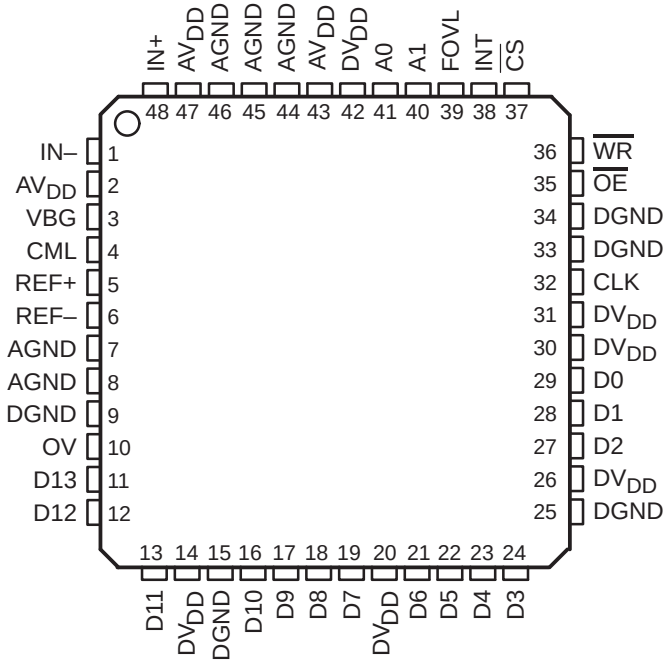
## features

- 14-Bit Resolution
- 1 MSPS and 3 MSPS Speed Grades Available
- On-Chip FIFO For Optimized Data Transfer
- Differential Nonlinearity (DNL) ±0.6 LSB Typ
- Integral Nonlinearity (INL) ±1.5 LSB Typ
- Internal Reference
- Differential Inputs
- Programmable Gain Amplifier
- μP Compatible Parallel Interface
- Timing Compatible With TI 6000 DSP Family
- 3.3-V Single Supply
- Power-Down Mode
- Monolithic CMOS Design

## applications

- xDSL Front Ends
- Communication
- Industrial Control
- Instrumentation
- Automotive

**PFB PACKAGE**
**(TOP VIEW)**




NC – No internal connection

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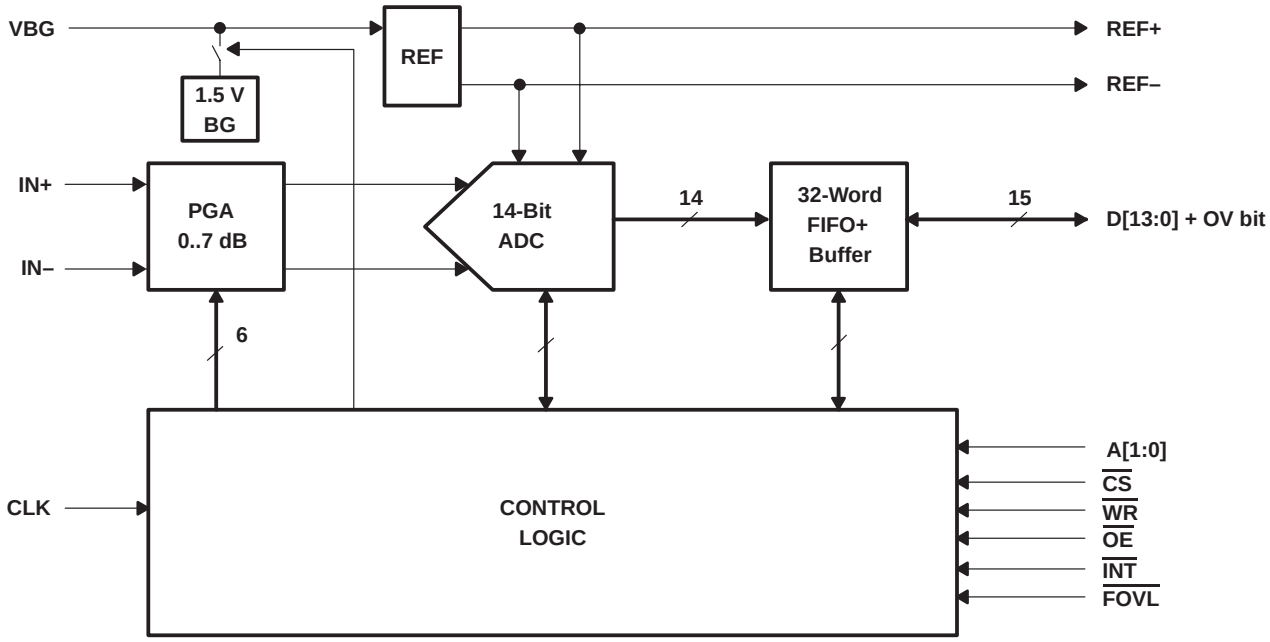
## description

The THS14F01 and THS14F03 are 14-bit, 1 MSPS/ 3 MSPS, single supply analog-to-digital converters with a FIFO, internal reference, differential inputs, programmable input gain, and an on-chip sample and hold amplifier.

Implemented with a CMOS process, the device has outstanding price/performance and power/speed ratios. The THS14F01 and THS14F03 are designed for use with 3.3-V systems, and with a high-speed µP compatible parallel interface, making them the first choice for solutions based on high-performance DSPs like the TI TMS320C6000 series.

The THS14F01 and THS14F03 are available in a TQFP-48 package in standard commercial and industrial temperature ranges.

## functional block diagram




**AVAILABLE OPTIONS**

| $T_A$ | PACKAGED DEVICE |
|---|---|
| | TQFP (PFB) |
| 0°C to 70°C | THS14F01CPFB, THS14F03CPFB |
| –40°C to 85°C | THS14F01IPFB, THS14F03IPFB |

## Terminal Functions

| TERMINAL NAME | TERMINAL NO. | I/O | DESCRIPTION |
|---|---|---|---|
| A[1:0] | 40, 41 | I | Address input |
| AGND | 7,8, 44, 45, 46 | P | Analog ground |
| $AV_{DD}$ | 2, 43, 47 | P | Analog power supply |
| CLK | 32 | I | Clock input |
| CML | 4 | | Reference midpoint. This pin requires a 0.1-μF capacitor to AGND. |
| $\overline{CS}$ | 37 | I | Chip select input. Active low |
| DGND | 9, 15, 25, 33, 34 | P | Digital ground |
| $DV_{DD}$ | 14, 20, 26, 30, 31, 42 | P | Digital power supply |
| D[13:0] | 11, 12, 13, 16, 17, 18, 19, 21, 22, 23, 24, 27, 28, 29 | I/O | Data inputs/outputs |
| FOVL | 39 | O | FIFO Overflow. Asserted when FIFO is full. Programmable polarity |
| IN+ | 48 | I | Positive differential analog input |
| IN– | 1 | I | Negative differential analog input |
| INT | 38 | O | Interrupt output. Asserted when FIFO trigger level is reached. Programmable polarity |
| $\overline{OE}$ | 35 | I | Output enable. Active low |
| OV | 10 | O | Out of range output |
| REF+ | 5 | O | Positive reference output. This pin requires a 0.1-μF capacitor to AGND. |
| REF– | 6 | O | Negative reference output. This pin requires a 0.1-μF capacitor to AGND. |
| VBG | 3 | I | Reference input. This pin requires a 1-μF capacitor to AGND. |
| $\overline{WR}$ | 36 | I | Write signal. Active low |

### absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply voltage, ($AV_{DD}$ to AGND) . . . . . 4V
Supply voltage, ($DV_{DD}$ to DGND) . . . . . 4V
Reference input voltage range, VBG . . . . . – 0.3 V to $AV_{DD}$ + 0.3 V
Analog input voltage range . . . . . – 0.3 V to $AV_{DD}$ + 0.3 V
Digital input voltage range . . . . . – 0.3 V to $DV_{DD}$ + 0.3 V
Operating free-air temperature range, $T_A$: C suffix . . . . . 0°C to 70°C
I suffix . . . . . –40°C to 85°C
Storage temperature range, $T_{stg}$ . . . . . –65°C to 150°C
Lead temperature 1.6 mm (1/16 inch) from case for 10 seconds . . . . . 260°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## recommended operating conditions

| | | MIN | NOM | MAX | UNIT |
|---|---|---|---|---|---|
| Supply voltage, $AV_{DD}$, $DV_{DD}$ | | 3 | 3.3 | 3.6 | V |
| High level digital input, $V_{IH}$ | | 2 | 3.3 | | V |
| Low level digital input, $V_{IL}$ | | | 0 | 0.8 | V |
| Load capacitance, $C_L$ | | | 5 | 15 | pF |
| Clock frequency, $f_{CLK}$ | THS14F01 | 0.1 | 1 | 1 | MHz |
| | THS14F03 | 0.1 | 3 | 3 | MHz |
| Clock duty cycle | | 40% | 50% | 60% | |
| Operating free-air temperature | C suffix | 0 | 25 | 70 | °C |
| | I suffix | –40 | 25 | 85 | |

## electrical characteristics over recommended operating conditions

| PARAMETER | | | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|---|---|
| **Power Supply** | | | | | | | |
| $I_{DDA}$ | Analog supply current | | | | 81 | 90 | mA |
| $I_{DDD}$ | Digital supply current | | | | 5 | 10 | mA |
| | Power | | | | 270 | 360 | mW |
| | Power down current | | | | 20 | | µA |
| **DC Characteristics**† | | | | | | | |
| | Resolution | | | | 14 | | Bits |
| DNL | Differential nonlinearity | | | | ±0.6 | ±1 | LSB |
| INL | Integral nonlinearity | THS14F01 | Best fit | | ±1.5 | ±2.5 | LSB |
| | | THS14F03 | | | ±1.5 | ±2.5 | |
| | Offset error | | IN+ = IN–, PGA = 0 dB | | | 0.3 | %FSR |
| | Gain error | | PGA = 0 dB | | | 1 | %FSR |
| **AC Characteristics**† | | | | | | | |
| ENOB | Effective number of bits | | | 11.2 | 11.5 | | Bits |
| THD | Total harmonic distortion | THS14F01/3 | $f_i$ = 100 kHz | | –81 | | dB |
| | | THS14F03 | $f_i$ = 1 MHz | | –78 | | |
| SNR | Signal-to-noise ratio | THS14F01/3 | $f_i$ = 100 kHz | | 72 | | dB |
| | | THS14F03 | $f_i$ = 1 MHz | 70 | 72 | | |
| SINAD | Signal-to-noise ratio + distortion | THS14F01/3 | $f_i$ = 100 kHz | | 70 | | dB |
| | | THS14F03 | $f_i$ = 1 MHz | 69 | 70 | | |
| SFDR | Spurious free dynamic range | THS14F01/3 | $f_i$ = 100 kHz | | 80 | | dB |
| | | THS14F03 | $f_i$ = 1 MHz | 73 | 80 | | |
| | Analog input bandwidth | | | | 140 | | MHz |

† FIFO trigger level = 10 samples. Performance is ensured with the output enable signal ($\overline{OE}$) being low during no more than one rising clock edge on CLK.

## electrical characteristics (continued)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|---|
| **Reference Voltage** | | | | | | |
| VBG | Bandgap voltage, internal mode | | 1.425 | 1.5 | 1.575 | V |
| | Input impedance | | | 40 | | kΩ |
| | Positive reference voltage, REF+ | | | 2.5 | | V |
| | Negative reference voltage, REF– | | | 0.5 | | V |
| | Reference difference, ΔREF, REF+ – REF– | | | 2 | | V |
| | Accuracy, internal reference | | | 5% | | |
| | Temperature coefficient | | | 40 | | ppm/°C |
| | Voltage coefficient | | | 200 | | ppm/V |
| **Analog Inputs** | | | | | | |
| | Positive analog input, IN+ | | 0 | | $AV_{DD}$ | V |
| | Negative analog input, IN– | | 0 | | $AV_{DD}$ | V |
| | Analog input voltage difference | ΔAin = IN+ – IN–, $V_{ref}$ = REF+ – REF– | $-V_{ref}$ | | $V_{ref}$ | V |
| | Input impedance | | | 25 | | kΩ |
| | PGA range | | 0 | | 7 | dB |
| | PGA step size | | | 1 | | dB |
| | PGA gain error | | | | ±0.25 | dB |
| **Digital Inputs** | | | | | | |
| $V_{IH}$ | High-level digital input | | 2 | | | V |
| $V_{IL}$ | Low-level digital input | | | | 0.8 | V |
| | Input capacitance | | | 5 | | pF |
| | Input current | | | | ±1 | μA |
| **Digital Outputs** | | | | | | |
| $V_{OH}$ | High-level digital output | $I_{OH}$ = 50 μA | 2.6 | | | V |
| $V_{OL}$ | Low-level digital output | $I_{OL}$ = 50 μA | | | 0.4 | V |
| $I_{OZ}$ | Output current, high impedance | | | | ±10 | μA |
| **Clock Timing (CS low)** | | | | | | |
| $f_{CLK}$ | Clock frequency | THS14F01 | 0.1 | 1 | 1 | MHz |
| | | THS14F03 | 0.1 | 3 | 3 | MHz |
| $t_d$ | Output delay time | | | | 25 | ns |
| | Latency | | | 9.5 | | Cycles |

## PARAMETER MEASUREMENT INFORMATION

### sample timing

The THS14F01/3 core is based on a pipeline architecture with a latency of 9.5 samples. The conversion results are stored in the FIFO 9.5 clock cycles after the input signal was sampled.

