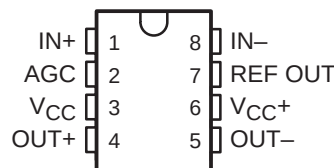


TL026C DIFFERENTIAL HIGH-FREQUENCY AMPLIFIER WITH AGC

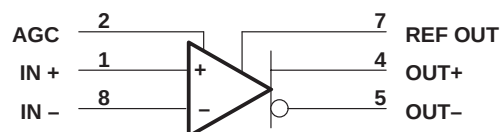
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- Low Output Common-Mode Sensitivity to AGC Voltages
- Input and Output Impedances Independent of AGC Voltage
- Peak Gain . . . 38 dB Typ
- Wide AGC Range . . . 50 dB Typ
- 3-dB Bandwidth . . . 50 MHz
- Other Characteristics Similar to NE592 and uA733

D OR P PACKAGE
(TOP VIEW)



symbol



description

This device is a monolithic two-stage high-frequency amplifier with differential inputs and outputs.

Internal feedback provides wide bandwidth, low phase distortion, and excellent gain stability. Variable gain based on signal summation provides large AGC control over a wide bandwidth with low harmonic distortion. Emitter-follower outputs enable the device to drive capacitive loads. All stages are current-source biased to obtain high common-mode and supply-voltage rejection ratios. The gain may be electronically attenuated by applying a control voltage to the AGC pin. No external compensation components are required.

This device is particularly useful in TV and radio IF and RF AGC circuits, as well as magnetic-tape and disk-file systems where AGC is needed. Other applications include video and pulse amplifiers where a large AGC range, wide bandwidth, low phase shift, and excellent gain stability are required.

The TL026C is characterized for operation from 0°C to 70°C.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)[†]

Supply voltage, V_{CC+} (see Note 1)	8 V
Supply voltage, V_{CC-} (see Note 1)	– 8 V
Differential input voltage	±5 V
Common-mode input voltage	±6 V
Output current	±10 mA
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range	0°C to 70°C
Storage temperature range	– 65°C to 150°C
Lead temperature range 1,6 mm (1/16 inch) from case for 10 seconds	260°C

[†] Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at these or any other conditions beyond those indicated in the recommended operating conditions section of this specification is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltages are with respect to the midpoint of V_{CC+} and V_{CC-} except differential input and output voltages.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	OPERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
P	1000 mW	8.0 mW/°C	640 mW

TL026C

DIFFERENTIAL HIGH-FREQUENCY AMPLIFIER WITH AGC

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recommended operating conditions

	MIN	NOM	MAX	UNIT
Supply voltage, V_{CC+}	3	6	8	V
Supply voltage, V_{CC-}	-3	-6	-8	V
Operating free-air temperature range, T_A	0		70	°C

electrical characteristics at 25°C operating free-air temperature, $V_{CC+} = \pm 6$ V, $V_{AGC} = 0$, REF OUT pin open (unless otherwise specified)

PARAMETER	FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
A_{VD} Large-signal differential voltage amplification	1	$V_{O(PP)} = 3$ V, $R_L = 2$ k Ω	65	85	105	V/V
ΔA_{VD} Change in voltage amplification	1	$V_{IPP} = 28.5$ mV, $R_L = 2$ k Ω , $V_{AGC} - V_{ref} = \pm 180$ mV		-50		dB
V_{ref} Voltage at REF OUT		$I_{ref} = -1$ mA to 100 μ A	1.3		1.5	V
BW Bandwidth (-3 dB)	2	$V_{O(PP)} = 1$ V, $V_{AGC} - V_{ref} = \pm 180$ mV		50		MHz
I_{IO} Input offset current				0.4	5	μ A
I_{IB} Input bias current				10	30	μ A
V_{ICR} Common-mode input voltage range	3		± 1			V
V_{OC} Common-mode output voltage	1	$R_L = \infty$	3.25	3.75	4.25	V
ΔV_{OC} Change in common-mode output voltage	1	$V_{AGC} = 0$ to 2 V, $R_L = \infty$			300	mV
V_{OO} Output offset voltage	1	$V_{ID} = 0$, $R_L = \infty$			0.75	V
$V_{O(PP)}$ Maximum peak-to-peak output voltage swing	1	$R_L = 2$ k Ω	3	4		V
r_i Input resistance at AGC, IN+, or IN -			10	30		k Ω
r_o Output resistance				20		Ω
$CMRR$ Common-mode rejection ratio	3	$V_{IC} = \pm 1$ V, $f = 100$ kHz	60	86		dB
		$V_{IC} = \pm 1$ V, $f = 5$ mHz		60		
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC+} / \Delta V_{IO}$)	4	$\Delta V_{CC+} = \pm 0.5$ V, $\Delta V_{CC-} = \pm 0.5$ V	50	70		dB
V_n Broadband equivalent noise voltage	4	BW = 1 kHz to 10 MHz		12		μ V
t_{pd} Propagation delay time	2	$\Delta V_O = 1$ V		6	10	ns
t_r Rise time	2	$\Delta V_O = 1$ V		4.5	12	ns
$I_{sink(max)}$ Maximum output sink current		$V_{ID} = 1$ V, $V_O = 3$ V	3	4		mA
I_{CC} Supply current		No load, No signal		22	27	mA



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TL026C

DIFFERENTIAL HIGH-FREQUENCY AMPLIFIER WITH AGC

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electrical characteristics over recommended operating free-air temperature range, $V_{CC\pm} = \pm 6\text{ V}$, $V_{AGC} = 0$, REF OUT pin open (unless otherwise specified)

