ADC bias control. Sets the bias current for the ADCs based on the mode of operation selected.	
		Setting	**ADC Bias Control**
		00	Normal operation (default)
		01	Extreme power saving
		10	Enhanced performance
		11	Power saving
[2:1]	RBIAS[1:0]	Record path bias control. Sets the bias current for the PGAs and mixers in the record path.	
		Setting	**Record Path Bias Control**
		00	Normal operation (default)
		01	Reserved
		10	Enhanced performance
		11	Power saving

R4: Record Mixer Left (Mixer 1) Control 0, 16,394 (0x400A)

This register controls the gain of single-ended inputs for the left channel record path. The left channel record mixer is referred to as Mixer 1.

Table 42.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	LINPG[2:0]			LINNG[2:0]			MX1EN

Table 43. Record Mixer Left (Mixer 1) Control 0 Register

Bits	Bit Name	Description	
[6:4]	LINPG[2:0]	Gain for a left channel single-ended input from the LINP pin, input to Mixer 1.	
		Setting	**Gain**
		000	Mute (default)
		001	−12 dB
		010	−9 dB
		011	−6 dB
		100	−3 dB
		101	0 dB
		110	3 dB
		111	6 dB
[3:1]	LINNG[2:0]	Gain for a left channel single-ended input from the LINN pin, input to Mixer 1.	
		Setting	**Gain**
		000	Mute (default)
		001	−12 dB
		010	−9 dB
		011	−6 dB
		100	−3 dB
		101	0 dB

CONTROL REGISTERS

Table 43. Record Mixer Left (Mixer 1) Control 0 Register (Continued)

Bits	Bit Name	Description	
		110	3 dB
		111	6 dB
0	MX1EN	Left channel mixer enable in the record path. Referred to as Mixer 1. 0 = mixer disabled (default). 1 = mixer enabled.	

R5: Record Mixer Left (Mixer 1) Control 1, 16,395 (0x400B)

This register controls the gain boost of the left channel differential PGA input and the gain for the left channel auxiliary input in the record path. The left channel record mixer is referred to as Mixer 1.

Table 44.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved			LDBOOST[1:0]		MX1AUXG[2:0]		

Table 45. Record Mixer Left (Mixer 1) Control 1 Register

Bits	Bit Name	Description	
[4:3]	LDBOOST[1:0]	Left channel differential PGA input gain boost, input to Mixer 1. The left differential input uses the LINP (positive signal) and LINN (negative signal) pins.	
		Setting	**Gain Boost**
		00	Mute (default)
		01	0 dB
		10	20 dB
		11	Reserved
[2:0]	MX1AUXG[2:0]	Left single-ended auxiliary input gain from the LAUX pin in the record path, input to Mixer 1.	
		Setting	**Auxiliary Input Gain**
		000	Mute (default)
		001	−12 dB
		010	−9 dB
		011	−6 dB
		100	−3 dB
		101	0 dB
		110	3 dB
		111	6 dB

R6: Record Mixer Right (Mixer 2) Control 0, 16,396 (0x400C)

This register controls the gain of single-ended inputs for the right channel record path. The right channel record mixer is referred to as Mixer 2.

Table 46.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	RINPG[2:0]			RINNG[2:0]			MX2EN

Table 47. Record Mixer Right (Mixer 2) Control 0 Register

Bits	Bit Name	Description	
[6:4]	RINPG[2:0]	Gain for a right channel single-ended input from the RINP pin, input to Mixer 2.	
		Setting	**Gain**
		000	Mute (default)
		001	−12 dB

CONTROL REGISTERS

Table 47. Record Mixer Right (Mixer 2) Control 0 Register (Continued)

Bits	Bit Name	Description	
		010	−9 dB
		011	−6 dB
		100	−3 dB
		101	0 dB
		110	3 dB
		111	6 dB
[3:1]	RINNG[2:0]	Gain for a right channel single-ended input from the RINN pin, input to Mixer 2.	
		Setting	**Gain**
		000	Mute (default)
		001	−12 dB
		010	−9 dB
		011	−6 dB
		100	−3 dB
		101	0 dB
		110	3 dB
		111	6 dB
0	MX2EN	Right channel mixer enable in the record path. Referred to as Mixer 2. 0 = mixer disabled (default). 1 = mixer enabled.	

R7: Record Mixer Right (Mixer 2) Control 1, 16,397 (0x400D)

This register controls the gain boost of the right channel differential PGA input and the gain for the right channel auxiliary input in the record path. The right channel record mixer is referred to as Mixer 2.

Table 48.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved			RDBOOST[1:0]		MX2AUXG[2:0]		

Table 49. Record Mixer Right (Mixer 2) Control 1 Register

Bits	Bit Name	Description	
[4:3]	RDBOOST[1:0]	Right channel differential PGA input gain boost, input to Mixer 2. The right differential input uses the RINP (positive signal) and RINN (negative signal) pins.	
		Setting	**Gain Boost**
		00	Mute (default)
		01	0 dB
		10	20 dB
		11	Reserved
[2:0]	MX2AUXG[2:0]	Right single-ended auxiliary input gain from the RAUX pin in the record path, input to Mixer 2.	
		Setting	**Auxiliary Input Gain**
		000	Mute (default)
		001	−12 dB
		010	−9 dB
		011	−6 dB
		100	−3 dB
		101	0 dB
		110	3 dB
		111	6 dB

CONTROL REGISTERS

R8: Left Differential Input Volume Control, 16,398 (0x400E)

This register enables the differential path and sets the volume control for the left differential PGA input.

Table 50.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
LDVOL[5:0]						LDMUTE	LDEN

Table 51. Left Differential Input Volume Control Register

Bits	Bit Name	Description	
[7:2]	LDVOL[5:0]	Left channel differential PGA input volume control. The left differential input uses the LINP (positive signal) and LINN (negative signal) pins. Each step corresponds to a 0.75 dB increase in gain. See Table 149 for a complete list of the volume settings.	
		Setting	**Volume**
		000000	−12 dB (default)
		000001	−11.25 dB
		…	…
		010000	0 dB
		…	…
		111110	34.5 dB
		111111	35.25 dB
1	LDMUTE	Left differential input mute control. 0 = mute (default). 1 = unmute.	
0	LDEN	Left differential PGA enable. When enabled, the LINP and LINN pins are used as a full differential pair. When disabled, these two pins are configured as two single-ended inputs with the signals routed around the PGA. 0 = disabled (default). 1 = enabled.	

R9: Right Differential Input Volume Control, 16,399 (0x400F)

This register enables the differential path and sets the volume control for the right differential PGA input.

Table 52.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
RDVOL[5:0]						RDMUTE	RDEN

Table 53. Right Differential Input Volume Control Register

Bits	Bit Name	Description	
[7:2]	RDVOL[5:0]	Right channel differential PGA input volume control. The right differential input uses the RINP (positive signal) and RINN (negative signal) pins. Each step corresponds to a 0.75 dB increase in gain. See Table 149 for a complete list of the volume settings.	
		Setting	**Volume**
		000000	−12 dB (default)
		000001	−11.25 dB
		…	…
		010000	0 dB
		…	…
		111110	34.5 dB
		111111	35.25 dB
1	RDMUTE	Right differential input mute control. 0 = mute (default). 1 = unmute.	

CONTROL REGISTERS

Table 53. Right Differential Input Volume Control Register (Continued)

Bits	Bit Name	Description
0	RDEN	Right differential PGA enable. When enabled, the RINP and RINN pins are used as a full differential pair. When disabled, these two pins are configured as two single-ended inputs with the signals routed around the PGA. 0 = disabled (default). 1 = enabled.

R10: Record Microphone Bias Control, 16,400 (0x4010)

This register controls the MICBIAS pin settings for biasing electret type analog microphones.

Table 54.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved				MPERF	MBI	Reserved	MBIEN

Table 55. Record Microphone Bias Control Register

Bits	Bit Name	Description
3	MPERF	Microphone bias is enabled for high performance or normal operation. High performance operation sources more current to the microphone. 0 = normal operation (default). 1 = high performance.
2	MBI	Microphone voltage bias as a fraction of AVDD. 0 = 0.90 × AVDD (default). 1 = 0.65 × AVDD.
0	MBIEN	Enables the MICBIAS output. 0 = disabled (default). 1 = enabled.