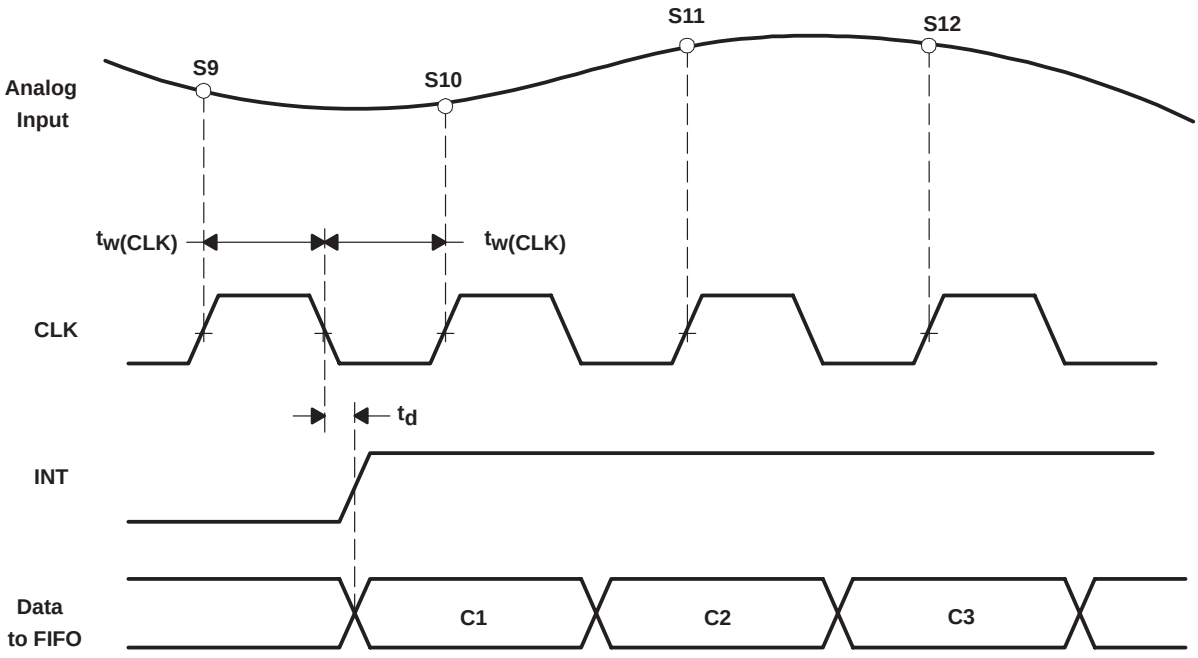



**Figure 1. Sample Timing**

INT goes active if the programmed FIFO level is reached. INT is either low or high active depending on the polarity bit (IP) within the control word. This signal is set synchronously to the CLK signal. It is reset by a read access to the FIFO once the number of samples in the FIFO is below the programmed threshold level.

## PARAMETER MEASUREMENT INFORMATION

The parallel interface of the THS14F01/3 ADC features 3-state buffers making it possible to directly connect it to a data bus. The output buffers are enabled by driving the OE input low.

Besides the sample results, it is also possible to read back the values of the control register, the PGA register, and the control register. Which register is read is determined by the address inputs A[1,0]. The ADC results are available at address 0.

The timing of the control signals is described in the following sections.

The FIFO can be disabled by setting FC to 0 (FIFO reset, default at power on). This makes it possible to access the device synchronously.

In this case the data is updated on every clock cycle.

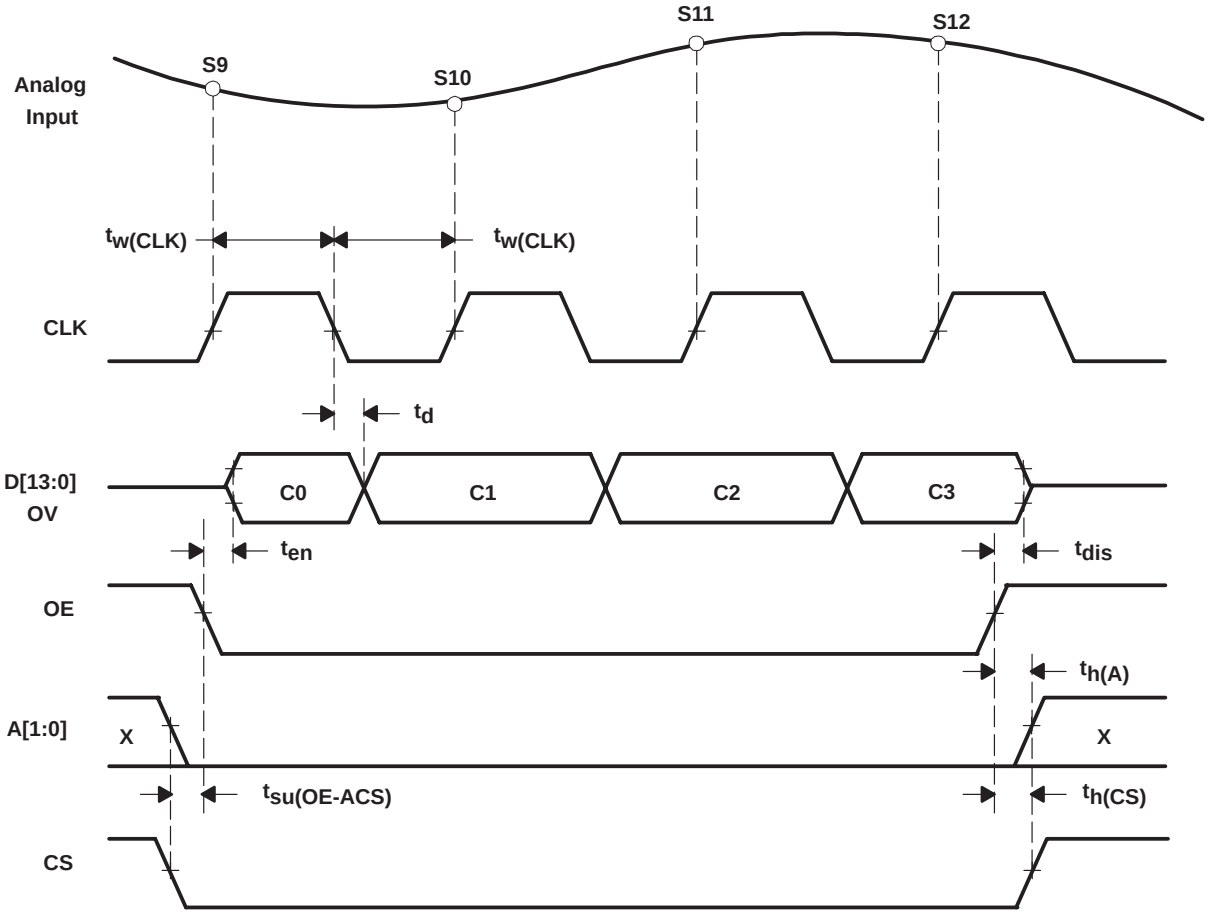



**Figure 2. Sample Timing**

## PARAMETER MEASUREMENT INFORMATION

### read timing (15-pF load)

| PARAMETER | | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| $t_{su(OE–ACS)}$ | Address and chip select setup time | 4 | | | ns |
| $t_{en}$ | Output enable | | | 15 | ns |
| $t_{dis}$ | Output disable | | | 10 | ns |
| $t_{h(A)}$ | Address hold time | 1 | | 15 | ns |
| $t_{h(CS)}$ | Chip select hold time | 0 | | | ns |

NOTE: All timing parameters refer to a 50% level.