PARAMETER	FIGURE	TEST CONDITIONS	MIN	TYP	MAX	UNIT
A_{VD} Large-signal differential voltage amplification	1	$V_{O(PP)} = 3\text{ V}$, $R_L = 2\text{ k}\Omega$	55		115	V/V
I_{IO} Input offset current					6	μA
I_{IB} Input bias current					40	μA
V_{ICR} Common-mode input voltage range	3		± 1			V
V_{OO} Output offset voltage	1	$V_{ID} = 0$, $R_L = \infty$			1.5	V
$V_{O(PP)}$ Maximum peak-to-peak output voltage swing	1	$R_L = 2\text{ k}\Omega$	2.8			V
r_i Input resistance at AGC, IN+, or IN –			8			$\text{k}\Omega$
CMRR Common-mode rejection ratio	3	$V_{IC} = \pm 1\text{ V}$, $f = 100\text{ kHz}$	50			dB
k_{SVR} Supply voltage rejection ratio ($\Delta V_{CC+} / \Delta V_{IO}$)	4	$\Delta V_{CC+} = \pm 0.5\text{ V}$, $\Delta V_{CC-} = \pm 0.5\text{ V}$	50			dB
$I_{\text{sink(max)}}$ Maximum output sink current		$V_{ID} = 1\text{ V}$, $V_O = 3\text{ V}$	2.8	4		mA
I_{CC} Supply current	1	No load, No signal			30	mA

PARAMETER MEASUREMENT INFORMATION

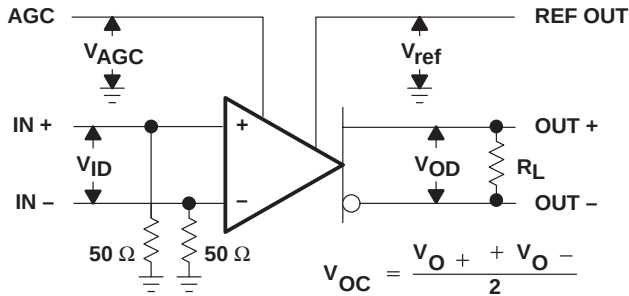


Figure 1. Test Circuit

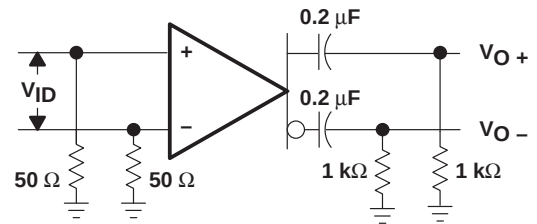


Figure 2. Test Circuit

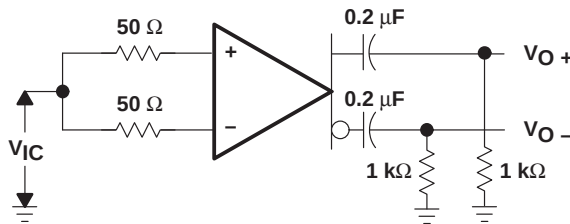


Figure 3. Test Circuit

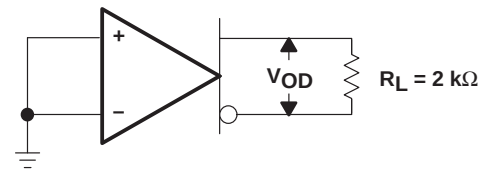


Figure 4. Test Circuit

TL026C

DIFFERENTIAL HIGH-FREQUENCY AMPLIFIER WITH AGC

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TYPICAL CHARACTERISTICS

DIFFERENTIAL VOLTAGE AMPLIFICATION
VS
DIFFERENTIAL GAIN-CONTROL VOLTAGE

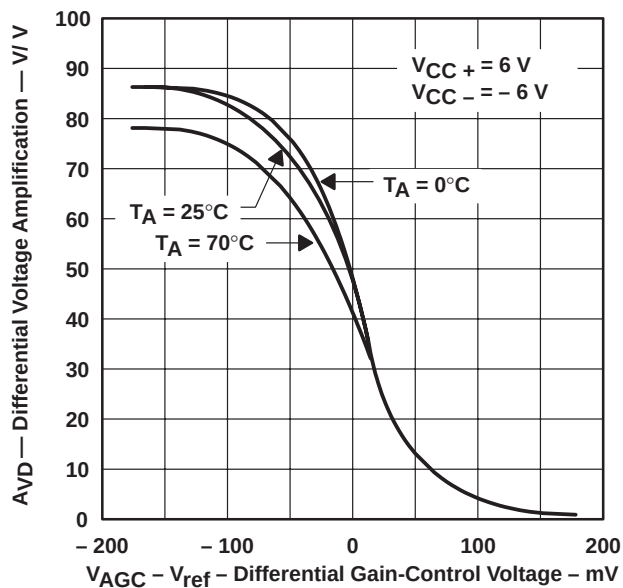


Figure 5

APPLICATION INFORMATION

gain characteristics

Figure 5 shows the differential voltage amplification versus the differential gain-control voltage ($V_{AGC} - V_{ref}$). V_{AGC} is the absolute voltage applied to the A_{GC} input and V_{ref} is the dc voltage at the REF OUT output. As V_{AGC} increases with respect to V_{ref} , the TL026C gain changes from maximum to minimum. As shown in Figure 5 for example, V_{AGC} would have to vary from approximately 180 mV less than V_{ref} to approximately 180 mV greater than V_{ref} to change the gain from maximum to minimum. The total signal change in V_{AGC} is defined by the following equation.

$$\begin{aligned}\Delta V_{AGC} &= V_{ref} + 180 \text{ mV} - (V_{ref} - 180 \text{ mV}) \\ \Delta V_{AGC} &= 360 \text{ mV}\end{aligned}\tag{1}$$