R11: ALC Control 0, 16,401 (0x4011)

Table 56.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PGASLEW[1:0]		ALCMAX[2:0]			ALCSEL[2:0]		

Table 57. ALC Control 0 Register

Bits	Bit Name	Description	
[7:6]	PGASLEW[1:0]	PGA volume slew time when the ALC is off. The slew time is the period of time that a volume increase or decrease takes to ramp up or ramp down to the target volume set in Register R8 (left differential input volume control) and Register R9 (right differential input volume control).	
		Setting	**Slew Time**
		00	24 ms (default)
		01	48 ms
		10	96 ms
		11	Off
[5:3]	ALCMAX[2:0]	The maximum ALC gain sets a limit to the amount of gain that the ALC can provide to the input signal. This protects small signals from excessive amplification.	
		Setting	**Maximum ALC Gain**
		000	−12 dB (default)
		001	−6 dB
		010	0 dB
		011	6 dB
		100	12 dB

CONTROL REGISTERS

Table 57. ALC Control 0 Register (Continued)

Bits	Bit Name	Description	
		101	18 dB
		110	24 dB
		111	30 dB
[2:0]	ALCSEL[2:0]	ALC select. These bits set the channels that are controlled by the ALC. When set to right only, the ALC responds only to the right channel input and controls the gain of the right PGA amplifier only. When set to left only, the ALC responds only to the left channel input and controls the gain of the left PGA amplifier only. When set to stereo, the ALC responds to the greater of the left or right channel and controls the gain of both the left and right PGA amplifiers. DSP control allows the PGA gain to be set within the DSP or from external GPIO inputs. These bits must be off if manual control of the volume is desired.	
		Setting	**Channels**
		000	Off (default)
		001	Right only
		010	Left only
		011	Stereo
		100	DSP control
		101	Reserved
		110	Reserved
		111	Reserved

R12: ALC Control 1, 16,402 (0x4012)

Table 58.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
		ALCHOLD[3:0]				ALCTARG[3:0]	

Table 59. ALC Control 1 Register

Bits	Bit Name	Description	
[7:4]	ALCHOLD[3:0]	ALC hold time. The ALC hold time is the amount of time that the ALC waits after a decrease in input level before increasing the gain to achieve the target level. The recommended minimum setting is 21 ms (0011) to prevent distortion of low frequency signals. The hold time doubles with every 1-bit increase.	
		Setting	**Hold Time**
		0000	2.67 ms (default)
		0001	5.34 ms
		0010	10.68 ms
		0011	21.36 ms
		0100	42.72 ms
		0101	85.44 ms
		0110	170.88 ms
		0111	341.76 ms
		1000	683.52 ms
		1001	1.367 sec
		1010	2.7341 sec
		1011	5.4682 sec
		1100	10.936 sec
		1101	21.873 sec
		1110	43.745 sec
		1111	87.491 sec
[3:0]	ALCTARG[3:0]	ALC target. The ALC target sets the desired ADC input level. The PGA gain is adjusted by the ALC to reach this target level. The recommended target level is between −16 dB and −10 dB to accommodate transients without clipping the ADC.	
		Setting	**ALC Target**

CONTROL REGISTERS

Table 59. ALC Control 1 Register (Continued)

Bits	Bit Name	Description	
		0000	−28.5 dB (default)
		0001	−27 dB
		0010	−25.5 dB
		0011	−24 dB
		0100	−22.5 dB
		0101	−21 dB
		0110	−19.5 dB
		0111	−18 dB
		1000	−16.5 dB
		1001	−15 dB
		1010	−13.5 dB
		1011	−12 dB
		1100	−10.5 dB
		1101	−9 dB
		1110	−7.5 dB
		1111	−6 dB

R13: ALC Control 2, 16,403 (0x4013)

Table 60.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
ALCATCK[3:0]				ALCDEC[3:0]			

Table 61. ALC Control 2 Register

Bits	Bit Name	Description	
[7:4]	ALCATCK[3:0]	ALC attack time. The attack time sets how fast the ALC starts attenuating after an increase in input level above the target. A typical setting for music recording is 384 ms, and a typical setting for voice recording is 24 ms.	
		Setting	**Attack Time**
		0000	6 ms (default)
		0001	12 ms
		0010	24 ms
		0011	48 ms
		0100	96 ms
		0101	192 ms
		0110	384 ms
		0111	768 ms
		1000	1.54 sec
		1001	3.07 sec
		1010	6.14 sec
		1011	12.29 sec
		1100	24.58 sec
		1101	49.15 sec
		1110	98.30 sec
		1111	196.61 sec
[3:0]	ALCDEC[3:0]	ALC decay time. The decay time sets how fast the ALC increases the PGA gain after a decrease in input level below the target. A typical setting for music recording is 24.58 seconds, and a typical setting for voice recording is 1.54 seconds.	
		Setting	**Decay Time**
		0000	24 ms
		0001	48 ms

CONTROL REGISTERS

Table 61. ALC Control 2 Register (Continued)

Bits	Bit Name	Description	
		0010	96 ms
		0011	192 ms
		0100	384 ms
		0101	768 ms
		0110	1.54 sec
		0111	3.07 sec
		1000	6.14 sec
		1001	12.29 sec
		1010	24.58 sec
		1011	49.15 sec
		1100	98.30 sec
		1101	196.61 sec
		1110	393.22 sec
		1111	786.43 sec

R14: ALC Control 3, 16,404 (0x4014)

Table 62.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
NGTYP[1:0]		NGEN	NGTHR[4:0]				

Table 63. ALC Control 3 Register

Bits	Bit Name	Description	
[7:6]	NGTYP[1:0]	Noise gate type. When the input signal falls below the threshold for 250 ms, the noise gate can hold a constant PGA gain, mute the ADC output, fade the PGA gain to the minimum gain value, or fade then mute.	
		Setting	**Noise Gate**
		00	Hold PGA constant (default)
		01	Mute ADC output (digital mute)
		10	Fade to PGA minimum value (analog fade)
		11	Fade then mute (analog fade/digital mute)
5	NGEN	Noise gate enable. 0 = disabled (default). 1 = enabled.	
[4:0]	NGTHR[4:0]	Noise gate threshold. When the input signal falls below the threshold for 250 ms, the noise gate is activated. A 1 LSB increase corresponds to a −1.5 dB change. See Table 150 for a complete list of the threshold settings.	
		Setting	**Threshold**
		00000	−76.5 dB (default)
		00001	−75 dB
		…	…
		11110	−31.5 dB
		11111	−30 dB

R15: Serial Port Control 0, 16,405 (0x4015)

Table 64.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	SPSRS	LRMOD	BPOL	LRPOL	CHPF[1:0]		MS

CONTROL REGISTERS

Table 65. Serial Port Control 0 Register

Bits	Bit Name	Description
6	SPSRS	Serial port sampling rate source. 0 = converter rate set in Register R17 (default). 1 = DSP rate set in Register R57.
5	LRMOD	LRCLK mode sets the LRCLK for either a 50% duty cycle or a pulse. The pulse mode should be at least 1 BCLK wide. 0 = 50% duty cycle (default). 1 = pulse mode.
4	BPOL	BCLK polarity sets the BCLK edge that triggers a change in audio data. This can be set for the falling or rising edge of the BCLK. 0 = falling edge (default). 1 = rising edge.
3	LRPOL	LRCLK polarity sets the LRCLK edge that triggers the beginning of the left channel audio frame. This can be set for the falling or rising edge of the LRCLK. 0 = falling edge (default). 1 = rising edge.
[2:1]	CHPF[1:0]	Channels per frame sets the number of channels per LRCLK frame.
		Setting / **Channels per LRCLK Frame**
		00 / Stereo (default)
		01 / TDM 4
		10 / TDM 8
		11 / Reserved
0	MS	Serial data port bus mode. Both LRCLK and BCLK are main of the serial port when set in main mode and are serial port subordinate in subordinate mode. 0 = subordinate mode (default). 1 = main mode.

R16: Serial Port Control 1, 16,406 (0x4016)

Table 66.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
BPF[2:0]			ADTDM	DATDM	MSBP	LRDEL[1:0]	

Table 67. Serial Port Control 1 Register

Bits	Bit Name	Description
[7:5]	BPF[2:0]	Number of bit clock cycles per LRCLK audio frame.
		Setting / **Bit Clock Cycles**
		000 / 64 (default)
		001 / Reserved
		010 / 48
		011 / 128
		100 / 256
		101 / Reserved
		110 / Reserved
		111 / Reserved
4	ADTDM	ADC serial audio data channel position in TDM mode. 0 = left first (default). 1 = right first.
3	DATDM	DAC serial audio data channel position in TDM mode. 0 = left first (default). 1 = right first.
2	MSBP	MSB position in the LRCLK frame.