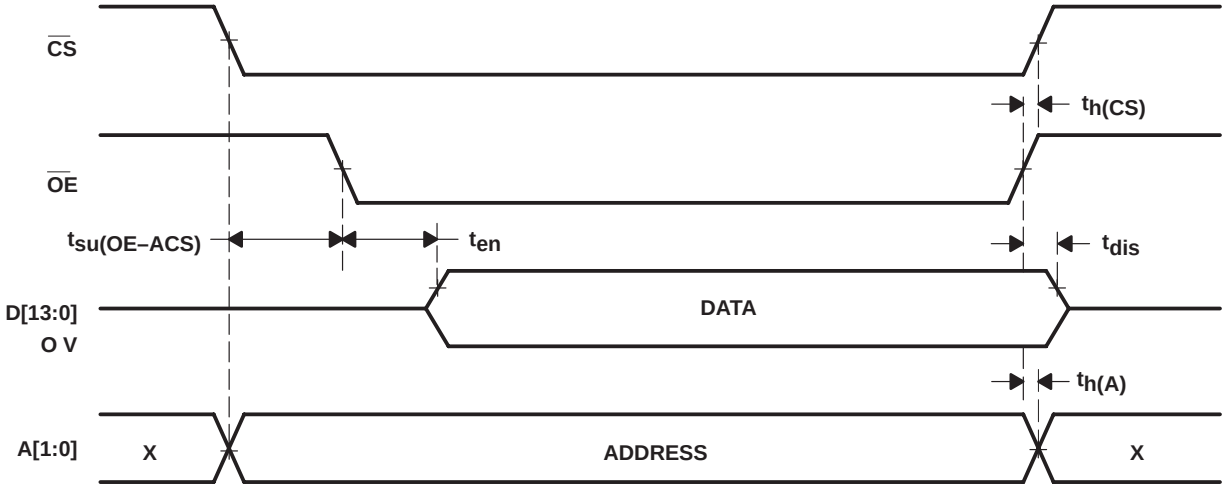



**Figure 3. Read Timing**

## PARAMETER MEASUREMENT INFORMATION

### write timing (15-pF load)

| PARAMETER | | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| $t_{su(WE–CS)}$ | Chip select setup time | 4 | | | ns |
| $t_{su(DA)}$ | Data and address setup time | 29 | | | ns |
| $t_{h(DA)}$ | Data and address hold time | 0 | | | ns |
| $t_{h(CS)}$ | Chip select hold time | 0 | | | ns |
| $t_{wH(WE)}$ | Write pulse duration high | 15 | | | ns |

NOTE: All timing parameters refer to a 50% level.