However, because V_{AGC} varies as the ac AGC signal varies and also differentially around V_{ref} , then V_{AGC} should have an ac signal component and a dc component. To preserve the dc and thermal tracking of the device, this dc voltage must be generated from V_{ref} . To apply proper bias to the AGC input, the external circuit used to generate V_{AGC} must combine these two voltages. Figures 6 and 7 show two circuits that will perform this operation and are easy to implement. The circuits use a standard dual operational amplifier for AGC feedback. By providing rectification and the required feedback gain, these circuits are also complete AGC systems.

circuit operation

Amplifier A1 amplifies and inverts the rectified and filtered AGC signal voltage V_C producing output voltage V_1 . Amplifier A2 is a differential amplifier that inverts V_1 again and adds the scaled V_{ref} voltage. This conditioning makes V_{AGC} the sum of the signal plus the scaled V_{ref} . As the signal voltage increases, V_{AGC} increases and the gain of the TL026C is reduced. This maintains a constant output level.

feedback circuit equations

Following the AGC input signal (Figures 6 and 7) from the OUT output through the feedback amplifiers to the AGC input produces the following equations:

1. AC output to diode D1, assuming sinusoidal signals

$$V_O = V_{OP} (\sin wt)\tag{2}$$

where:

$$V_{OP} = \text{peak voltage of } V_O$$

2. Diode D1 and capacitor C1 output

$$V_C = V_{OP} - V_F\tag{3}$$

where:

$$V_F = \text{forward voltage drop of D1}$$

$$V_C = \text{voltage across capacitor C1}$$

3. A1 output

$$V_1 = -\frac{R_2}{R_1} V_C\tag{4}$$

4. A2 output ($R_3 = R_4$)

$$V_{AGC} = \frac{R_2}{R_1} V_C + 2\frac{R_6}{R_5 + R_6} V_{ref}\tag{5}$$

TL026C

DIFFERENTIAL HIGH-FREQUENCY AMPLIFIER WITH AGC

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Amplifier A2 inverts V1 producing a positive AGC signal voltage. Therefore, the input voltage to the TL026C AGC pin consists of an AGC signal equal to:

$$\frac{R2}{R1} V_C \quad (6)$$

and a dc voltage derived from V_{ref} , defined as the quiescent value of V_{AGC} .

$$V_{AGC}(q) = 2 \frac{R6}{R5 + R6} V_{ref} \quad (7)$$

For the initial resistor calculations, V_{ref} is assumed to be typically 1.4 V making quiescent V_{AGC} approximately 1.22 V ($V_{AGC}(q) = V_{ref} - 180$ mV). This voltage allows the TL026C to operate at maximum gain under no-signal and low-signal conditions. In addition, with V_{ref} used as both internal and external reference, its variation from device to device automatically adjusts the overall bias and makes AGC operation essentially independent of the absolute value of V_{ref} . The resistor divider needs to be calculated only once and is valid for the full tolerance of V_{ref} .

output voltage limits (see Figures 6 and 7)

The output voltage level desired must fall within the following limits:

1. Because the data sheet minimum output swing is 3 V peak-to-peak using a 2-k Ω load resistor, the user-selected design limit for the peak output swing should not exceed 1.5 V.
2. The voltage drop of the rectifying diode determines the lower voltage limit. When a silicon diode is used, this voltage is approximately 0.7 V. The output voltage V_O must have sufficient amplitude to exceed the rectifying diode drop. A Schottky diode can be used to reduce the V_O level required.

gain calculations for a peak output voltage of 1 V

A peak output voltage of 1 V was chosen for gain calculations because it is approximately midway between the limits of conditions 1 and 2 in the preceding paragraph.

Using equation 3 ($V_C = V_{OP} - V_d$), V_C is calculated as follows:

$$V_C = 1 \text{ V} - 0.7 \text{ V}$$

$$V_C = 0.3 \text{ V}$$

Therefore, the gain of A1 must produce a voltage V1 that is equal to or greater than the total change in V_{AGC} for maximum TL026C gain change.

With a total change in V_{AGC} of 360 mV and using equation 4, the calculation is as follows:

$$-\frac{V1}{V_C} = \frac{\Delta V_{AGC}}{V_C} = \frac{R2}{R1} = \frac{0.36}{0.3} = 1.2$$

If $R1$ is 10 k Ω , $R2$ is 1.2 times $R1$ or 12 k Ω .

Since the output voltage for this circuit must be between 0.85 V and 1.3 V, the component values in Figures 6 and 7 provide a nominal 1-V peak output limit. This limit is the best choice to allow for temperature variations of the diode and minimum output voltage specification.