CONTROL REGISTERS

Table 67. Serial Port Control 1 Register (Continued)

Bits	Bit Name	Description	
		0 = MSB first (default). 1 = LSB first.	
[1:0]	LRDEL[1:0]	Data delay from LRCLK edge (in BCLK units).	
		Setting	**Delay (Bit Clock Cycles)**
		00	1 (default)
		01	0
		10	8
		11	16

R17: Converter Control 0, 16,407 (0x4017)

Table 68.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	DAPAIR[1:0]		DAOSR	ADOSR	CONVSR[2:0]		

Table 69. Converter Control 0 Register

Bits	Bit Name	Description		
[6:5]	DAPAIR[1:0]	On-chip DAC serial data selection in TDM 4 or TDM 8 mode.		
		Setting	**Pair**	
		00	First pair (default)	
		01	Second pair	
		10	Third pair	
		11	Fourth pair	
4	DAOSR	DAC oversampling ratio. This bit cannot be set for 64× when CONVSR[2:0] is set to 96 kHz. 0 = 128× (default). 1 = 64×.		
3	ADOSR	ADC oversampling ratio. This bit cannot be set for 64× when CONVSR[2:0] is set to 96 kHz. 0 = 128× (default). 1 = 64×.		
[2:0]	CONVSR[2:0]	Converter sampling rate. The ADCs and DACs operate at the sampling rate set in this register. The converter rate selected is a ratio of the base sampling rate, f_S. The base sampling rate is determined by the operating frequency of the core clock.		
		Setting	**Sampling Rate**	**Base Sampling Rate (f_S = 48 kHz)**
		000	f_S	48 kHz, base (default)
		001	$f_S/6$	8 kHz
		010	$f_S/4$	12 kHz
		011	$f_S/3$	16 kHz
		100	$f_S/2$	24 kHz
		101	$f_S/1.5$	32 kHz
		110	$f_S/0.5$	96 kHz
		111	Reserved	

R18: Converter Control 1, 16,408 (0x4018)

Table 70.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved						ADPAIR[1:0]	

CONTROL REGISTERS

Table 71. Converter Control 1 Register

Bits	Bit Name	Description	
[1:0]	ADPAIR[1:0]	On-chip ADC serial data selection in TDM 4 or TDM 8 mode.	
		Setting	Pair
		00	First pair (default)
		01	Second pair
		10	Third pair
		11	Fourth pair

R19: ADC Control, 16,409 (0x4019)

Table 72.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	ADCPOL	HPF	DMPOL	DMSW	INSEL	ADCEN[1:0]	

Table 73. ADC Control Register

Bits	Bit Name	Description	
6	ADCPOL	Invert input polarity. 0 = normal (default). 1 = inverted.	
5	HPF	ADC high-pass filter select. At 48 kHz, f_{3dB} = 2 Hz. 0 = off (default). 1 = on.	
4	DMPOL	Digital microphone data polarity swap. 0 = invert polarity. 1 = normal (default).	
3	DMSW	Digital microphone channel swap. Normal operation sends the left channel on the rising edge of the clock and the right channel on the falling edge of the clock. 0 = normal (default). 1 = swap left and right channels.	
2	INSEL	Digital microphone input select. When asserted, the on-chip ADCs are off, BCLK is main at 128 × f_S, and ADC_SDATA is expected to have left and right channels interleaved. 0 = digital microphone inputs off, ADCs enabled (default). 1 = digital microphone inputs enabled, ADCs off.	
[1:0]	ADCEN[1:0]	ADC enable.	
		Setting	**ADCs Enabled**
		00	Both off (default)
		01	Left on
		10	Right on
		11	Both on

R20: Left Input Digital Volume, 16,410 (0x401A)

Table 74.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
LADVOL[7:0]							

Table 75. Left Input Digital Volume Register

Bits	Bit Name	Description
[7:0]	LADVOL[7:0]	Controls the digital volume attenuation for left channel inputs from either the left ADC or the left digital microphone input. Each bit corresponds to a 0.375 dB step with slewing between settings. See Table 151 for a complete list of the volume settings.

CONTROL REGISTERS

Table 75. Left Input Digital Volume Register (Continued)

Bits	Bit Name	Description	
		Setting	**Volume Attenuation**
		00000000	0 dB (default)
		00000001	−0.375 dB
		00000010	−0.75 dB
		…	…
		11111110	−95.25 dB
		11111111	−95.625 dB

R21: Right Input Digital Volume, 16,411 (0x401B)

Table 76.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
RADVOL[7:0]							

Table 77. Right Input Digital Volume Register

Bits	Bit Name	Description	
[7:0]	RADVOL[7:0]	Controls the digital volume attenuation for right channel inputs from either the right ADC or the right digital microphone input. Each bit corresponds to a 0.375 dB step with slewing between settings. See Table 151 for a complete list of the volume settings.	
		Setting	**Volume Attenuation**
		00000000	0 dB (default)
		00000001	−0.375 dB
		00000010	−0.75 dB
		…	…
		11111110	−95.25 dB
		11111111	−95.625 dB

R22: Playback Mixer Left (Mixer 3) Control 0, 16,412 (0x401C)

Table 78.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	MX3RM	MX3LM	MX3AUXG[3:0]				MX3EN

Table 79. Playback Mixer Left (Mixer 3) Control 0 Register

Bits	Bit Name	Description	
6	MX3RM	Mixer input mute. Mutes the right DAC input to the left channel playback mixer (Mixer 3). 0 = muted (default). 1 = unmuted.	
5	MX3LM	Mixer input mute. Mutes the left DAC input to the left channel playback mixer (Mixer 3). 0 = muted (default). 1 = unmuted.	
[4:1]	MX3AUXG[3:0]	Mixer input gain. Controls the left channel auxiliary input gain to the left channel playback mixer (Mixer 3).	
		Setting	**Gain**
		0000	Mute (default)
		0001	−15 dB
		0010	−12 dB
		0011	−9 dB
		0100	−6 dB
		0101	−3 dB

CONTROL REGISTERS

Table 79. Playback Mixer Left (Mixer 3) Control 0 Register (Continued)

Bits	Bit Name	Description	
		0110	0 dB
		0111	3 dB
		1000	6 dB
0	MX3EN	Mixer 3 enable. 0 = disabled (default). 1 = enabled.	

R23: Playback Mixer Left (Mixer 3) Control 1, 16,413 (0x401D)

Table 80.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
MX3G2[3:0]				MX3G1[3:0]			

Table 81. Playback Mixer Left (Mixer 3) Control 1 Register

Bits	Bit Name	Description	
[7:4]	MX3G2[3:0]	Bypass gain control. The signal from the right channel record mixer (Mixer 2) bypasses the converters and gain can be applied before the left playback mixer (Mixer 3).	
		Setting	**Gain**
		0000	Mute (default)
		0001	−15 dB
		0010	−12 dB
		0011	−9 dB
		0100	−6 dB
		0101	−3 dB
		0110	0 dB
		0111	3 dB
		1000	6 dB
[3:0]	MX3G1[3:0]	Bypass gain control. The signal from the left channel record mixer (Mixer 1) bypasses the converters and gain can be applied before the left playback mixer (Mixer 3).	
		Setting	**Gain**
		0000	Mute (default)
		0001	−15 dB
		0010	−12 dB
		0011	−9 dB
		0100	−6 dB
		0101	−3 dB
		0110	0 dB
		0111	3 dB
		1000	6 dB

R24: Playback Mixer Right (Mixer 4) Control 0, 16,414 (0x401E)

Table 82.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	MX4RM	MX4LM	MX4AUXG[3:0]				MX4EN

CONTROL REGISTERS

Table 83. Playback Mixer Right (Mixer 4) Control 0 Register

Bits	Bit Name	Description	
6	MX4RM	Mixer input mute. Mutes the right DAC input to the right channel playback mixer (Mixer 4). 0 = muted (default). 1 = unmuted.	
5	MX4LM	Mixer input mute. Mutes the left DAC input to the right channel playback mixer (Mixer 4). 0 = muted (default). 1 = unmuted.	
[4:1]	MX4AUXG[3:0]	Mixer input gain. Controls the right channel auxiliary input gain to the right channel playback mixer (Mixer 4).	
		Setting	**Gain**
		0000	Mute (default)
		0001	−15 dB
		0010	−12 dB
		0011	−9 dB
		0100	−6 dB
		0101	−3 dB
		0110	0 dB
		0111	3 dB
		1000	6 dB
0	MX4EN	Mixer 4 enable. 0 = disabled (default). 1 = enabled.	