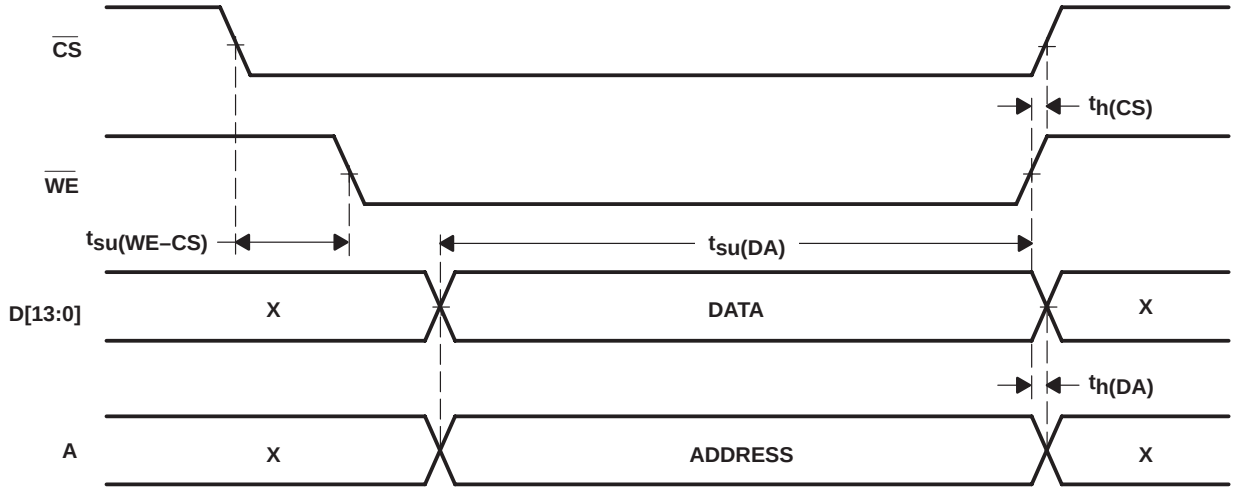



**Figure 4. Write Timing**

## TYPICAL CHARACTERISTICS

**POWER vs FREQUENCY**

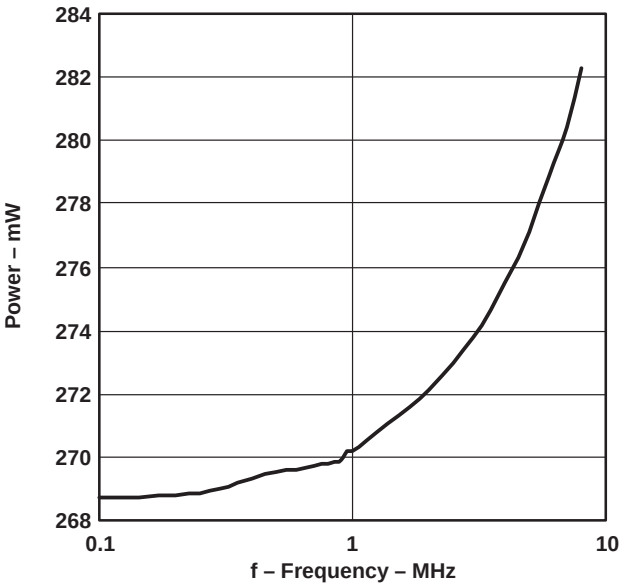



Figure 5

**SUPPLY CURRENT vs TIME**

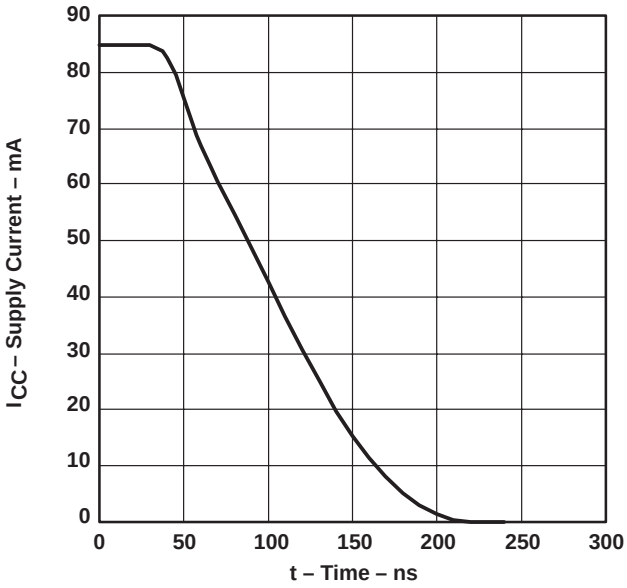



Figure 6

**FAST FOURIER TRANSFORM**

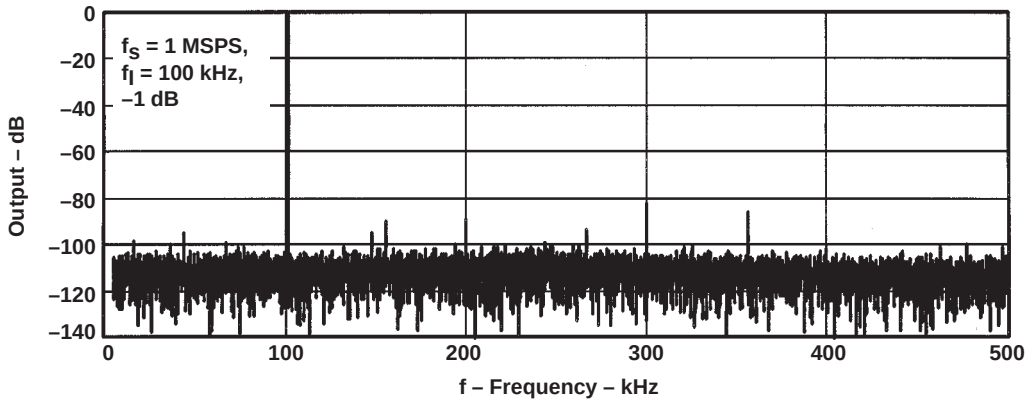



Figure 7

## TYPICAL CHARACTERISTICS

FAST FOURIER TRANSFORM

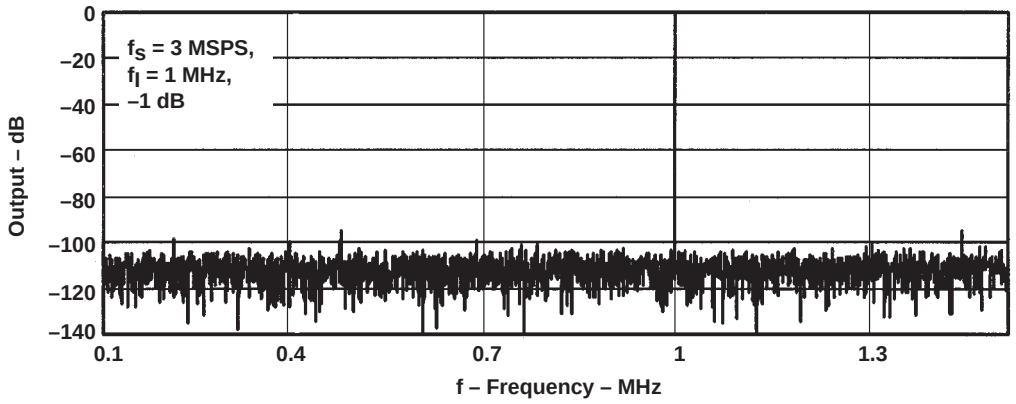



**Figure 8**

INTEGRAL NONLINEARITY

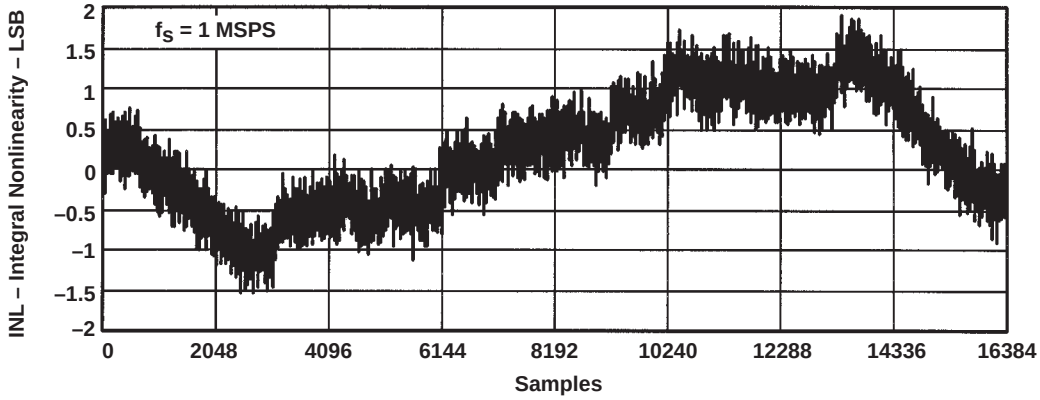



**Figure 9**

INTEGRAL NONLINEARITY

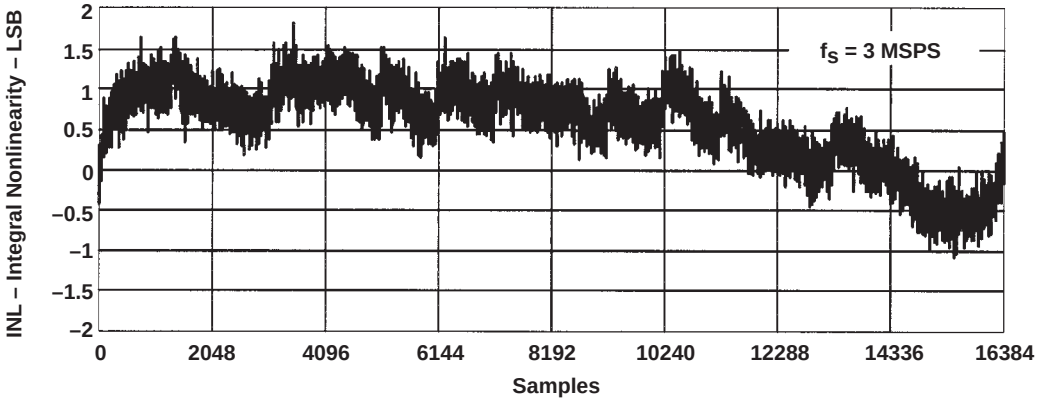



**Figure 10**

## TYPICAL CHARACTERISTICS

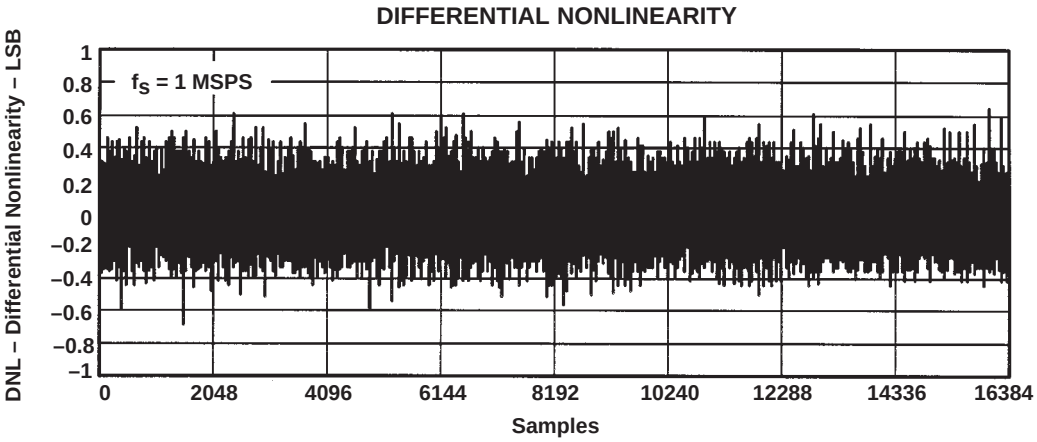



**Figure 11**

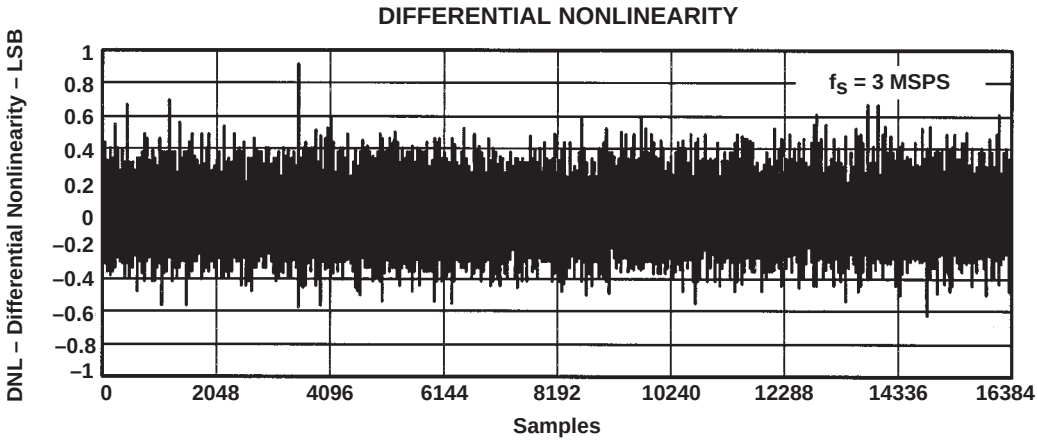



**Figure 12**

## TYPICAL CHARACTERISTICS

**TOTAL HARMONIC DISTORTION vs FREQUENCY**

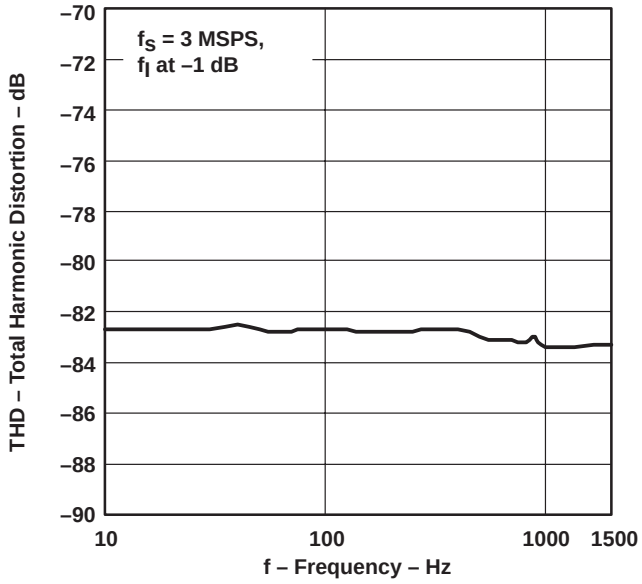



**Figure 13**

**SIGNAL-TO-NOISE RATIO vs FREQUENCY**

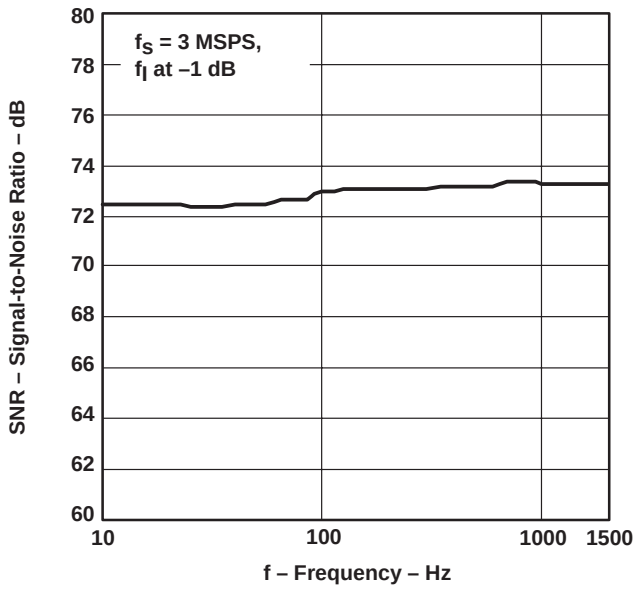



**Figure 14**

## PRINCIPLES OF OPERATION

### registers

The device contains several registers. The A register is selected by the values of bits A1 and A0:

| A1 | A0 | Register |
|---|---|---|
| 0 | 0 | Conversion result |
| 0 | 1 | PGA |
| 1 | 0 | Offset |
| 1 | 1 | Control |

Tables 1 and 2 describe how to read the conversion results and how to configure the data converter. The default values (were applicable) show the state after a power-on reset.

**Table 1. Conversion Result Register, Address 0, Read**

| BIT | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| Function | MSB | ... | … | … | … | … | … | … | … | … | … | … | … | LSB |

The output can be configured for two's complement or straight binary format (see D11/control register).

The output code is given by:

| 2s complement: | | Straight binary: | |
|---|---|---|---|
| –8192 | at ΔIN = –ΔREF | 0 | at ΔIN = –ΔREF |
| 0 | at ΔIN = 0 | 8192 | at ΔIN = 0 |
| 8191 | ΔIN = –ΔREF – 1 LSB | 16383 | at ΔIN = –ΔREF – 1 LSB |

$$1\ \text{LSB} = \frac{2\Delta\text{REF}}{16384}$$

**Table 2. PGA Gain Register, Address 1, Read/Write**

| BIT | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| Function | X | X | X | X | X | X | X | X | X | X | X | G2 | G1 | G0 |
| Default | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

The PGA gain is determined by writing to G2–0.

Gain (dB) = 1dB × G2–0. max = 7dB. The range of G2–0 is 0 to 7.

**Table 3. Offset Register, Address 2, Read/Write**

| BIT | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| Function | X | X | X | X | X | X | MSB | … | … | … | … | … | … | LSB |