APPLICATION INFORMATION

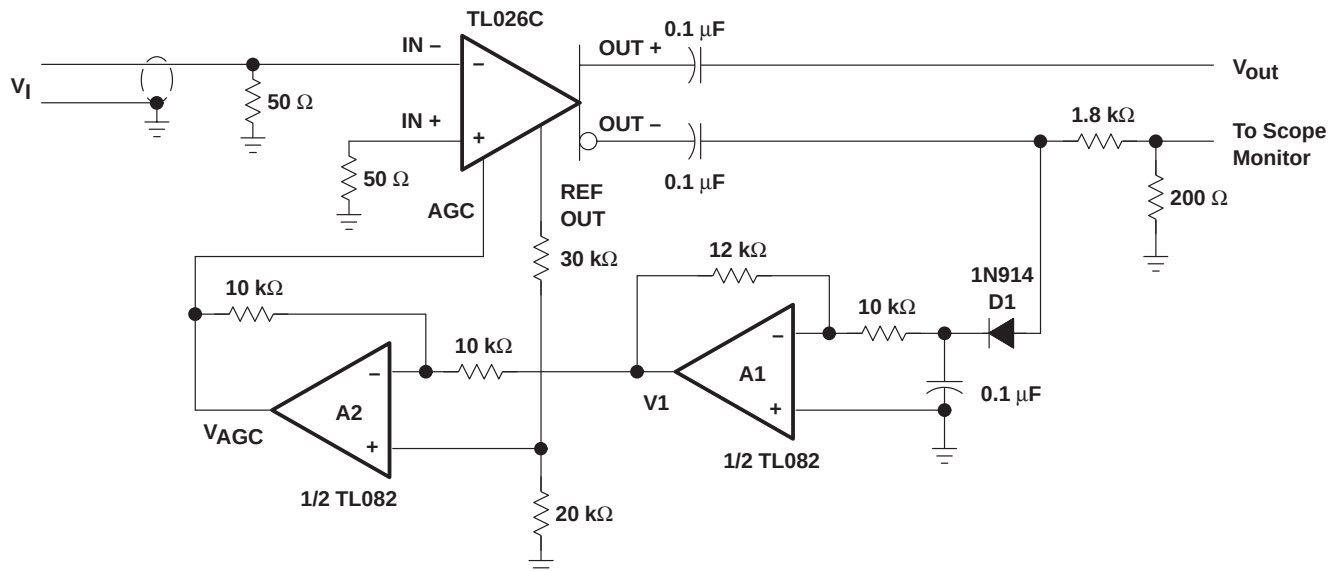
The circuit values in Figures 6 and 7 will produce the best results in this general application. Because of rectification and device input constraints, the circuit in Figure 6 will not provide attenuation and has about 32 dB of control range. The circuit shown in Figure 7 will have approximately 25% variation in the peak output voltage limit due to the variation in gain of the TL592 device to device. In addition, if a lower output voltage is desired, the output of the TL026C can be used for approximately 40 mV of controlled signal.

considerations for the use of the TL026C

To obtain the most reliable results, RF breadboarding techniques must be used. A groundplane board should be used and power supplies should be bypassed with 0.1- μ F capacitors. Input leads and output leads should be as short as possible and separated from each other.

A peak input voltage greater than 200 mV will begin to saturate the input stages of the TL026C and, while the circuit is in the AGC mode, the output signal may become distorted.

To observe the output signal of TL026C or TL592, low-capacitance FET probes or the output voltage divider technique shown in Figure 6 should be used.



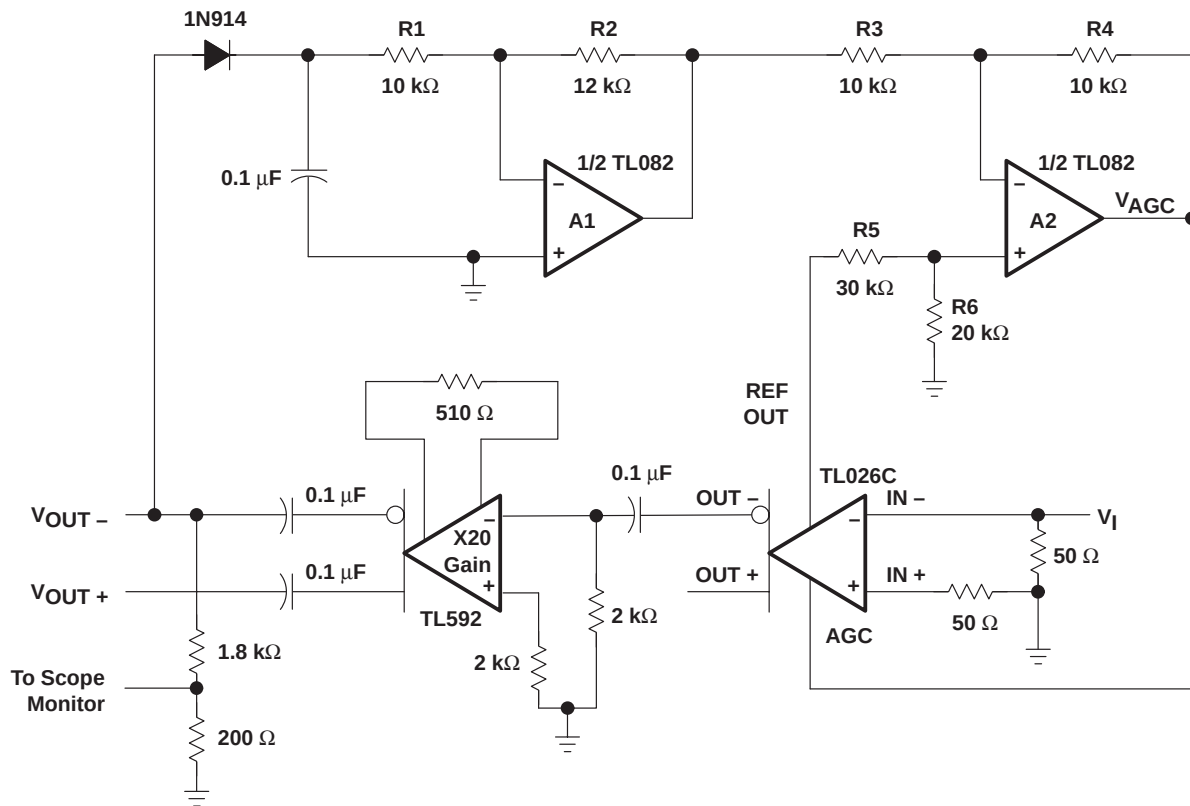
NOTE: $V_{CC+} = 6\text{ V}$ and $V_{CC-} = -6\text{ V}$ for TL026C and amplifiers A1 and A2.

Figure 6. Typical Application Circuit With No Attenuation

DIFFERENTIAL HIGH-FREQUENCY AMPLIFIER WITH AGC

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NOTE: $V_{CC+} = 6\text{ V}$ and $V_{CC-} = -6\text{ V}$ for TL026C and amplifiers A1 and A2.