R25: Playback Mixer Right (Mixer 4) Control 1, 16,415 (0x401F)

Table 84.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
MX4G2[3:0]				MX4G1[3:0]			

Table 85. Playback Mixer Right (Mixer 4) Control 1 Register

Bits	Bit Name	Description	
[7:4]	MX4G2[3:0]	Bypass gain control. The signal from the right channel record mixer (Mixer 2) bypasses the converters and gain can be applied before the right playback mixer (Mixer 4).	
		Setting	**Gain**
		0000	Mute (default)
		0001	−15 dB
		0010	−12 dB
		0011	−9 dB
		0100	−6 dB
		0101	−3 dB
		0110	0 dB
		0111	3 dB
		1000	6 dB
[3:0]	MX4G1[3:0]	Bypass gain control. The signal from the left channel record mixer (Mixer 1) bypasses the converters and gain can be applied before the right playback mixer (Mixer 4).	
		Setting	**Gain**
		0000	Mute (default)
		0001	−15 dB
		0010	−12 dB
		0011	−9 dB

CONTROL REGISTERS

Table 85. Playback Mixer Right (Mixer 4) Control 1 Register (Continued)

Bits	Bit Name	Description	
		0100	−6 dB
		0101	−3 dB
		0110	0 dB
		0111	3 dB
		1000	6 dB

R26: Playback L/R Mixer Left (Mixer 5) Line Output Control, 16,416 (0x4020)

Table 86.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved			MX5G4[1:0]		MX5G3[1:0]		MX5EN

Table 87. Playback L/R Mixer Left (Mixer 5) Line Output Control Register

Bits	Bit Name	Description	
[4:3]	MX5G4[1:0]	Mixer input gain boost. The signal from the right channel playback mixer (Mixer 4) can be enabled and boosted in the playback L/R mixer left (Mixer 5).	
		Setting	**Gain Boost**
		00	Mute (default)
		01	0 dB output (−6 dB gain on each of the two inputs)
		10	6 dB output (0 dB gain on each of the two inputs)
		11	Reserved
[2:1]	MX5G3[1:0]	Mixer input gain boost. The signal from the left channel playback mixer (Mixer 3) can be enabled and boosted in the playback L/R mixer left (Mixer 5).	
		Setting	**Gain Boost**
		00	Mute (default)
		01	0 dB output (−6 dB gain on each of the two inputs)
		10	6 dB output (0 dB gain on each of the two inputs)
		11	Reserved
0	MX5EN	Mixer 5 enable. 0 = disabled (default). 1 = enabled.	

R27: Playback L/R Mixer Right (Mixer 6) Line Output Control, 16,417 (0x4021)

Table 88.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved			MX6G4[1:0]		MX6G3[1:0]		MX6EN

Table 89. Playback L/R Mixer Right (Mixer 6) Line Output Control Register

Bits	Bit Name	Description	
[4:3]	MX6G4[1:0]	Mixer input gain boost. The signal from the right channel playback mixer (Mixer 4) can be enabled and boosted in the playback L/R mixer right (Mixer 6).	
		Setting	**Gain Boost**
		00	Mute (default)
		01	0 dB output (−6 dB gain on each of the two inputs)
		10	6 dB output (0 dB gain on each of the two inputs)
		11	Reserved
[2:1]	MX6G3[1:0]	Mixer input gain boost. The signal from the left channel playback mixer (Mixer 3) can be enabled and boosted in the playback L/R mixer right (Mixer 6).	

CONTROL REGISTERS

Table 89. Playback L/R Mixer Right (Mixer 6) Line Output Control Register (Continued)

Bits	Bit Name	Description	
		Setting	**Gain Boost**
		00	Mute (default)
		01	0 dB output (−6 dB gain on each of the two inputs)
		10	6 dB output (0 dB gain on each of the two inputs)
		11	Reserved
0	MX6EN	Mixer 6 enable. 0 = disabled (default). 1 = enabled.	

R28: Playback L/R Mixer Mono Output (Mixer 7) Control, 16,418 (0x4022)

Table 90.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved					MX7[1:0]		MX7EN

Table 91. Playback L/R Mixer Mono Output (Mixer 7) Control Register

Bits	Bit Name	Description	
[2:1]	MX7[1:0]	L/R mono playback mixer (Mixer 7). Mixes the left and right playback mixers (Mixer 3 and Mixer 4) with either a 0 dB or 6 dB gain boost. Additionally, this mixer can operate as a common-mode output, which is used as the virtual ground in a capless headphone configuration.	
		Setting	**Gain Boost**
		00	Common-mode output (default)
		01	0 dB output (−6 dB gain on each of the two inputs)
		10	6 dB output (0 dB gain on each of the two inputs)
		11	Reserved
0	MX7EN	Mixer 7 enable. 0 = disabled (default). 1 = enabled.	

R29: Playback Headphone Left Volume Control, 16,419 (0x4023)

Table 92.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
LHPVOL[5:0]						LHPM	HPEN

Table 93. Playback Headphone Left Volume Control Register

Bits	Bit Name	Description	
[7:2]	LHPVOL[5:0]	Headphone volume control for left channel, LHP output. Each 1-bit step corresponds to a 1 dB increase in volume. See Table 152 for a complete list of the volume settings.	
		Setting	**Volume**
		000000	−57 dB (default)
		…	…
		111001	0 dB
		…	…
		111111	6 dB
1	LHPM	Headphone mute for left channel, LHP output (active low). 0 = mute. 1 = unmute (default).	
0	HPEN	Headphone volume control enable. Logical OR with the HPMODE bit in Register R30. If either the HPEN bit or the HPMODE bit is set to 1, the headphone output is enabled.	

CONTROL REGISTERS

Table 93. Playback Headphone Left Volume Control Register (Continued)

Bits	Bit Name	Description
		0 = disabled (default). 1 = enabled.

R30: Playback Headphone Right Volume Control, 16,420 (0x4024)

Table 94.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
RHPVOL[5:0]						RHPM	HPMODE

Table 95. Playback Headphone Right Volume Control Register

Bits	Bit Name	Description	
[7:2]	RHPVOL[5:0]	Headphone volume control for right channel, RHP output. Each 1-bit step corresponds to a 1 dB increase in volume. See Table 152 for a complete list of the volume settings.	
		Setting	**Volume**
		000000	−57 dB (default)
		…	…
		111001	0 dB
		…	…
		111111	6 dB
1	RHPM	Headphone mute for right channel, RHP output (active low). 0 = mute. 1 = unmute (default).	
0	HPMODE	RHP and LHP output mode. These pins can be configured for either line outputs or headphone outputs. Logical OR with the HPEN bit in Register R29. If either the HPMODE bit or the HPEN bit is set to 1, the headphone output is enabled. 0 = enable line output (default). 1 = enable headphone output.	

R31: Playback Line Output Left Volume Control, 16,421 (0x4025)

Table 96.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
LOUTVOL[5:0]						LOUTM	LOMODE

Table 97. Playback Line Output Left Volume Control Register

Bits	Bit Name	Description	
[7:2]	LOUTVOL[5:0]	Line output volume control for left channel, LOUTN and LOUTP outputs. Each 1-bit step corresponds to a 1 dB increase in volume. See Table 152 for a complete list of the volume settings.	
		Setting	**Volume**
		000000	−57 dB (default)
		…	…
		111001	0 dB
		…	…
		111111	6 dB
1	LOUTM	Line output mute for left channel, LOUTN and LOUTP outputs (active low). 0 = mute. 1 = unmute (default).	
0	LOMODE	Line output mode for left channel, LOUTN and LOUTP outputs. These pins can be configured for either line outputs or headphone outputs. To drive earpiece speakers, set this bit to 1 (headphone output). 0 = line output (default).	

CONTROL REGISTERS

Table 97. Playback Line Output Left Volume Control Register (Continued)

Bits	Bit Name	Description
		1 = headphone output.

R32: Playback Line Output Right Volume Control, 16,422 (0x4026)

Table 98.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
ROUTVOL[5:0]						ROUTM	ROMODE

Table 99. Playback Line Output Right Volume Control Register

Bits	Bit Name	Description	
[7:2]	ROUTVOL[5:0]	Line output volume control for right channel, ROUTN and ROUTP outputs. Each 1-bit step corresponds to a 1 dB increase in volume. See Table 152 for a complete list of the volume settings.	
		Setting	**Volume**
		000000	−57 dB (default)
		…	…
		111001	0 dB
		…	…
		111111	6 dB
1	ROUTM	Line output mute for right channel, ROUTN and ROUTP outputs (active low). 0 = mute. 1 = unmute (default).	
0	ROMODE	Line output mode for right channel, ROUTN and ROUTP outputs. These pins can be configured for either line outputs or headphone outputs. To drive earpiece speakers, set this bit to 1 (headphone output). 0 = line output (default). 1 = headphone output.	

R33: Playback Mono Output Control, 16,423 (0x4027)

Table 100.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
MONOVOL[5:0]						MONOM	MOMODE

Table 101. Playback Mono Output Control Register

Bits	Bit Name	Description	
[7:2]	MONOVOL[5:0]	Mono output volume control. Each 1-bit step corresponds to a 1 dB increase in volume. If MX7[1:0] in Register R28 is set for common-mode output, volume control is disabled. See Table 152 for a complete list of the volume settings.	
		Setting	**Volume**
		000000	−57 dB (default)
		…	…
		111001	0 dB
		…	…
		111111	6 dB
1	MONOM	Mono output mute (active low). 0 = mute. 1 = unmute (default).	
0	MOMODE	Headphone mode enable. If MX7[1:0] in Register R28 is set for common-mode output for a capless headphone configuration, this bit should be set to 1 (headphone output). 0 = line output (default). 1 = headphone output.	