| Default | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

The offset correction range is from –128 to 127 LSB. This value is added to the conversion results from the ADC.

## PRINCIPLES OF OPERATION

**Table 4. Control Register, Address 3, Read**

| BIT | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| Function | PWD | REF | FOR | TM2 | TM1 | TM0 | OFF | IP | FP | FC | F3 | F2 | F1 | F0 |

**Table 5. Control Register, Address 3, Write**

| BIT | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|---|---|---|---|---|---|---|---|---|---|---|---|---|---|---|
| Function | PWD | REF | FOR | TM2 | TM1 | TM0 | OFF | IP | FP | FC | F3 | F2 | F1 | F0 |
| Default | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| | | | |
|---|---|---|---|
| PWD: | Power down | 0 = normal operation | 1 = power down |
| REF: | Reference select | 0 = internal reference | 1 = external reference |
| FOR: | Output format | 0 = straight binary | 1 = 2s complement |
| TM2–0: | Test mode | 000 = normal operation | |
| | | 001 = both inputs = REF– | |
| | | 010 = IN+ at $V_{ref}/2$, IN– at REF– | |
| | | 011 = IN+ at REF+, IN– at REF– | |
| | | 100 = normal operation | |
| | | 101 = both inputs = REF+ | |
| | | 110 = IN+ at REF–, IN– at $V_{ref}/2$ | |
| | | 111 = IN+ at REF–, IN– at REF+ | |
| OFF: | Offset correction | 0 = enable | 1 = disable |
| IP: | INT polarity | 0 = low active | 1 = high active |
| FP: | FIFO FOVL polarity | 0 = low active | 1 = high active |
| FC: | FIFO control | 0 = disable FIFO | 1 = enable FIFO |
| F3–0: | FIFO threshold | Sets the FIFO threshold for the INT signal in steps of 2 ranging from 0 to 30 | |

## APPLICATION INFORMATION

### FIFO description

The FIFO is based on a circular buffer (see Figure 15, in this example the FIFO is 16 words long). The buffer is accessed using two pointers, one for the ADC writing to the FIFO, one for the processor (DSP) reading from the buffer. Both pointers move in a clockwise direction. If the distance between the ADC write pointer and the DSP read pointer is greater or equal a programmable threshold, the INT signal is asserted. If this INT signal is connected to an external interrupt pin of the processor, it is possible to read out the stored values in the FIFO at once during the interrupt service routine. If the ADC write pointer reaches the position of the DSP read pointer, an overflow occurs. In this case, the overflow bit in the ADC register is set and the FOVL is asserted.