Figure 7. Typical Application Circuit With Attenuation

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TL026CD	Active	Production	SOIC (D) 8	75 TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C
TL026CD.A	Active	Production	SOIC (D) 8	75 TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C
TL026CDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C
TL026CDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	0 to 70	TL026C
TL026CP	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TL026CP
TL026CP.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TL026CP
TL026CPSR	Active	Production	SO (PS) 8	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	T026
TL026CPSR.A	Active	Production	SO (PS) 8	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	T026

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

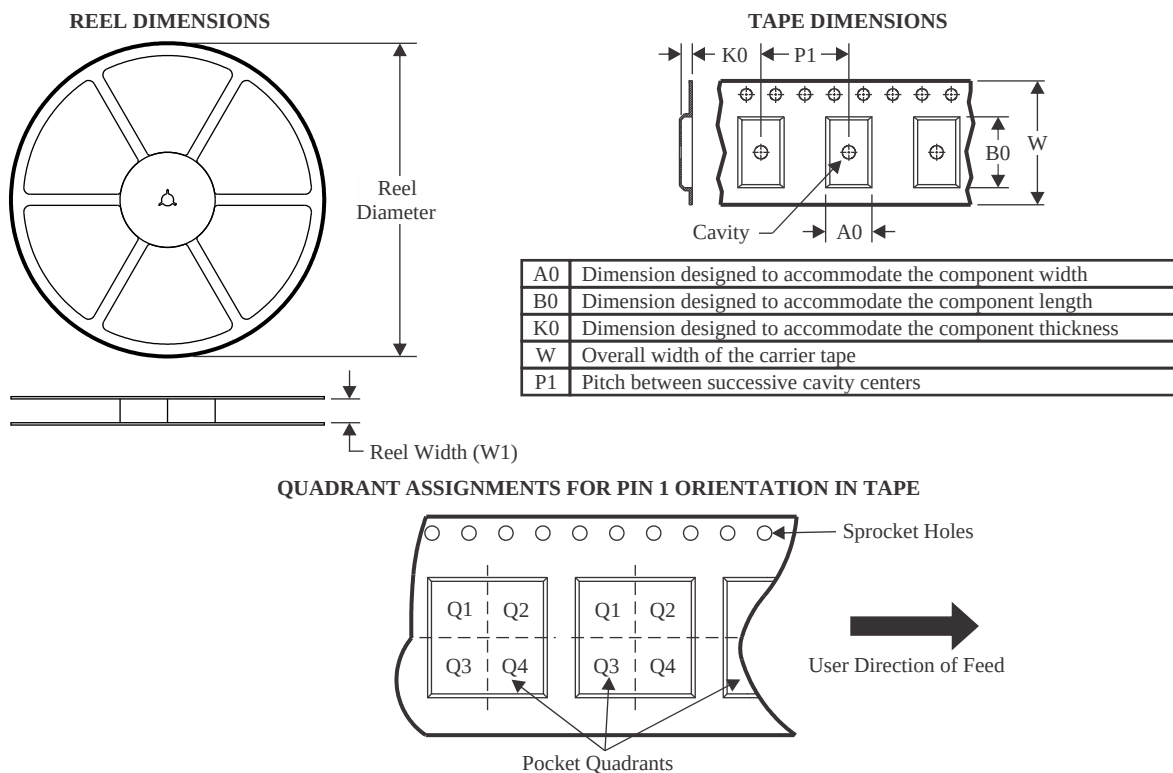
⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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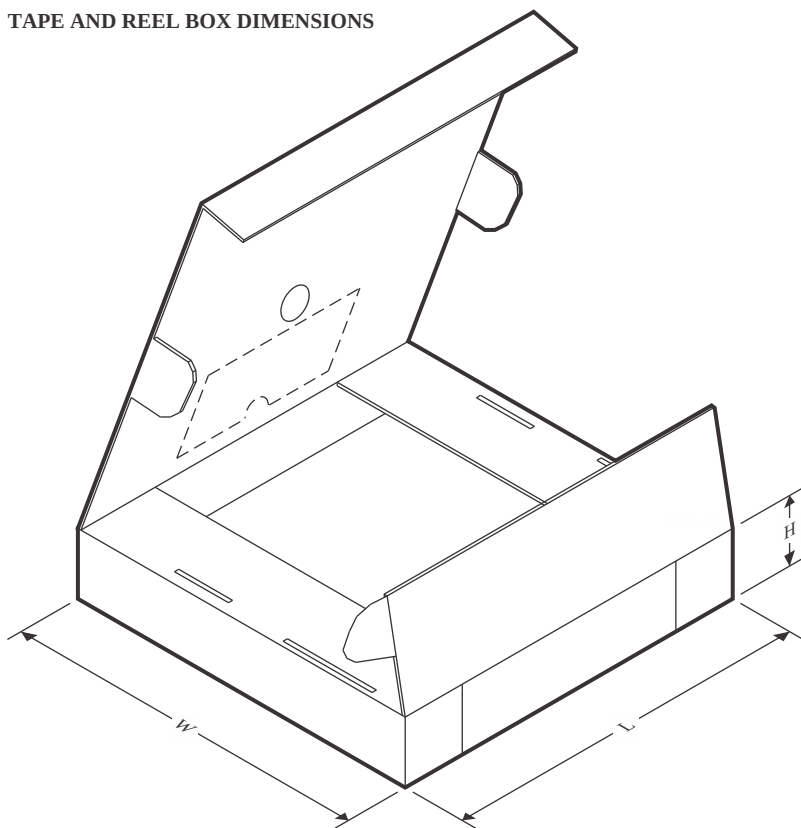
TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TL026CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TL026CPSR	SO	PS	8	2000	330.0	16.4	8.35	6.6	2.4	12.0	16.0	Q1