CONTROL REGISTERS

R34: Playback Pop/Click Suppression, 16,424 (0x4028)

Table 102.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved			POPMODE	POPLESS	ASLEW[1:0]		Reserved

Table 103. Playback Pop/Click Suppression Register

Bits	Bit Name	Description	
4	POPMODE	Pop suppression circuit power saving mode. The pop suppression circuits charge faster in normal operation; however, after they are charged, they can be put into low power operation. 0 = normal (default). 1 = low power.	
3	POPLESS	Pop suppression disable. The pop suppression circuits are enabled by default. They can be disabled to save power; however, disabling the circuits increases the risk of pops and clicks. 0 = enabled (default). 1 = disabled.	
[2:1]	ASLEW[1:0]	Analog volume slew rate for playback volume controls.	
		Setting	**Slew Rate**
		00	21.25 ms (default)
		01	42.5 ms
		10	85 ms
		11	Off

R35: Playback Power Management, 16,425 (0x4029)

Table 104.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
HPBIAS[1:0]		DACBIAS[1:0]		PBIAS[1:0]		PREN	PLEN

Table 105. Playback Power Management Register

Bits	Bit Name	Description	
[7:6]	HPBIAS[1:0]	Headphone bias control.	
		Setting	**Headphone Bias Control**
		00	Normal operation (default)
		01	Extreme power saving
		10	Enhanced performance
		11	Power saving
[5:4]	DACBIAS[1:0]	DAC bias control.	
		Setting	**DAC Bias Control**
		00	Normal operation (default)
		01	Extreme power saving
		10	Enhanced performance
		11	Power saving
[3:2]	PBIAS[1:0]	Playback path channel bias control.	
		Setting	**Playback Path Bias Control**
		00	Normal operation (default)
		01	Reserved
		10	Enhanced performance
		11	Power saving
1	PREN	Playback right channel enable. 0 = disabled (default).	

CONTROL REGISTERS

Table 105. Playback Power Management Register (Continued)

Bits	Bit Name	Description
		1 = enabled.
0	PLEN	Playback left channel enable. 0 = disabled (default). 1 = enabled.

R36: DAC Control 0, 16,426 (0x402A)

Table 106.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
DACMONO[1:0]		DACPOL	Reserved		DEMPH	DACEN[1:0]	

Table 107. DAC Control 0 Register

Bits	Bit Name	Description	
[7:6]	DACMONO[1:0]	DAC mono mode. The DAC channels can be set to mono mode within the DAC and output on the left channel, the right channel, or both channels.	
		Setting	**Mono Mode**
		00	Stereo (default)
		01	Left channel in mono mode
		10	Right channel in mono mode
		11	Both channels in mono mode
5	DACPOL	Invert input polarity of the DACs. 0 = normal (default). 1 = inverted.	
2	DEMPH	DAC de-emphasis filter enable. The de-emphasis filter is designed for use with a sampling rate of 44.1 kHz only. 0 = disabled (default). 1 = enabled.	
[1:0]	DACEN[1:0]	DAC enable.	
		Setting	**DACs Enabled**
		00	Both off (default)
		01	Left on
		10	Right on
		11	Both on

R37: DAC Control 1, 16,427 (0x402B)

Table 108.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
LDAVOL[7:0]							

Table 109. DAC Control 1 Register

Bits	Bit Name	Description	
[7:0]	LDAVOL[7:0]	Controls the digital volume attenuation for left channel inputs from the left DAC. Each bit corresponds to a 0.375 dB step with slewing between settings. See Table 151 for a complete list of the volume settings.	
		Setting	**Volume Attenuation**
		00000000	0 dB (default)
		00000001	−0.375 dB
		00000010	−0.75 dB
		…	…
		11111110	−95.25 dB

CONTROL REGISTERS

Table 109. DAC Control 1 Register (Continued)

Bits	Bit Name	Description	
		11111111	−95.625 dB

R38: DAC Control 2, 16,428 (0x402C)

Table 110.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
RDAVOL[7:0]							

Table 111. DAC Control 2 Register

Bits	Bit Name	Description	
[7:0]	RDAVOL[7:0]	Controls the digital volume attenuation for right channel inputs from the right DAC. Each bit corresponds to a 0.375 dB step with slewing between settings. See Table 151 for a complete list of the volume settings.	
		Setting	**Volume Attenuation**
		00000000	0 dB (default)
		00000001	−0.375 dB
		00000010	−0.75 dB
		…	…
		11111110	−95.25 dB
		11111111	−95.625 dB

R39: Serial Port Pad Control, 16,429 (0x402D)

The optional pull-up/pull-down resistors are nominally 250 kΩ. When enabled, these pull-up/pull-down resistors set the serial port signals to a defined state when the signal source becomes three-state.

Table 112.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
ADCSDP[1:0]		DACSDP[1:0]		LRCLKP[1:0]		BCLKP[1:0]	

Table 113. Serial Port Pad Control Register

Bits	Bit Name	Description	
[7:6]	ADCSDP[1:0]	ADC_SDATA pad pull-up/pull-down configuration.	
		Setting	**Configuration**
		00	Pull-up
		01	Reserved
		10	None (default)
		11	Pull-down
[5:4]	DACSDP[1:0]	DAC_SDATA pad pull-up/pull-down configuration.	
		Setting	**Configuration**
		00	Pull-up
		01	Reserved
		10	None (default)
		11	Pull-down
[3:2]	LRCLKP[1:0]	LRCLK pad pull-up/pull-down configuration.	
		Setting	**Configuration**
		00	Pull-up
		01	Reserved
		10	None (default)

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Table 113. Serial Port Pad Control Register (Continued)

Bits	Bit Name	Description	
		11	Pull-down
[1:0]	BCLKP[1:0]	BCLK pad pull-up/pull-down configuration.	
		Setting	**Configuration**
		00	Pull-up
		01	Reserved
		10	None (default)
		11	Pull-down

R40: Control Port Pad Control 0, 16,431 (0x402F)

The optional pull-up/pull-down resistors are nominally 250 kΩ. When enabled, these pull-up/pull-down resistors set the control port signals to a defined state when the signal source becomes three-state.

Table 114.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CDATP[1:0]		CLCHP[1:0]		SCLP[1:0]		SDAP[1:0]	

Table 115. Control Port Pad Control 0 Register

Bits	Bit Name	Description	
[7:6]	CDATP[1:0]	CDATA pad pull-up/pull-down configuration.	
		Setting	**Configuration**
		00	Pull-up
		01	Reserved
		10	None (default)
		11	Pull-down
[5:4]	CLCHP[1:0]	$\overline{\text{CLATCH}}$ pad pull-up/pull-down configuration.	
		Setting	**Configuration**
		00	Pull-up
		01	Reserved
		10	None (default)
		11	Pull-down
[3:2]	SCLP[1:0]	SCL/CCLK pad pull-up/pull-down configuration.	
		Setting	**Configuration**
		00	Pull-up
		01	Reserved
		10	None (default)
		11	Pull-down
[1:0]	SDAP[1:0]	SDA/COUT pad pull-up/pull-down configuration.	
		Setting	**Configuration**
		00	Pull-up
		01	Reserved
		10	None (default)
		11	Pull-down

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R41: Control Port Pad Control 1, 16,432 (0x4030)

With IOVDD set to 3.3 V, the low and high drive strengths of the SDA/COUT pin are approximately 2.0 mA and 4.0 mA, respectively. With IOVDD set to 1.8 V, the low and high drive strengths are approximately 0.8 mA and 1.7 mA, respectively. The high drive strength mode may be useful for generating a stronger ACK pulse in I^2C mode, if needed.

Table 116.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved							SDASTR

Table 117. Control Port Pad Control 1 Register

Bits	Bit Name	Description
0	SDASTR	SDA/COUT pin drive strength. 0 = low (default). 1 = high.

R42: Jack Detect Pin Control, 16,433 (0x4031)

With IOVDD set to 3.3 V, the low and high drive strengths of the JACKDET/MICIN pin are approximately 2.0 mA and 4.0 mA, respectively. With IOVDD set to 1.8 V, the low and high drive strengths are approximately 0.8 mA and 1.7 mA, respectively. The optional pull-up/ pull-down resistors are nominally 250 kΩ. When enabled, these pull-up/pull-down resistors set the input signals to a defined state when the signal source becomes three-state.

Table 118.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved		JDSTR	Reserved	JDP[1:0]		Reserved	

Table 119. Jack Detect Pin Control Register

Bits	Bit Name	Description	
5	JDSTR	JACKDET/MICIN pin drive strength. 0 = low (default). 1 = high.	
[3:2]	JDP[1:0]	JACKDET/MICIN pad pull-up/pull-down configuration.	
		Setting	**Configuration**
		00	Pull-up
		01	Reserved
		10	None (default)
		11	Pull-down

R67: Dejitter Control, 16,438 (0x4036)

The dejitter control register allows the size of the dejitter window to be set, and also allows all dejitter circuits in the device to be activated or bypassed. Dejitter circuits protect against duplicate samples or skipped samples due to jitter from the serial ports in subordinate mode. Disabling and reenabling certain subsystems in the device—that is, the ADCs, serial ports, SigmaDSP core, and DACs—during operation can cause the associated dejitter circuits to fail. As a result, audio data fails to be output to the next subsystem in the device.