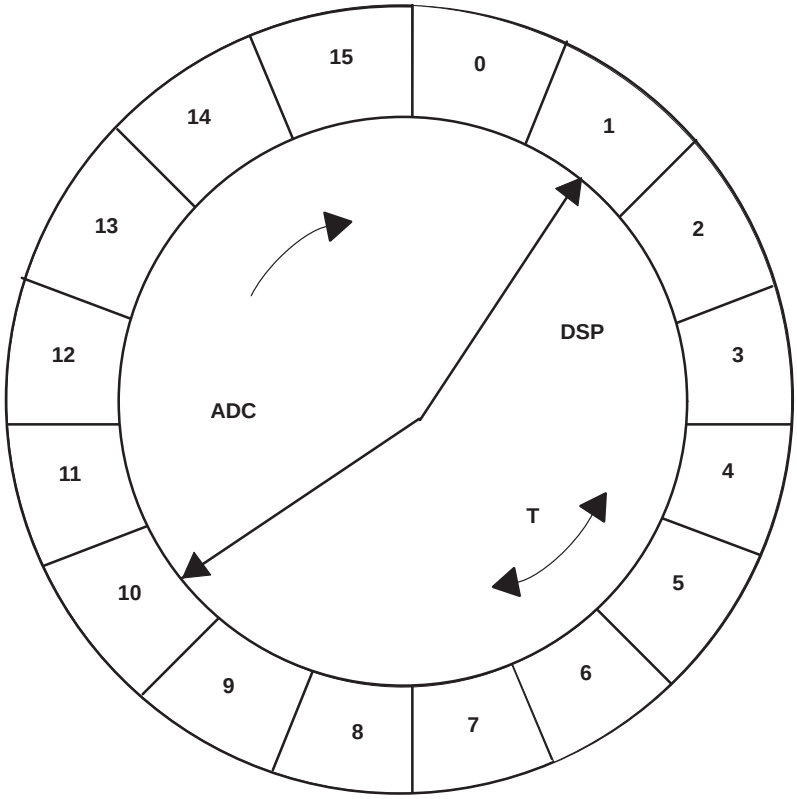



**Figure 15. Circular Buffer**

## APPLICATION INFORMATION

### DMA transfer and FIFO

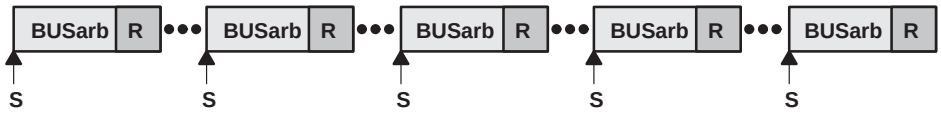
The FIFO makes it possible to use the available interface bandwidth of the host processor more efficiently. The following is a description based on the TMS320C6201 DSP from TI.

The TMS320C6201 memory interface has a limited bandwidth, for example 200MWPS at a clock rate of 200 MHz. The THS14F04x interface is asynchronous with a maximum speed of 300MWPS, which is approximately 7 clock cycles.

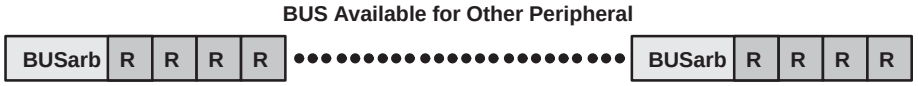
If the DSP uses the DMA controller to read data from the DSP, the following conditions exist:

- DMA bus arbitration: 16 clock cycles
- THS14F0x read access: 7 clock cycles

If, for example, 10 samples need to be read from the ADC without the FIFO, the memory interface will be allocated for (10 + 7) $\times$ 16 = 272 clock cycles in total.




With a FIFO programmed to a 10 sample threshold, the memory interface will be allocated for 16 + 7 $\times$ 10 = 86 clock cycles in total.




### driving the analog input

The THS14F01/3 ADCs have a fully differential input. A differential input is advantageous with respect to SNR, SFDR, and THD performance because the signal peak-to-peak level is 50% of a comparable single-ended input.

There are three basic input configurations:

- Fully differential
- Transformer coupled single-ended to differential
- Single-ended

## APPLICATION INFORMATION

### fully differential configuration

In this configuration, the ADC converts the difference (ΔIN) of the two input signals on IN+ and IN–.

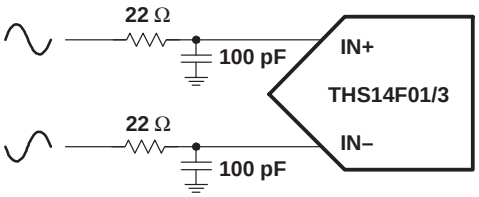



**Figure 16. Differential Input**

The resistors and capacitors on the inputs decouple the driving source output from the ADC input and also serve as first order low pass filters to attenuate out of band noise.

The input range on both inputs is 0 V to $AV_{DD}$. The full-scale value is determined by the voltage reference. The positive full-scale output is reached, if ΔIN equals ΔREF, the negative full-scale output is reached, if ΔIN equals –ΔREF.

| ΔIN [V] | OUTPUT |
|---|---|
| –ΔREF | – full scale |
| 0 | 0 |
| ΔREF | + full scale |

### transformer coupled single-ended to differential configuration

If the application requires the best SNR, SFDR, and THD performance, the input should be transformer coupled.

The signal amplitude on both inputs of the ADC is one half as high as in a single-ended configuration thus increasing the ADC ac performance.

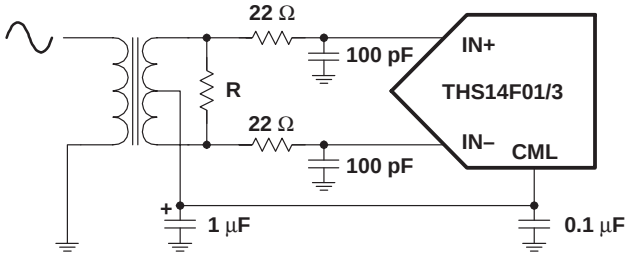



**Figure 17. Transformer Coupled**

| IN [$V_{PEAK}$] | OUTPUT [PEAK] |
|---|---|
| –ΔREF | – full scale† |
| 0 | 0 |
| ΔREF | + full scale† |

† n = 1 (winding ratio)

The resistor R of the transformer coupled input configuration must be set to match the signal source impedance $R = n^2 Rs$, where Rs is the source impedance and n is the transformer winding ratio.

## APPLICATION INFORMATION

### single-ended configuration

In this configuration, the input signal is level shifted by ΔREF/2.

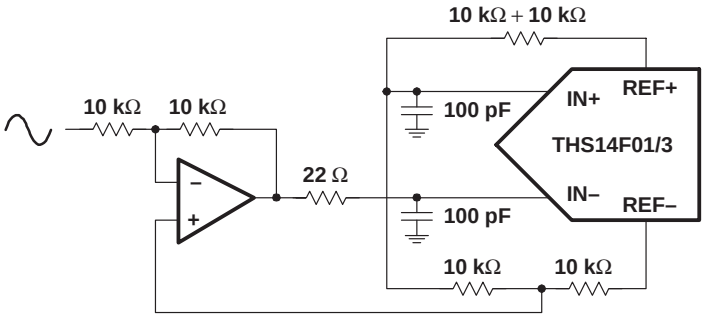



**Figure 18. Single-Ended With Level Shift**

The following table shows the input voltages for negative full-scale output, zero output, and positive full-scale output:

| ΔIN+ [V] | OUTPUT |
|---|---|
| –ΔREF | – full scale |
| 0 | 0 |
| ΔREF | + full scale |

Note that the resistors of the op-amp and the op-amp all introduce gain and offset errors. Those errors can be trimmed by varying the values of the resistors.

Because of the added offset, the op-amp does not necessarily operate in the best region of its transfer curve (best linearity around zero) and therefore may introduce unacceptable distortion. For ac signals, an alternative is described in the following section.

## APPLICATION INFORMATION

### ac-coupled single-ended configuration

If the application does not require the signal bandwidth to include dc, the level shift shown in Figure 4 is not necessary.