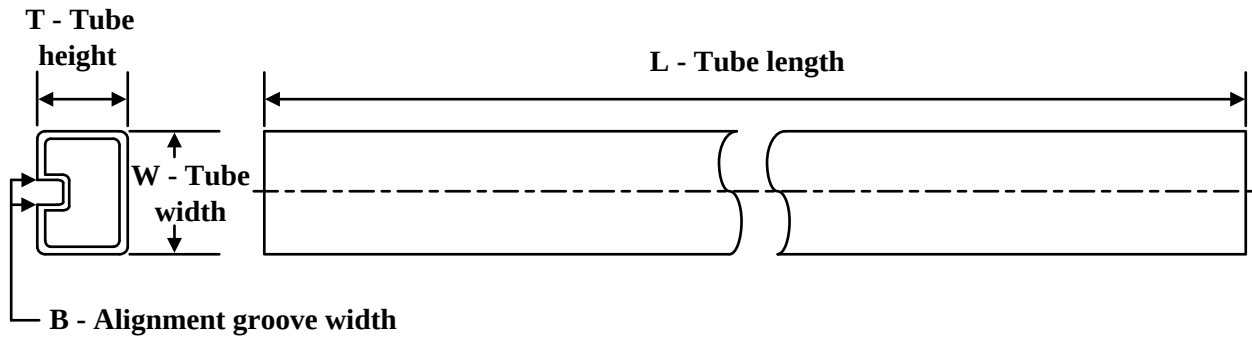
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TL026CDR	SOIC	D	8	2500	353.0	353.0	32.0
TL026CPSR	SO	PS	8	2000	353.0	353.0	32.0

TUBE



*All dimensions are nominal

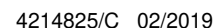
Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
TL026CD	D	SOIC	8	75	507	8	3940	4.32
TL026CD.A	D	SOIC	8	75	507	8	3940	4.32
TL026CP	P	PDIP	8	50	506	13.97	11230	4.32
TL026CP.A	P	PDIP	8	50	506	13.97	11230	4.32



PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.

2. This drawing is subject to change without notice.

3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.

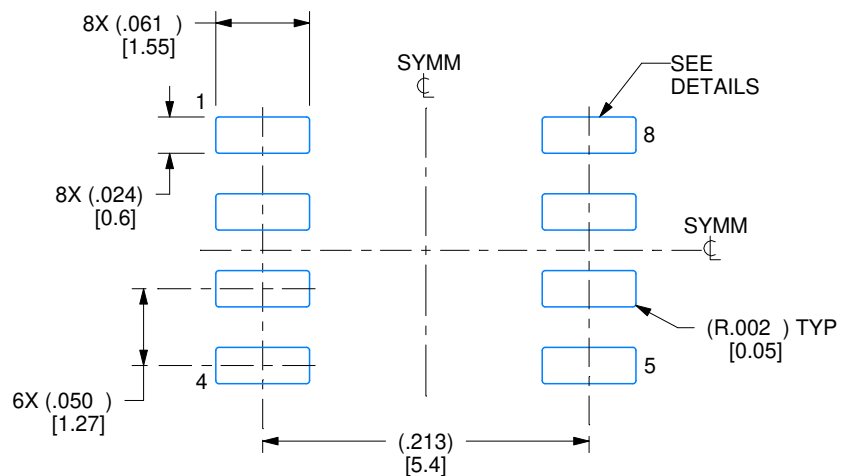
4. This dimension does not include interlead flash.

5. Reference JEDEC registration MS-012, variation AA.

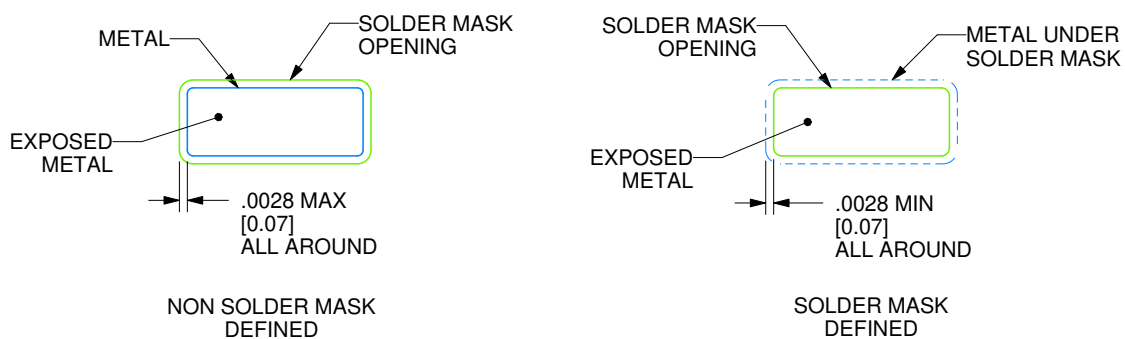
D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

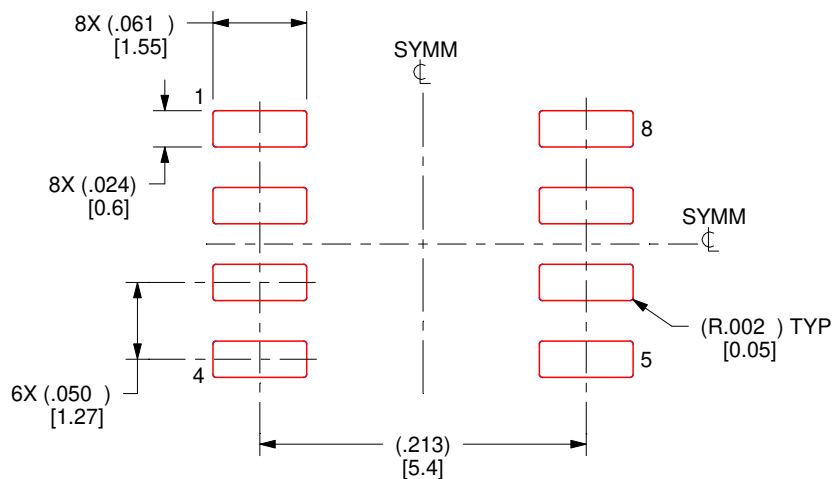
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.125 MM] THICK STENCIL
SCALE:8X