When the serial ports are in main mode, the dejitter circuit can be bypassed by setting the dejitter window to 0. When the serial ports are in subordinate mode, the dejitter circuit can be reinitialized prior to outputting audio from the device, guaranteeing that audio is output to the next subsystem in the device. Any time that audio must pass through the ADCs, serial port, sound engine/DSP core, or DACs, the dejitter circuit can be bypassed and reset by setting the dejitter window size to 0. In this way, the dejitter circuit can be immediately reactivated, without a wait period, by setting the dejitter window size to the default value of 3.

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Table 120.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
DEJIT[7:0]							

Table 121. Dejitter Control Register

Bits	Bit Name	Description	
[7:0]	DEJIT[7:0]	Dejitter window size.	
		Window Size	**Core Clock Cycles**
		00000000	0
		…	…
		00000011	3 (default)
		…	…
		00000101	5

R43 to R47: Cyclic Redundancy Check Registers, 16,576 to 16,580 (0x40C0 to 0x40C4)

The cyclic redundancy check (CRC) constantly checks the validity of the program RAM contents. SigmaStudio generates a 32-bit hash sum, which must be written to four consecutive read-only 8-bit register locations. CRC must then be enabled. Every 1024 frames (21 ms at 48 kHz), the IC generates its own 32-bit code and compares it to the one stored in these registers. If the codes do not match, a GPIO pin is set high (CRC flag). This output flag must be enabled using the output CRC error sticky setting in the GPIO pin control register (see Table 125). The 1-bit CRC error flag is reset when the CRCEN bit goes low. For example, a GPIO pin can be connected to an interrupt pin on an external microcontroller, which triggers a rewrite of the corrupted memory.

By default, CRC is disabled (the CRCEN bit is set to 0). To enable continuous CRC checking, the user can set the CRCEN bit to 1 after loading a program and sending the correct CRC, which is calculated by SigmaStudio. If an error occurs, it can be cleared by setting the CRCEN bit low, fixing the error (presumably by reloading the program), and then setting the CRCEN bit high again.

Table 122.

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x40C0	CRC[31:24]							
0x40C1	CRC[23:16]							
0x40C2	CRC[15:8]							
0x40C3	CRC[7:0]							
0x40C4	Reserved							CRCEN

Table 123. Cyclic Redundancy Check Registers

	Address			
Register	**Decimal**	**Hex**	**Bit Name**	**Description**
R43	16,576	0x40C0	CRC[31:24]	CRC hash sum, Bits[31:24] (read-only register)
R44	16,577	0x40C1	CRC[23:16]	CRC hash sum, Bits[23:16] (read-only register)
R45	16,578	0x40C2	CRC[15:8]	CRC hash sum, Bits[15:8] (read-only register)
R46	16,579	0x40C3	CRC[7:0]	CRC hash sum, Bits[7:0] (read-only register)
R47	16,580	0x40C4	CRCEN	CRC enable 0 = disabled (default) 1 = enabled

R48 to R51: GPIO Pin Control, 16,582 to 16,585 (0x40C6 to 0x40C9)

The GPIO pin control register sets the functionality of each GPIO pin as shown in Table 125. The GPIO functions use the same pins as the serial port and must be enabled in the serial data/GPIO pin configuration register (Address 0x40F4). When the GPIO pins are set to I^2C/SPI

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port control mode, the pins are set through writes to memory locations described in Table 32. The value of the optional internal pull-up is nominally 250 kΩ.

The output CRC error and output watchdog error settings are sticky, that is, once set, they remain set until the ADAU1761 is reset.

Table 124.

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x40C6	Reserved				GPIO0[3:0]			
0x40C7	Reserved				GPIO1[3:0]			
0x40C8	Reserved				GPIO2[3:0]			
0x40C9	Reserved				GPIO3[3:0]			

Table 125. GPIO Pin Functionality Bit Settings

GPIOx[3:0] Bits	GPIO Pin Function
0000	Input without debounce (default)
0001	Input with debounce (0.3 ms)
0010	Input with debounce (0.6 ms)
0011	Input with debounce (0.9 ms)
0100	Input with debounce (5 ms)
0101	Input with debounce (10 ms)
0110	Input with debounce (20 ms)
0111	Input with debounce (40 ms)
1000	Input controlled by I^2C/SPI port
1001	Output set by I^2C/SPI port, with pull-up
1010	Output set by I^2C/SPI port, no pull-up
1011	Output set by DSP core, with pull-up
1100	Output set by DSP core, no pull-up
1101	Reserved
1110	Output CRC error (sticky)
1111	Output watchdog error (sticky)

Table 126. GPIO Pin Control Registers

	Address			
Register	Decimal	Hex	Bit Name	Description
R48	16,582	0x40C6	GPIO0[3:0]	GPIO 0 pin function (see Table 125)
R49	16,583	0x40C7	GPIO1[3:0]	GPIO 1 pin function (see Table 125)
R50	16,584	0x40C8	GPIO2[3:0]	GPIO 2 pin function (see Table 125)
R51	16,585	0x40C9	GPIO3[3:0]	GPIO 3 pin function (see Table 125)

R52 to R56: Watchdog Registers, 16,592 to 16,596 (0x40D0 to 0x40D4)

A program counter watchdog is used when the core does block processing (which can span several samples). The watchdog flags an error if the program counter reaches a specific 24-bit value (ranging from 0x000000 to 0xFFFFFF) that is set in the register map. This value consists of three consecutive 8-bit register locations. The error flag sends a high signal to one of the GPIO pins. The watchdog function must be enabled by setting the DOGEN bit high in Register R52 (Address 0x40D0).

The watchdog error bit (DOGER) is the 1-bit watchdog error flag that can be sent to a GPIO pin, as described in Table 125. This error flag can connect, for example, to an interrupt pin on a microcontroller in the system. The flag is reset when the DOGEN bit goes low. This flag can also be read back over the control port from Register R56 (Address 0x40D4).

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Table 127.

Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x40D0	Reserved							DOGEN
0x40D1	DOG[23:16]							
0x40D2	DOG[15:8]							
0x40D3	DOG[7:0]							
0x40D4	Reserved							DOGER

Table 128. Watchdog Registers

Register	Address		Bit Name	Description			
	Decimal	Hex					
R52	16,592	0x40D0	DOGEN	Watchdog enable bit. 0 = disabled (default). 1 = enabled.			
R53	16,593	0x40D1	DOG[23:16]	Watchdog value, Bits[23:16] (MSB).			
R54	16,594	0x40D2	DOG[15:8]	Watchdog value, Bits[15:8].			
R55	16,595	0x40D3	DOG[7:0]	Watchdog value, Bits[7:0].			
				DOG[23:16]	**DOG[15:8]**	**DOG[7:0]**	**Hex Value**
				00000000 … 11111111	00000000 … 11111111	00000000 … 11111111	0x000000 (default) … 0xFFFFFF
R56	16,596	0x40D4	DOGER	Watchdog error (read-only bit). 0 = no error (default). 1 = error.			

R57: DSP Sampling Rate Setting, 16,619 (0x40EB)

Table 129.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved				DSPSR[3:0]			

Table 130. DSP Sampling Rate Setting Register

Bits	Bit Name	Description		
[3:0]	DSPSR[3:0]	SigmaDSP core sampling rate. The DSP sampling rate is a ratio of the base sampling rate, f_S. The base sampling rate is determined by the operating frequency of the core clock. For most applications, the SigmaDSP core sampling rate should equal the converter sampling rate (set using the CONVSR[2:0] bits in Register R17) and the serial port sampling rate (set using the SPSR[2:0] bits in Register R64).		
		Setting	**Sampling Rate**	**Base Sampling Rate (f_S = 48 kHz)**
		0000	f_S/0.5	96 kHz, base
		0001	f_S	48 kHz (default)
		0010	f_S/1.5	32 kHz
		0011	f_S/2	24 kHz
		0100	f_S/3	16 kHz
		0101	f_S/4	12 kHz
		0110	f_S/6	8 kHz
		0111	Serial input data rate	
		1000	Serial output data rate	
		1111	None	

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R58: Serial Input Route Control, 16,626 (0x40F2)

Table 131.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved				SINRT[3:0]			

Table 132. Serial Input Route Control Register

Bits	Bit Name	Description	
[3:0]	SINRT[3:0]	Serial data input routing. This register sets the input where the DACs receive serial data. This location can be from the DSP or from any TDM slot on the serial port.	
		Setting	**Routing**
		0000	DSP to DACs [L, R] (default)
		0001	Serial input [L0, R0] to DACs [L, R]
		0010	Reserved
		0011	Serial input [L1, R1] to DACs [L, R]
		0100	Reserved
		0101	Serial input [L2, R2] to DACs [L, R]
		0110	Reserved
		0111	Serial input [L3, R3] to DACs [L, R]
		1000	Reserved
		1001	Serial input [R0, L0] to DACs [L, R]
		1010	Reserved
		1011	Serial input [R1, L1] to DACs [L, R]
		1100	Reserved
		1101	Serial input [R2, L2] to DACs [L, R]
		1110	Reserved
		1111	Serial input [R3, L3] to DACs [L, R]