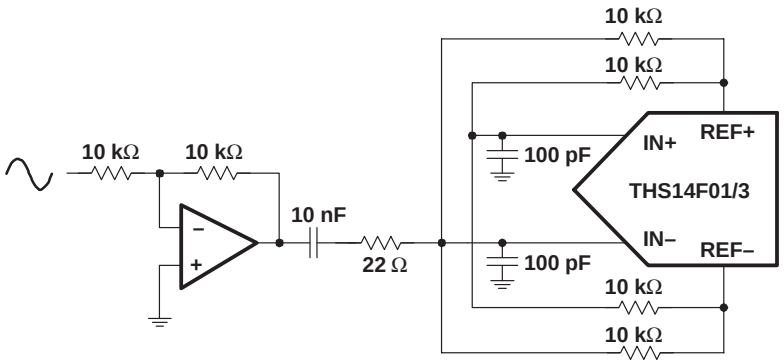



**Figure 19. Single-Ended With Level Shift**

Because the signal swing on the op-amp is centered around ground, it is more likely that the signal stays within the linear region of the op-amp transfer function, thus increasing the overall ac performance.

| IN [$V_{PEAK}$] | OUTPUT [$_{PEAK}$] |
|---|---|
| –ΔREF | – full scale |
| 0 | 0 |
| ΔREF | + full scale |

Compared to the transformer-coupled configuration, the swing on IN– is twice as big, which can decrease the ac performance (SNR, SFD, and THD).

## APPLICATION INFORMATION

### internal/external reference operation

The THS14F01/3 ADC can either be operated using the built-in band gap reference or using an external precision reference in case very high dc accuracy is needed.

The REF+ and REF+ outputs are given by:

$$\text{REF} + = \text{VBG}\left(1 + \frac{2}{3}\right) \text{ and } \text{REF}- = \text{VBG}\left(1 - \frac{2}{3}\right)$$

If the built-in reference is used, VBG equals 1.5 V which results in REF+ = 2.5 V, REF– = 0.5 V and $\Delta$REF = 2 V.

The internal reference can be disabled by writing 1 to D12 (REF) in the control register (address 3). The band gap reference is then disconnected and can be substituted by a voltage on the VBG pin.

### programmable gain amplifier

The on-chip programmable gain amplifier (PGA) has eight gain settings. The gain can be changed by writing to the PGA gain register (address 1). The range is 0 to 7dB in steps of one dB.

### out of range indication

The OV output of the ADC indicates an out of range condition. Every time the difference on the analog inputs exceeds the differential reference, this signal is asserted. This signal is updated the same way as the digital data outputs and therefore subject to the same pipeline delay.

### offset compensation

With the offset register it is possible to automatically compensate system offset errors, including errors caused by additional signal conditioning circuitry. If the offset compensation is enabled (D7 (OFF) in the control register), the value in the offset register (address 2) is automatically subtracted from the output of the ADC.

In order to set the correct value of the offset compensation register, the ADC result when the input signal is 0 must be read by the host processor and written to the offset register (address 2).

### test modes

The ADC core operation can be tested by selecting one of the available test modes (see control register description). The test modes apply various voltages to the differential input depending on the setting in the control register.

### digital I/O

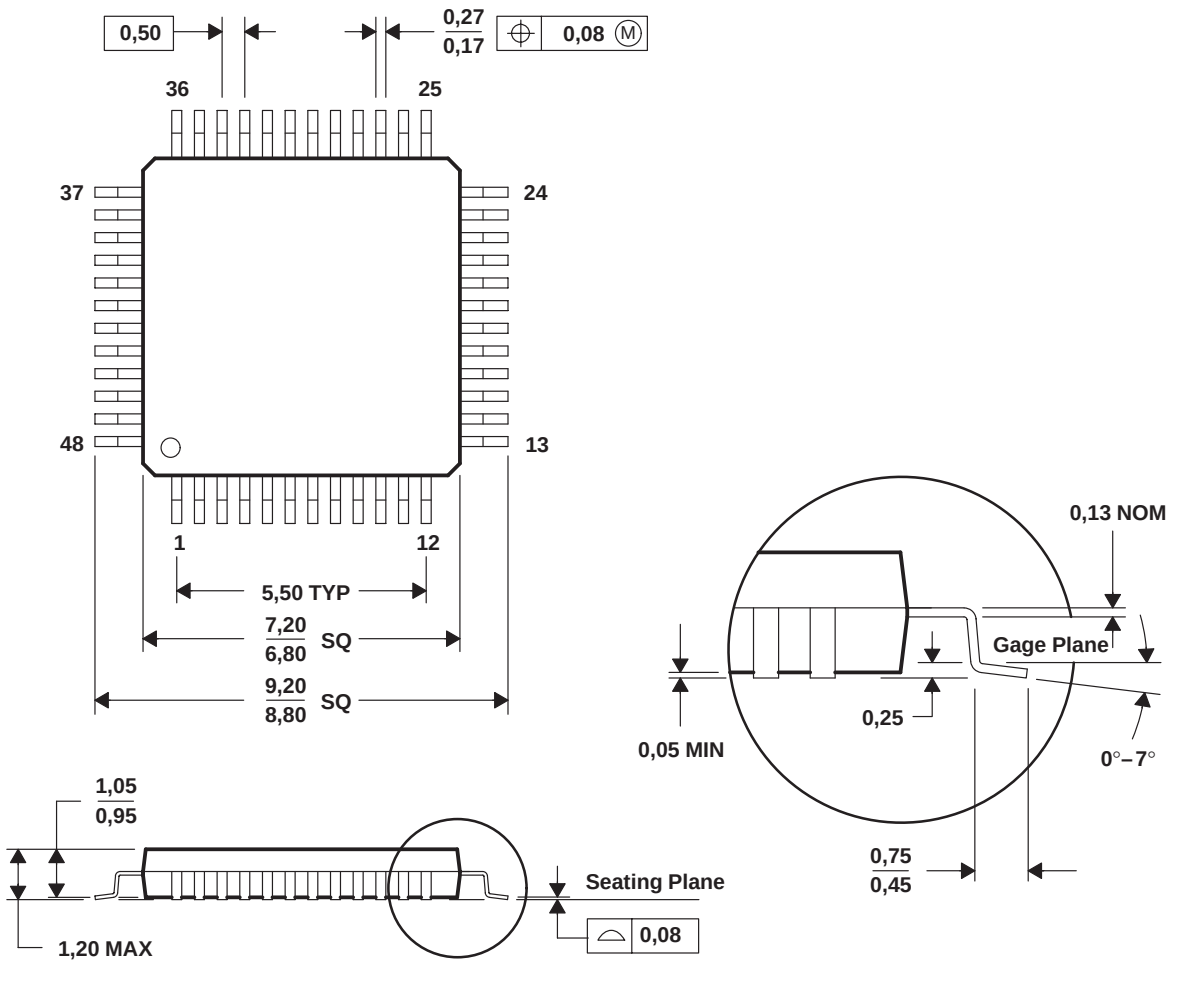
The digital inputs and outputs of the THS14F01/3 ADC are 3-V CMOS compatible. In order to avoid current feed back errors, the capacitive load on the digital outputs should be as low as possible (50 pF max). Series resistors (100 $\Omega$) on the digital outputs can improve the performance by limiting the current during output transitions.

The parallel interface of the THS14F01/3 ADC features 3-state buffers, making it possible to directly connect it to a data bus. The output buffers are enabled by driving the $\overline{\text{OE}}$ input low.

Refer to the read and write timing diagrams in the parameter measurement information section for information on read and write access.

## MECHANICAL DATA

**PFB (S-PQFP-G48)** **PLASTIC QUAD FLATPACK**




NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Falls within JEDEC MS-026

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