4214825/C 02/2019

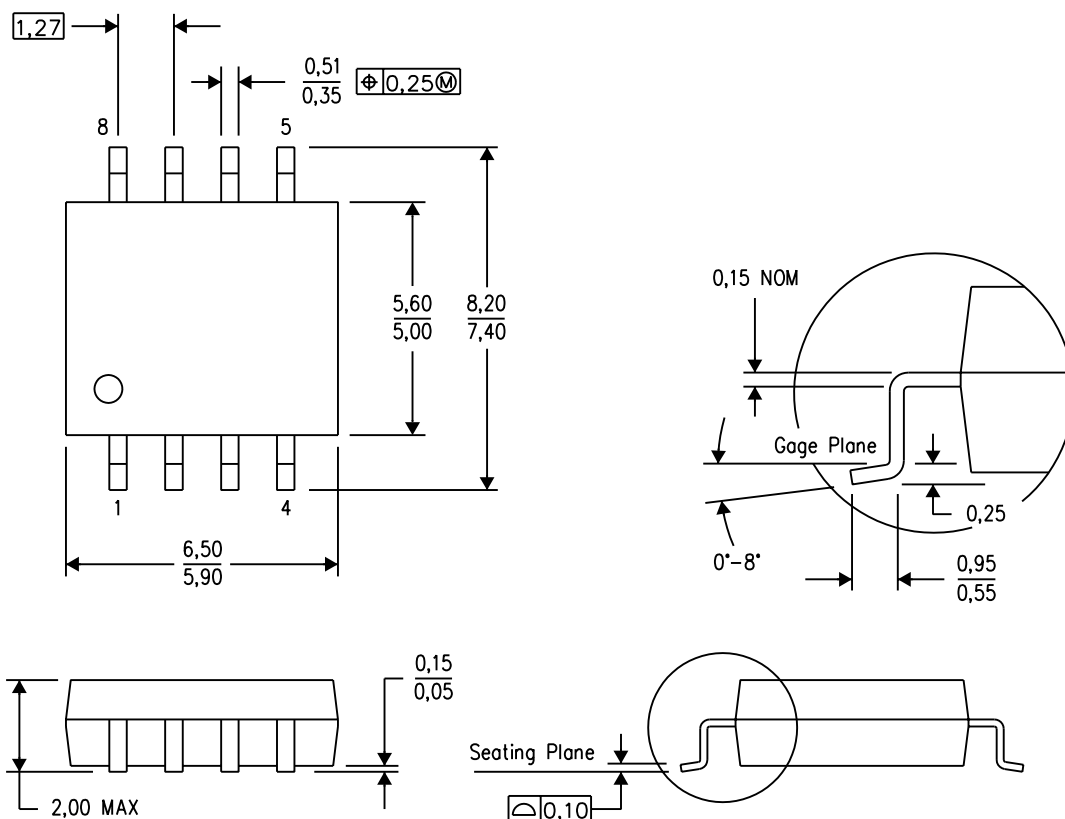
NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