R59: Serial Output Route Control, 16,627 (0x40F3)

Table 133.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved				SOUTRT[3:0]			

Table 134. Serial Output Route Control Register

Bits	Bit Name	Description	
[3:0]	SOUTRT[3:0]	Serial data output routing. This register sets the output where the ADCs send serial data. This location can be to the DSP or to any TDM slot on the serial port.	
		Setting	**Routing**
		0000	ADCs [L, R] to DSP (default)
		0001	ADCs [L, R] to serial output [L0, R0]
		0010	Reserved
		0011	ADCs [L, R] to serial output [L1, R1]
		0100	Reserved
		0101	ADCs [L, R] to serial output [L2, R2]
		0110	Reserved
		0111	ADCs [L, R] to serial output [L3, R3]
		1000	Reserved
		1001	ADCs [L, R] to serial output [R0, L0]
		1010	Reserved
		1011	ADCs [L, R] to serial output [R1, L1]

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Table 134. Serial Output Route Control Register (Continued)

Bits	Bit Name	Description	
		1100	Reserved
		1101	ADCs [L, R] to serial output [R2, L2]
		1110	Reserved
		1111	ADCs [L, R] to serial output [R3, L3]

R60: Serial Data/GPIO Pin Configuration, 16,628 (0x40F4)

The serial data/GPIO pin configuration register controls the functionality of the serial data port pins. If the bits in this register are set to 1, these pins are configured as GPIO interfaces to the SigmaDSP. If these bits are set to 0, they are configured as serial data I/O port pins.

Table 135.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved				LRGP3	BGP2	SDOGP1	SDIGP0

*Table 136. Table SEQ Table * ARABIC 85 . Serial Data/GPIO Pin Configuration Register*

Bits	Bit Name	Description
3	LRGP3	LRCLK or GPIO3 pin configuration select. 0 = LRCLK enabled (default). 1 = GPIO3 enabled.
2	BGP2	BCLK or GPIO2 pin configuration select. 0 = BCLK enabled (default). 1 = GPIO2 enabled.
1	SDOGP1	ADC_SDATA or GPIO1 pin configuration select. 0 = ADC_SDATA enabled (default). 1 = GPIO1 enabled.
0	SDIGP0	DAC_SDATA or GPIO0 pin configuration select. 0 = DAC_SDATA enabled (default). 1 = GPIO0 enabled.

R61: DSP Enable, 16,629 (0x40F5)

Table 137.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved							DSPEN

Table 138. DSP Enable Register

Bits	Bit Name	Description
0	DSPEN	Enables the DSP. Set this bit before writing to the parameter RAM and before setting the DSPRUN bit in Register R62 (Address 0x40F6). 0 = DSP disabled (default). 1 = DSP enabled.

R62: DSP Run, 16,630 (0x40F6)

Table 139.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved							DSPRUN

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Table 140. DSP Run Register

Bits	Bit Name	Description
0	DSPRUN	Run the DSP. Set the DSPEN bit in Register R61 (Address 0x40F5) before setting this bit. 0 = DSP off (default). 1 = run the DSP.

R63: DSP Slew Modes, 16,631 (0x40F7)

The DSP slew modes register sets the slew source for each output. The slew source can be either the DSP (digital slew) or the codec (analog slew). When these bits are set to Logic 0, the codec provides volume slew according to the ASLEW[1:0] bits in Register R34 (playback pop/click suppression register, Address 0x4028). When these bits are set to Logic 1, the slew is provided and defined by the DSP program, disabling the codec volume slew.

Table 141.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved			MOSLW	ROSLW	LOSLW	RHPSLW	LHPSLW

Table 142. DSP Slew Modes Register

Bits	Bit Name	Description
4	MOSLW	Mono output slew generation. 0 = codec (default). 1 = DSP.
3	ROSLW	Line output right slew generation. 0 = codec (default). 1 = DSP.
2	LOSLW	Line output left slew generation. 0 = codec (default). 1 = DSP.
1	RHPSLW	Headphone right slew generation. 0 = codec (default). 1 = DSP.
0	LHPSLW	Headphone left slew generation. 0 = codec (default). 1 = DSP.

R64: Serial Port Sampling Rate, 16,632 (0x40F8)

Table 143.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved					SPSR[2:0]		

Table 144. Serial Port Sampling Rate Register

Bits	Bit Name	Description		
[2:0]	SPSR[2:0]	Serial port sampling rate. The serial port sampling rate is a ratio of the base sampling rate, f_S. The base sampling rate is determined by the operating frequency of the core clock. For most applications, the serial port sampling rate should equal the converter sampling rate (set using the CONVSR[2:0] bits in Register R17) and the DSP sampling rate (set using the DSPSR[3:0] bits in Register R57).		
		Setting	**Sampling Rate**	**Base Sampling Rate (f_S = 48 kHz)**
		000	f_S	48 kHz, base (default)
		001	$f_S/6$	8 kHz
		010	$f_S/4$	12 kHz
		011	$f_S/3$	16 kHz

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Table 144. Serial Port Sampling Rate Register (Continued)

Bits	Bit Name	Description		
		100	$f_S/2$	24 kHz
		101	$f_S/1.5$	32 kHz
		110	$f_S/0.5$	96 kHz
		111	Reserved	

R65: Clock Enable 0, 16,633 (0x40F9)

This register disables or enables the digital clock engine for different blocks within the ADAU1761. For maximum power saving, use this register to disable blocks that are not being used.

Table 145.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	SLEWPD	ALCPD	DECPD	SOUTPD	INTPD	SINPD	SPPD

Table 146. Clock Enable 0 Register

Bits	Bit Name	Description
6	SLEWPD	Codec slew digital clock engine enable. When powered down, the analog playback path volume controls are disabled and stay set to their current state. 0 = powered down (default). 1 = enabled.
5	ALCPD	ALC digital clock engine enable. 0 = powered down (default). 1 = enabled.
4	DECPD	Decimator resync (dejitter) digital clock engine enable. 0 = powered down (default). 1 = enabled.
3	SOUTPD	Serial routing outputs digital clock engine enable. 0 = powered down (default). 1 = enabled.
2	INTPD	Interpolator resync (dejitter) digital clock engine enable. 0 = powered down (default). 1 = enabled.
1	SINPD	Serial routing inputs digital clock engine enable. 0 = powered down (default). 1 = enabled.
0	SPPD	Serial port digital clock engine enable. 0 = powered down (default). 1 = enabled.

R66: Clock Enable 1, 16,634 (0x40FA)

This register enables Digital Clock Generator 0 and Digital Clock Generator 1. Digital Clock Generator 0 generates sample rates for the ADCs, DACs, and DSP. Digital Clock Generator 1 generates BCLK and LRCLK for the serial port when the part is in main mode. For maximum power saving, use this register to disable clocks that are not being used.

Table 147.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved						CLK1	CLK0

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Table 148. Clock Enable 1 Register

Bits	Bit Name	Description
1	CLK1	Digital Clock Generator 1. 0 = off (default). 1 = on.
0	CLK0	Digital Clock Generator 0. 0 = off (default). 1 = on.

Table 149. R8 and R9 Volume Settings

Binary Value	Volume Setting (dB)
000000	−12
000001	−11.25
000010	−10.5
000011	−9.75
000100	−9
000101	−8.25
000110	−7.5
000111	−6.75
001000	−6
001001	−5.25
001010	−4.5
001011	−3.75
001100	−3
001101	−2.25
001110	−1.5
001111	−0.75
010000	0
010001	0.75
010010	1.5
010011	2.25
010100	3
010101	3.75
010110	4.5
010111	5.25
011000	6
011001	6.75
011010	7.5
011011	8.25
011100	9
011101	9.75
011110	10.5
011111	11.25
100000	12
100001	12.75
100010	13.5
100011	14.25
100100	15
100101	15.75
100110	16.5

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Table 149. R8 and R9 Volume Settings (Continued)

Binary Value	Volume Setting (dB)
100111	17.25
101000	18
101001	18.75
101010	19.5
101011	20.25
101100	21
101101	21.75
101110	22.5
101111	23.25
110000	24
110001	24.75
110010	25.5
110011	26.25
110100	27
110101	27.75
110110	28.5
110111	29.25
111000	30
111001	30.75
111010	31.5
111011	32.25
111100	33
111101	33.75
111110	34.5
111111	35.25

Table 150. R14 Noise Gate Threshold

Binary Value	Noise Gate Threshold (dB)
00000	−76.5
00001	−75
00010	−73.5
00011	−72
00100	−70.5
00101	−69
00110	−67.5
00111	−66
01000	−64.5
01001	−63
01010	−61.5
01011	−60
01100	−58.5
01101	−57
01110	−55.5
01111	−54
10000	−52.5
10001	−51
10010	−49.5
10011	−48
10100	−46.5