MECHANICAL DATA

PS (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE

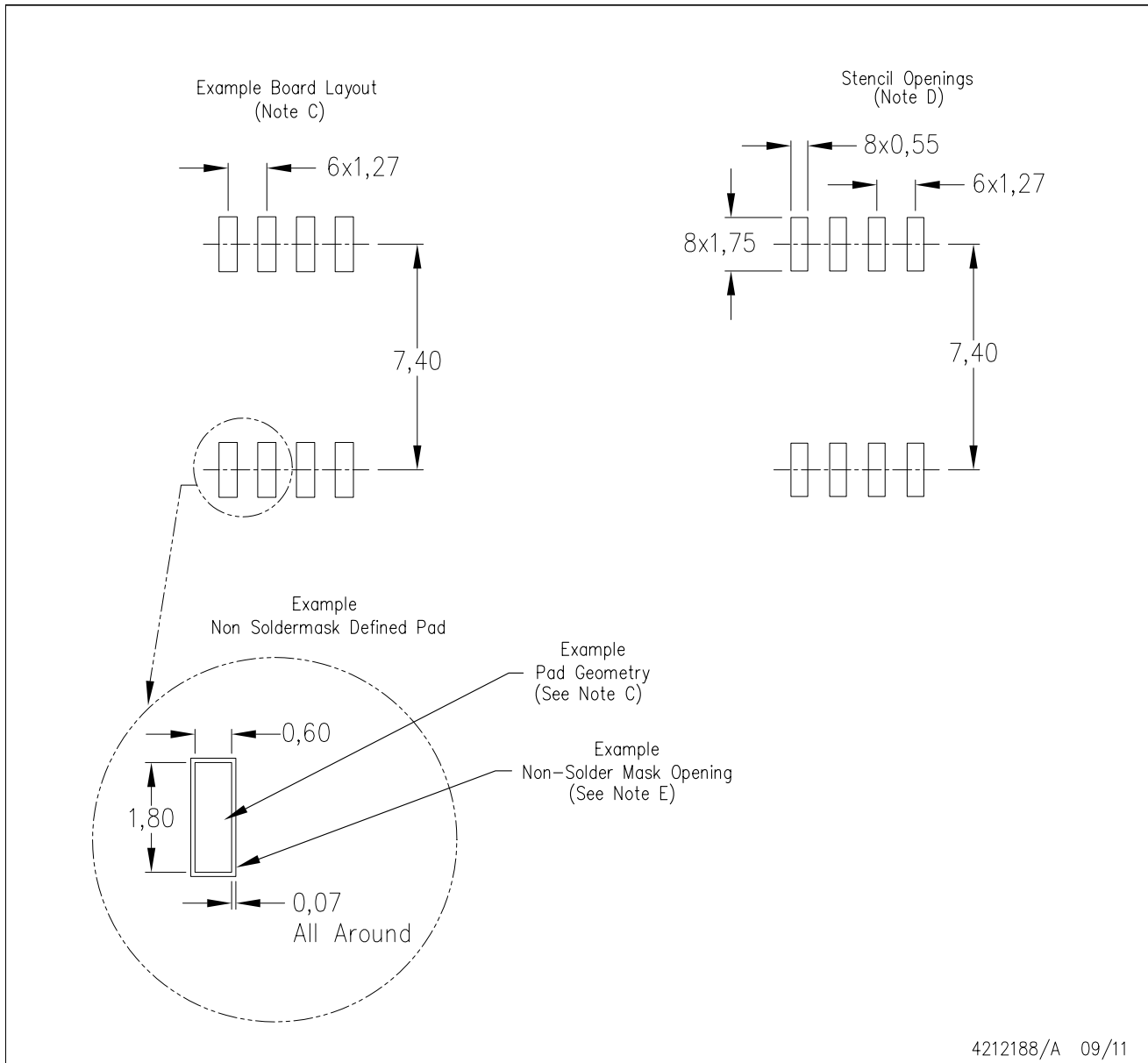


4040063/C 03/03

- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.

PS (R-PDSO-G8)

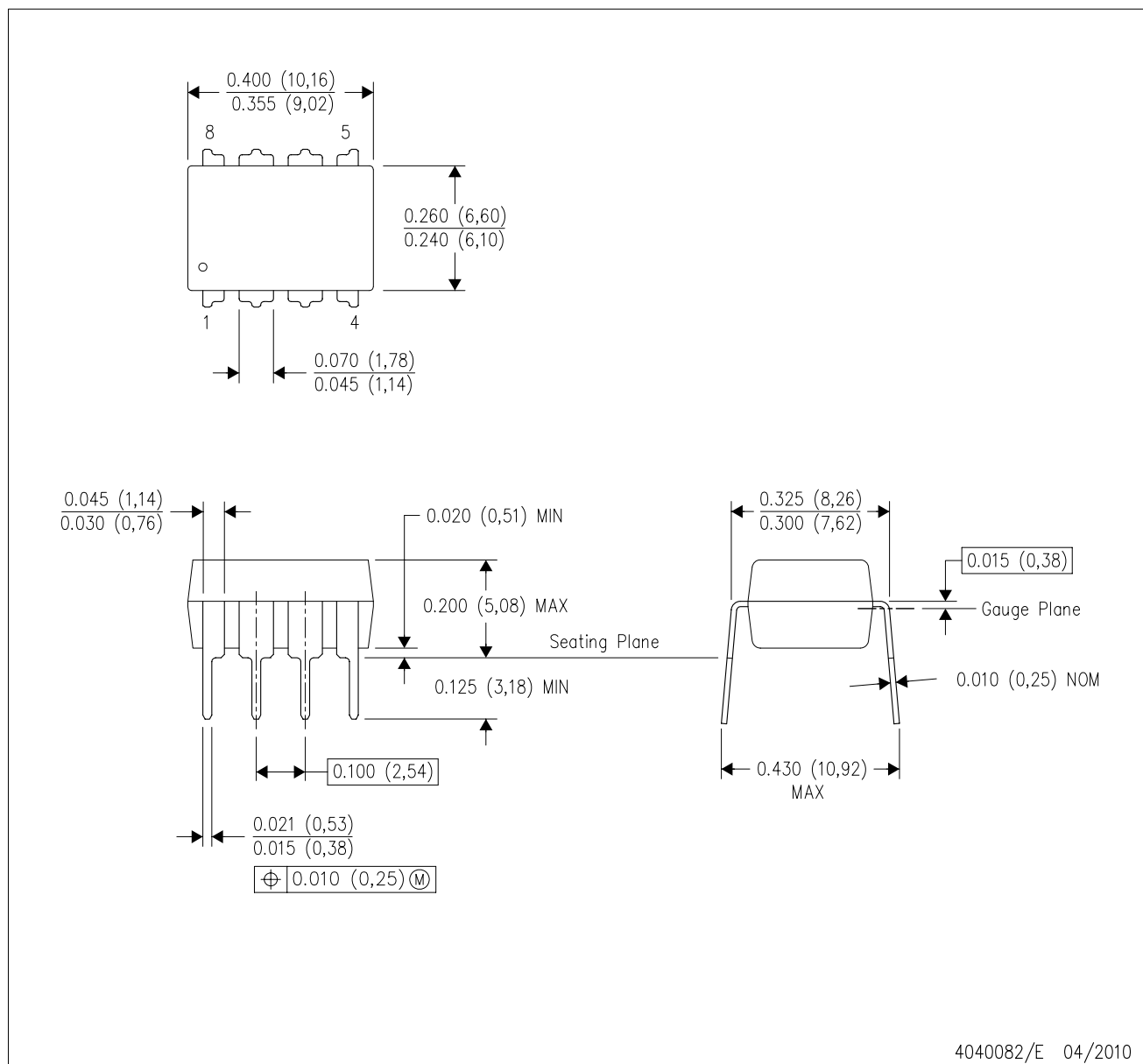
PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Publication IPC-7351 is recommended for alternate designs.
 - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
 - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MS-001 variation BA.

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