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Table 150. R14 Noise Gate Threshold (Continued)

Binary Value	Noise Gate Threshold (dB)
10101	−45
10110	−43.5
10111	−42
11000	−40.5
11001	−39
11010	−37.5
11011	−36
11100	−34.5
11101	−33
11110	−31.5
11111	−30

Table 151. R20, R21, R37, and R38 Volume Settings

Binary Value	Volume Attenuation (dB)
00000000	0
00000001	−0.375
00000010	−0.75
00000011	−1.125
00000100	−1.5
00000101	−1.875
00000110	−2.25
00000111	−2.625
00001000	−3
00001001	−3.375
00001010	−3.75
00001011	−4.125
00001100	−4.5
00001101	−4.875
00001110	−5.25
00001111	−5.625
00010000	−6
00010001	−6.375
00010010	−6.75
00010011	−7.125
00010100	−7.5
00010101	−7.875
00010110	−8.25
00010111	−8.625
00011000	−9
00011001	−9.375
00011010	−9.75
00011011	−10.125
00011100	−10.5
00011101	−10.875
00011110	−11.25
00011111	−11.625
00100000	−12
00100001	−12.375
00100010	−12.75

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Table 151. R20, R21, R37, and R38 Volume Settings (Continued)

Binary Value	Volume Attenuation (dB)
00100011	−13.125
00100100	−13.5
00100101	−13.875
00100110	−14.25
00100111	−14.625
00101000	−15
00101001	−15.375
00101010	−15.75
00101011	−16.125
00101100	−16.5
00101101	−16.875
00101110	−17.25
00101111	−17.625
00110000	−18
00110001	−18.375
00110010	−18.75
00110011	−19.125
00110100	−19.5
00110101	−19.875
00110110	−20.25
00110111	−20.625
00111000	−21
00111001	−21.375
00111010	−21.75
00111011	−22.125
00111100	−22.5
00111101	−22.875
00111110	−23.25
00111111	−23.625
01000000	−24
01000001	−24.375
01000010	−24.75
01000011	−25.125
01000100	−25.5
01000101	−25.875
01000110	−26.25
01000111	−26.625
01001000	−27
01001001	−27.375
01001010	−27.75
01001011	−28.125
01001100	−28.5
01001101	−28.875
01001110	−29.25
01001111	−29.625
01010000	−30
01010001	−30.375
01010010	−30.75
01010011	−31.125

CONTROL REGISTERS

Table 151. R20, R21, R37, and R38 Volume Settings (Continued)

Binary Value	Volume Attenuation (dB)
01010100	−31.5
01010101	−31.875
01010110	−32.25
01010111	−32.625
01011000	−33
01011001	−33.375
01011010	−33.75
01011011	−34.125
01011100	−34.5
01011101	−34.875
01011110	−35.25
01011111	−35.625
01100000	−36
01100001	−36.375
01100010	−36.75
01100011	−37.125
01100100	−37.5
01100101	−37.875
01100110	−38.25
01100111	−38.625
01101000	−39
01101001	−39.375
01101010	−39.75
01101011	−40.125
01101100	−40.5
01101101	−40.875
01101110	−41.25
01101111	−41.625
01110000	−42
01110001	−42.375
01110010	−42.75
01110011	−43.125
01110100	−43.5
01110101	−43.875
01110110	−44.25
01110111	−44.625
01111000	−45
01111001	−45.375
01111010	−45.75
01111011	−46.125
01111100	−46.5
01111101	−46.875
01111110	−47.25
01111111	−47.625
10000000	−48
10000001	−48.375
10000010	−48.75
10000011	−49.125
10000100	−49.5

CONTROL REGISTERS

Table 151. R20, R21, R37, and R38 Volume Settings (Continued)

Binary Value	Volume Attenuation (dB)
10000101	−49.875
10000110	−50.25
10000111	−50.625
10001000	−51
10001001	−51.375
10001010	−51.75
10001011	−52.125
10001100	−52.5
10001101	−52.875
10001110	−53.25
10001111	−53.625
10010000	−54
10010001	−54.375
10010010	−54.75
10010011	−55.125
10010100	−55.5
10010101	−55.875
10010110	−56.25
10010111	−56.625
10011000	−57
10011001	−57.375
10011010	−57.75
10011011	−58.125
10011100	−58.5
10011101	−58.875
10011110	−59.25
10011111	−59.625
10100000	−60
10100001	−60.375
10100010	−60.75
10100011	−61.125
10100100	−61.5
10100101	−61.875
10100110	−62.25
10100111	−62.625
10101000	−63
10101001	−63.375
10101010	−63.75
10101011	−64.125
10101100	−64.5
10101101	−64.875
10101110	−65.25
10101111	−65.625
10110000	−66
10110001	−66.375
10110010	−66.75
10110011	−67.125
10110100	−67.5
10110101	−67.875

CONTROL REGISTERS

Table 151. R20, R21, R37, and R38 Volume Settings (Continued)

Binary Value	Volume Attenuation (dB)
10110110	−68.25
10110111	−68.625
10111000	−69
10111001	−69.375
10111010	−69.75
10111011	−70.125
10111100	−70.5
10111101	−70.875
10111110	−71.25
10111111	−71.625
11000000	−72
11000001	−72.375
11000010	−72.75
11000011	−73.125
11000100	−73.5
11000101	−73.875
11000110	−74.25
11000111	−74.625
11001000	−75
11001001	−75.375
11001010	−75.75
11001011	−76.125
11001100	−76.5
11001101	−76.875
11001110	−77.25
11001111	−77.625
11010000	−78
11010001	−78.375
11010010	−78.75
11010011	−79.125
11010100	−79.5
11010101	−79.875
11010110	−80.25
11010111	−80.625
11011000	−81
11011001	−81.375
11011010	−81.75
11011011	−82.125
11011100	−82.5
11011101	−82.875
11011110	−83.25
11011111	−83.625
11100000	−84
11100001	−84.375
11100010	−84.75
11100011	−85.125
11100100	−85.5
11100101	−85.875
11100110	−86.25

CONTROL REGISTERS

Table 151. R20, R21, R37, and R38 Volume Settings (Continued)

Binary Value	Volume Attenuation (dB)
11100111	−86.625
11101000	−87
11101001	−87.375
11101010	−87.75
11101011	−88.125
11101100	−88.5
11101101	−88.875
11101110	−89.25
11101111	−89.625
11110000	−90
11110001	−90.375
11110010	−90.75
11110011	−91.125
11110100	−91.5
11110101	−91.875
11110110	−92.25
11110111	−92.625
11111000	−93
11111001	−93.375
11111010	−93.75
11111011	−94.125
11111100	−94.5
11111101	−94.875
11111110	−95.25
11111111	−95.625

Table 152. R29 through R33 Volume Settings

Binary Value	Volume Setting (dB)
000000	−57
000001	−56
000010	−55
000011	−54
000100	−53
000101	−52
000110	−51
000111	−50
001000	−49
001001	−48
001010	−47
001011	−46
001100	−45
001101	−44
001110	−43
001111	−42
010000	−41
010001	−40
010010	−39
010011	−38
010100	−37

CONTROL REGISTERS

Table 152. R29 through R33 Volume Settings (Continued)

Binary Value	Volume Setting (dB)
010101	−36
010110	−35
010111	−34
011000	−33
011001	−32
011010	−31
011011	−30
011100	−29
011101	−28
011110	−27
011111	−26
100000	−25
100001	−24
100010	−23
100011	−22
100100	−21
100101	−20
100110	−19
100111	−18
101000	−17
101001	−16
101010	−15
101011	−14
101100	−13
101101	−12
101110	−11
101111	−10
110000	−9
110001	−8
110010	−7
110011	−6
110100	−5
110101	−4
110110	−3
110111	−2
111000	−1
111001	0
111010	1
111011	2
111100	3
111101	4
111110	5
111111	6

OUTLINE DIMENSIONS

Package Drawing (Option)	Package Type	Package Description
CP-32-4	LFCSP	32-Lead Lead Frame Chip Scale Package

For the latest package outline information and land patterns (footprints), go to Package Index.

ORDERING GUIDE

Model[1]	Temperature Range	Package Description	Packing Quantity	Package Option
ADAU1761BCPZ	−40°C to +85°C	32-Lead Lead Frame Chip Scale Package [LFCSP]	Tray, 490	CP-32-4
ADAU1761BCPZ-R7	−40°C to +85°C	32-Lead Lead Frame Chip Scale Package [LFCSP]	Reel, 1500	CP-32-4
ADAU1761BCPZ-RL	−40°C to +85°C	32-Lead Lead Frame Chip Scale Package [LFCSP]	Reel, 5000	CP-32-4

[1] Z = RoHS Compliant Part.

EVALUATION BOARDS

Model[1]	Description
EVAL-ADAU1761Z	Evaluation Board

[1] Z = RoHS Compliant Part.

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